

## 1.25 GHz Dual Integrated DCL with PPMU, Level Setting DACs, and On-Chip Calibration Registers

### FEATURES

- ▶ 1.25 GHz, 2.5 Gbps data rate
- ▶ 3-level driver with high-Z and reflection clamps
- ▶ Window and differential comparators
- ▶  $\pm 25$  mA active load
- ▶ Per pin parametric measurement unit (PMU) with a  $-1.5$  V to  $+4.5$  V range
- ▶ Low leakage mode (typically  $<5$  nA)
- ▶ Integrated 16-bit DACs with offset and gain correction
- ▶ 1.2 W power dissipation per channel (ADATE320)
- ▶ 1.3 W power dissipation per channel (ADATE320-1)
- ▶ Driver
  - ▶ Voltage range:  $-1.5$  V to  $+4.5$  V
  - ▶ Precision trimmed termination:  $50.0 \Omega$
  - ▶ Unterminated swing: 50 mV minimum to 6.0 V maximum
  - ▶ 400 ps minimum pulse width, 1.0 V programmed swing
  - ▶ 25 ps deterministic jitter
- ▶ Comparator
  - ▶ Differential and single-ended window modes
  - ▶ 100 ps equivalent input rise/fall time (ERT/EFT)
  - ▶ 250 mV current mode logic (CML) outputs (ADATE320)
  - ▶ 400 mV CML outputs (ADATE320-1)
- ▶ Load
- ▶ Per pin PMU (PPMU)
  - ▶ Force voltage/compliance range:  $-1.5$  V to  $+4.5$  V
  - ▶ 5 current ranges
    - ▶  $\pm 40$  mA,  $\pm 1$  mA,  $\pm 100 \mu\text{A}$ ,  $\pm 10 \mu\text{A}$ ,  $\pm 2 \mu\text{A}$
  - ▶ Dedicated go/no-go comparators
- ▶ DC levels
  - ▶ Fully integrated and dedicated 16-bit DACs
  - ▶ On-chip gain and offset calibration registers with automatic add/multiply function
- ▶ 84-lead, 10 mm  $\times$  10 mm LFCSP (0.4 mm pitch)

### APPLICATIONS

- ▶ Automatic test equipment (ATE)
- ▶ Semiconductor/board test systems
- ▶ Instrumentation and characterization equipment

### GENERAL DESCRIPTION

The ADATE320 is a complete, single-chip ATE solution that performs the pin electronics functions of a driver, comparator, and active load (DCL), and a four quadrant per pin parametric measurement unit (PPMU). Dedicated 16-bit digital-to-analog converters (DACs) with on-chip calibration registers provide all the necessary dc levels for operation of the device.

The driver features three active modes: high, low, and terminate, as well as a high impedance inhibit state. The inhibit state, in conjunction with the integrated dynamic clamps, facilitates significant attenuation of transmission line reflections when the driver is not actively terminating the line. The open-circuit drive capability is  $-1.5$  V to  $+4.5$  V to accommodate a standard range of ATE and instrumentation applications.

The ADATE320 can be used as a dual, single-ended pin electronics channel or as a single differential channel. In addition to per channel high speed window comparators, the ADATE320 provides a programmable threshold differential comparator for differential ATE applications.

All dc levels for DCL and PPMU functions are generated by dedicated, on-chip, 16-bit DACs. To facilitate the programming of accurate levels, the ADATE320 includes an integrated calibration function to correct for the gain and offset errors of each functional block. Correction coefficients can be stored on chip, and any values written to the DACs adjust automatically using the appropriate correction factors.

The ADATE320 uses a serial programmable interface (SPI) bus to program all functional blocks, DACs, and on-chip calibration constants. It also has an on-chip temperature sensor and overvoltage/undervoltage fault clamps that monitor and report the device temperature and any output pin or transient PPMU voltage faults that may occur during operation.

The ADATE320 is available in two options. The standard option has high speed comparator outputs with 250 mV output swing. The ADATE320-1 has 400 mV output swing. See the [Ordering Guide](#) for more information.

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## REVISION HISTORY

### 6/2024—Rev. D to Rev. E

Change to Temperature Sensor Gain Parameter, Table 11.....	19
Changes to Evaluation Boards.....	78

### 4/2024—Rev. C to Rev. D

Changes to Figure 143.....	67
Updated Outline Dimensions.....	78
Added Evaluation Boards.....	78

### 9/2022—Rev. B to Rev. C

Deleted Figure 136; Renumbered Sequentially.....	64
Changes to Figure 136.....	64



## SPECIFICATIONS

$V_{CC} = 8.0\text{ V}$ ,  $V_{DD} = 1.8\text{ V}$ ,  $V_{EE} = -5.0\text{ V}$ ,  $V_{TTCx} = V_{TTDx} = 1.2\text{ V}$ ,  $V_{REF} = 2.500\text{ V}$ ,  $V_{REFGND} = 0.000\text{ V}$ . All default test conditions are as defined in Table 30. All specified values are at  $T_J = 60^\circ\text{C}$ , where  $T_J$  corresponds to the typical internal temperature sensor reading (V\_THERM pin), unless otherwise noted. Temperature coefficients are measured around  $T_J = 40^\circ\text{C}$ ,  $60^\circ\text{C}$ ,  $80^\circ\text{C}$ , and  $100^\circ\text{C}$ . Typical values are based on the statistical mean of the design, simulation analyses, and/or limited bench evaluation data. Typical values are neither tested nor guaranteed. Test level codes are defined in the [Explanation of Test Levels](#) section.

## ELECTRICAL SPECIFICATIONS

Table 1.

Parameter	Min	Typ	Max	Unit	Test Level	Test Conditions/Comments
<b>DUTx PIN CHARACTERISTICS</b>						
Output Leakage Current						
DCL Disable						
PPMU Range E	-10.0	+5.0	+10.0	nA	P	$-1.5\text{ V} < V_{DUTx} < +4.5\text{ V}$ , PPMU and DCL disabled, PPMU Range E, $V_{CLx} = -2.5\text{ V}$ , $V_{CHx} = +7.5\text{ V}$
PPMU Range A to Range D		5.0		nA	C <sub>T</sub>	$-1.5\text{ V} < V_{DUTx} < +4.5\text{ V}$ , PPMU and DCL disabled, PPMU Range A, Range B, Range C, and Range D, $V_{CLx} = -2.5\text{ V}$ , $V_{CHx} = +7.5\text{ V}$
Driver High-Z Mode	-0.4		+0.4	μA	P	$-1.5\text{ V} < V_{DUTx} < +4.5\text{ V}$ , PPMU disabled and DCL enabled, RCV active, $V_{CLx} = -2.5\text{ V}$ , $V_{CHx} = +7.5\text{ V}$
Capacitance		0.4		pF	S	Drive $V_{ITx} = 0.0\text{ V}$
Voltage Range	-1.5		+4.5	V	D	
<b>POWER SUPPLIES</b>						
Positive DCL Supply, $V_{CC}$	7.6	8.0	8.4	V	D	Power measured with the DUTx pin high-Z, 10 K to 0.0 V Defines dc power supply rejection (PSR) conditions
Negative DCL Supply, $V_{EE}$	-5.25	-5.0	-4.75	V	D	Defines dc PSR conditions
Digital Supply, $V_{DD}$	1.7	1.8	1.9	V	D	
Comparator Termination, $V_{TTCx}$	0.5	1.2	1.8	V	D	$V_{TTC0}$ is not electrically connected to $V_{TTC1}$
Driver Termination, $V_{TTDx}$	0.0	1.2	1.8	V	D	$V_{TTD0}$ is not electrically connected to $V_{TTD1}$
Positive DCL Supply Current, $I_{CC}$						Load and PPMU power-down
ADATE320	145	169	185	mA	P	
ADATE320-1	145	169	185	mA	P	
Negative DCL Supply Current, $I_{EE}$						Load and PPMU power-down
ADATE320	190	222	235	mA	P	
ADATE320-1	220	247	265	mA	P	
Digital Core Supply Current, $I_{DD}$	-125	+10	+125	μA	P	Quiescent (SPI is static)
Comparator Termination Supply Current, $V_{TTCx}$						$0.5\text{ V} \leq V_{TTCx} \leq 1.8\text{ V}$
ADATE320		41		mA	C <sub>T</sub>	
ADATE320-1		66		mA	C <sub>T</sub>	
Driver Termination Supply Current, $V_{TTDx}$		0		mA	C <sub>T</sub>	$0.0\text{ V} \leq V_{TTDx} \leq 1.8\text{ V}$ , $(\overline{\text{DATx}} + \overline{\text{DATx}})/2 = (\overline{\text{RCVx}} + \overline{\text{RCVx}})/2 = V_{TTDx}$
Total Power Dissipation						Load and PPMU power-down
ADATE320	2.10	2.52	2.75	W	P	
ADATE320-1	2.25	2.66	2.90	W	P	

## SPECIFICATIONS

## DRIVER SPECIFICATIONS

$V_{IH} - V_{IL} \geq 100$  mV to meet dc and ac performance specifications.

Table 2.

Parameter	Min	Typ	Max	Unit	Test Level	Test Conditions/Comments
<b>DC SPECIFICATIONS</b>						
High Speed Differential Input Characteristics						
High Speed Input Termination Resistance: DATx/DAT $\bar{x}$ , RCVx/RCV $\bar{x}$	48	50	52	$\Omega$	P	Impedance between VT $\bar{T}$ Dx and respective DATx and RCVx pins; force 4 mA into each pin, measure voltage from VT $\bar{T}$ Dx; calculate resistance ( $\Delta V/\Delta I$ )
Input Voltage Range: DATx/DAT $\bar{x}$ , RCVx/RCV $\bar{x}$	0.0		1.8	V	P <sub>F</sub>	
Input Voltage Differential	0.2	0.4	1.0	V	P <sub>F</sub>	DATx - DAT $\bar{x}$  ,  RCVx - RCV $\bar{x}$
Output Characteristics						
Output Range						
High, V <sub>IH</sub>	-1.4		+4.5	V	D	
Low, V <sub>IL</sub>	-1.5		+4.4	V	D	
Output Term Range, V <sub>IT</sub>	-1.5		+4.5	V	D	
Functional Amplitude (V <sub>IH</sub> - V <sub>IL</sub> )	0.05		6.0	V	D	
DC Output Current Limit						
Source	75		120	mA	P	Drive high, V <sub>IH</sub> = 4.5 V, V <sub>DUTx</sub> = -2.0 V, measure current
Sink	-120		-75	mA	P	Drive low, V <sub>IL</sub> = -1.5 V, V <sub>DUTx</sub> = 5.0 V, measure current
Output Resistance, $\pm 40$ mA	46	48.5	52	$\Omega$	P	$\Delta V_{DUTx}/\Delta I_{DUTx}$ ; source: V <sub>IHx</sub> = 3.0 V, I <sub>DUTx</sub> = 1 mA, 40 mA; sink: V <sub>IL</sub> = 0.0 V, I <sub>DUTx</sub> = -1 mA, -40 mA
<b>DC ACCURACY</b>						
V <sub>IH</sub> , V <sub>IL</sub> , V <sub>IT</sub>						
Offset Error	-500		+500	mV	P	Measured at DAC Code 0x4000 (0.0 V), uncalibrated
Offset Temperature Coefficient (TC)		$\pm 200$		$\mu V/^\circ C$	C <sub>T</sub>	
Gain	1.0		1.1	V/V	P	Gain derived from measurement at DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V); based on an ideal DAC transfer function (see Table 24)
Gain TC		$\pm 50$		ppm/ $^\circ C$	C <sub>T</sub>	
Differential Nonlinearity (DNL)		$\pm 250$		$\mu V$	C <sub>T</sub>	After two-point gain/offset calibration; calibration points at 0x4000 (0.0 V) output; 0x8CCC (3.0 V); measured over full specified output range
Integral Nonlinearity (INL) Focused Range	-5		+5	mV	P	After two-point gain/offset calibration; calibration points at 0x4000 (0.0 V) and 0x8CCC (3.0 V); measured over -0.5 V to +3.5 V output range
INL Full Range	-20		+20	mV	P	After two-point gain/offset calibration; calibration points at 0x4000 (0.0 V) and 0x8CCC (3.0 V); measured over full specified output range
Resolution		153		$\mu V$	D	
DUTGND Voltage Accuracy	-5	$\pm 1$	+5	mV	P	Over $\pm 0.1$ V range; measured over -0.5 V to +3.5 V focused driver output range
DC Levels Interaction						
V <sub>IH</sub> vs. V <sub>IL</sub>		$\pm 1.0$		mV	C <sub>T</sub>	Monitor interaction on V <sub>IH</sub> = +4.5 V; sweep V <sub>IL</sub> = -1.5 V to +4.4 V, V <sub>IT</sub> = +1.0 V
V <sub>IH</sub> vs. V <sub>IT</sub>		$\pm 1.0$		mV	C <sub>T</sub>	Monitor interaction on V <sub>IH</sub> = +4.5 V; sweep V <sub>IT</sub> = -1.5 V to +4.5 V, V <sub>IL</sub> = 0.0 V
V <sub>IL</sub> vs. V <sub>IH</sub>		$\pm 1.0$		mV	C <sub>T</sub>	Monitor interaction on V <sub>IL</sub> = -1.5 V; sweep V <sub>IH</sub> = -1.4 V to +4.5 V, V <sub>IT</sub> = +1.0 V

## SPECIFICATIONS

Table 2. (Continued)

Parameter	Min	Typ	Max	Unit	Test Level	Test Conditions/Comments
VIL vs. VIT		±1.0		mV	C <sub>T</sub>	Monitor interaction on VIL = -1.5 V; sweep VIT = -1.5 V to +4.5 V, VIH = +2.0 V
VIT vs. VIH		±2.0		mV	C <sub>T</sub>	Monitor interaction on VIT = 1.0 V; sweep VIH = -1.4 V to +4.5 V, VIL = -1.5 V
VIT vs. VIL		±2.0		mV	C <sub>T</sub>	Monitor interaction on VIT = 1.0 V; sweep VIL = -1.5 V to +4.4 V, VIH = +4.5 V
Overall Voltage Accuracy Focused Range		±5		mV	C <sub>T</sub>	VIH - VIL ≥ 100 mV; sum of INL, dc interaction, DUTGND and TC errors over ±5°C, after calibration
VIH, VIL, VIT DC PSR		+15		mV/V	C <sub>T</sub>	Measured at calibration points, see Table 1 for power supply ranges
<b>AC SPECIFICATIONS</b>						
All ac specifications performed after dc calibration						
Toggle DATx, VIL = 0.0 V, terminated						
<b>Rise/Fall Times</b>						
0.2 V Programmed Swing						
t <sub>RISE</sub>		150		ps	C <sub>B</sub>	20% to 80%, VIH = 0.2 V
t <sub>FALL</sub>		170		ps	C <sub>B</sub>	20% to 80%, VIH = 0.2 V
0.5 V Programmed Swing						
t <sub>RISE</sub>		150		ps	C <sub>B</sub>	20% to 80%, VIH = 0.5 V
t <sub>FALL</sub>		170		ps	C <sub>B</sub>	20% to 80%, VIH = 0.5 V
1.0 V Programmed Swing						
t <sub>RISE</sub>		150		ps	C <sub>B</sub>	20% to 80%, VIH = 1.0 V
t <sub>FALL</sub>		170		ps	C <sub>B</sub>	20% to 80%, VIH = 1.0 V
2.0 V Programmed Swing						
t <sub>RISE</sub>	120	160	230	ps	P	20% to 80%, VIH = 2.0 V
t <sub>FALL</sub>	120	180	230	ps	P	20% to 80%, VIH = 2.0 V
4.0 V Programmed Swing						
t <sub>RISE</sub>		320		ps	C <sub>B</sub>	10% to 90%, VIH = 4.0 V, unterminated
t <sub>FALL</sub>		320		ps	C <sub>B</sub>	10% to 90%, VIH = 4.0 V, unterminated
t <sub>RISE</sub> to t <sub>FALL</sub> Mismatch		-20		ps	C <sub>B</sub>	t <sub>RISE</sub> - t <sub>FALL</sub> (20% to 80%) within one channel, VIH = 2.0 V, VIL = 0.0 V, terminated
<b>Trailing Edge Timing Error</b>						
Toggle DATx						
Programmed Swing						
VIL = 0.0 V, terminated, 400 ps ≤ pulse width (PW) ≤ 10 ns						
0.2 V		±15		ps	C <sub>B</sub>	VIH = 0.2 V
0.5 V		±15		ps	C <sub>B</sub>	VIH = 0.5 V
1.0 V		±15		ps	C <sub>B</sub>	VIH = 1.0 V
2.0 V		±15		ps	C <sub>B</sub>	VIH = 2.0 V
<b>Maximum Toggle Rate</b>						
Toggle DATx						
VIL = 0.0 V, terminated ≤10% amplitude loss						
0.2 V		2.8		Gbps	C <sub>B</sub>	VIH = 0.2 V
0.5 V		3.2		Gbps	C <sub>B</sub>	VIH = 0.5 V
1.0 V		3.2		Gbps	C <sub>B</sub>	VIH = 1.0 V
2.0 V		2.8		Gbps	C <sub>B</sub>	VIH = 2.0 V
<b>Dynamic Performance</b>						
Toggle DATx, drive VIL to/from VIH						
Propagation Delay						
VIH = 2.0 V, VIL = 0.0 V, terminated						
Time		750		ps	C <sub>B</sub>	
TC		2		ps/°C	C <sub>T</sub>	
<b>Delay Matching</b>						
VIH = 2.0 V, VIL = 0.0 V, terminated						
Edge to Edge		10		ps	C <sub>B</sub>	t <sub>LH0</sub> - t <sub>HL0</sub> ; t <sub>LH1</sub> - t <sub>HL1</sub>
Channel to Channel		35		ps	C <sub>B</sub>	t <sub>LH0</sub> - t <sub>LH1</sub> ; t <sub>HL0</sub> - t <sub>HL1</sub>
Delay Change vs. Duty Cycle		±7		ps	C <sub>B</sub>	VIH = 2.0 V, VIL = 0.0 V, terminated, 1 MHz, 5% to 95%
Overshoot and Undershoot		50		mV	C <sub>B</sub>	VIH = 2.0 V, VIL = 0.0 V, terminated, minimum driver CLC

## SPECIFICATIONS

Table 2. (Continued)

Parameter	Min	Typ	Max	Unit	Test Level	Test Conditions/Comments
Settling Time (VIH to VIL)						Toggle DATx
To Within 3% of Final Value		1		ns	C <sub>B</sub>	VIH = 2.0 V, VIL = 0.0 V, from 50% crossing, terminated
To Within 1% of Final Value		10		ns	C <sub>B</sub>	VIH = 2.0 V, VIL = 0.0 V, from 50% crossing, terminated
Dynamic Performance						
Drive Active to/from VIT						Toggle RCVx, VIH = 2.0 V, VIT = 1.0 V, VIL = 0.0 V, terminated
Transition Time						20% to 80%
Active to VIT		200		ps	C <sub>B</sub>	
VIT to Active		170		ps	C <sub>B</sub>	
Propagation Delay		1.0		ns	C <sub>B</sub>	
TC		2		ps/°C	C <sub>T</sub>	
Drive Active to/from Inhibit						Toggle RCVx, VIH = 1.0 V, VIL = -1.0 V, terminated
Transition Time						20% to 80%
Inhibit to Active		250		ps	C <sub>B</sub>	
Active to Inhibit		850		ps	C <sub>B</sub>	
Propagation Delay						
Inhibit to VIH		2.1		ns	C <sub>B</sub>	
Inhibit to VIL		2.5		ns	C <sub>B</sub>	
Matching Inhibit to VIL vs. Inhibit to VIH		0.4		ns	C <sub>B</sub>	
VIH to Inhibit		2.5		ns	C <sub>B</sub>	
VIL to Inhibit		2.1		ns	C <sub>B</sub>	
Input/Output Spike		125		mV p-p	C <sub>B</sub>	VIH = 0.0 V, VIL = 0.0 V, terminated, toggle RCVx
Cable Loss Compensation (CLC)						VIH = 2.0 V, VIL = 0.0 V, terminated
Amplitude		20		%	C <sub>B</sub>	Maximum CLC setting
Resolution		3		Bits	D	
Time Constant 1		400		ps	S	Maximum CLC setting
Time Constant 2		1.5		ns	S	Maximum CLC setting

## REFLECTION CLAMP SPECIFICATIONS

Clamp accuracy specifications apply only when VCHx - VCLx > 0.8 V.

Table 3.

Parameter	Min	Typ	Max	Unit	Test Level	Test Conditions/Comments
VCH						
Functional Range	-0.5		+5.0	V	D	
Offset Error	-300		+300	mV	P	Driver high-Z, sinking 1 mA, measured at DAC Code 0x4000 (0.0 V), uncalibrated
Offset TC		±0.25		mV/°C	C <sub>T</sub>	
Gain	1.0		1.1	V/V	P	Driver high-Z, sinking 1 mA, gain derived from measurements at DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V), based on an ideal DAC transfer function (see Table 24)
Gain TC		±25		ppm/°C	C <sub>T</sub>	
Resolution		153		μV	D	
DNL		±250		μV	C <sub>T</sub>	Driver high-Z, sinking 1 mA, after two-point gain/offset calibration; calibration points at DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V), measured over the functional range
INL	-20		+20	mV	P	Driver high-Z, sinking 1 mA, after two-point gain/offset calibration; calibration points at DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V), measured over the functional range

## SPECIFICATIONS

Table 3. (Continued)

Parameter	Min	Typ	Max	Unit	Test Level	Test Conditions/Comments
VCL						
Functional Range	-2.0		+3.5	V	D	
Offset Error	-300		+300	mV	P	Driver high-Z, sourcing 1 mA, measured at DAC Code 0x4000 (0.0 V), uncalibrated
Offset TC		±0.25		mV/°C	C <sub>T</sub>	
Gain	1.0		1.1	V/V	P	Drive high-Z, sourcing 1 mA, gain derived from measurements at DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V), based on an ideal DAC transfer function (see Table 24)
Gain TC		±25		ppm/°C	C <sub>T</sub>	
Resolution		153		μV	D	
DNL		±250		μV	C <sub>T</sub>	Drive high-Z, sourcing 1 mA, after two-point gain/offset calibration; calibration points at DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V), measured over the functional range
INL	-20		+20	mV	P	Drive high-Z, sourcing 1 mA, after two-point gain/offset calibration; calibration points at DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V), measured over the functional range
DC CLAMP CURRENT LIMIT						
VCHx	-105		-60	mA	P	VCHx = -1.0 V, VCLx = -2.0 V, V <sub>DUTx</sub> = 4.5 V
VCLx	+60		+105	mA	P	VCHx = 5.0 V, VCLx = 4.0 V, V <sub>DUTx</sub> = -1.5 V
DUTGND VOLTAGE ACCURACY	-10	±2	+10	mV	P	Over ±0.1 V range, measured at end points of VCHx and VCLx functional range

## NORMAL WINDOW COMPARATOR (NWC) SPECIFICATIONS

Table 4.

Parameter	Min	Typ	Max	Unit	Test Level	Test Conditions/Comments
DC SPECIFICATIONS						
Input Voltage Range	-1.5		+4.5	V	D	
Differential Voltage Range	±0.1		±6.0	V	D	
Input Offset Voltage	-250		+250	mV	P	Measured at DAC Code 0x4000 (0.0 V); uncalibrated
Input Offset Voltage TC		±150		μV/°C	C <sub>T</sub>	
Gain	1.0		1.1	V/V	P	Gain derived from measurements at DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V); based on an ideal DAC transfer function (see Table 24)
Gain TC		±10		ppm/°C	C <sub>T</sub>	
Threshold Resolution		153		μV	D	
Threshold DNL		±0.25		mV	C <sub>T</sub>	Measured over -1.5 V to +4.5 V functional range after two-point gain/offset calibration; calibration points at DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V)
Threshold INL						After two-point gain/offset calibration; calibration points at DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V)
Focused Range	-5		+5	mV	P	Measured over -0.5 V to +3.5 V range
Full Range	-7		+7	mV	P	Measured over -1.5 V to +4.5 V range
DUTGND Voltage Accuracy	-5	±1	+5	mV	P	Over ±0.1 V range; measured over -0.5 V to +3.5 V focused NWC input range
Uncertainty Band		10		mV	C <sub>B</sub>	V <sub>DUTx</sub> = 0.0 V, sweep comparator threshold to determine the uncertainty band
Programmable Hysteresis		100		mV	C <sub>B</sub>	
Hysteresis Resolution		4		Bits	D	
DC PSR		±5		mV/V	C <sub>T</sub>	Measured at DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V) calibration points



## SPECIFICATIONS

Table 4. (Continued)

Parameter	Min	Typ	Max	Unit	Test Level	Test Conditions/Comments
Digital Output Characteristics						
Internal Pull-up Resistance to Comparator, VTTCx	46	50	54	$\Omega$	P	Source 1 mA and 10 mA from the output pin in high state, measure $\Delta V$ to calculate resistance; $R = \Delta V/9$ mA; repeat for all output pins
Common-Mode Voltage						Measured relative to $V_{TTCx}$ , with 100 $\Omega$ differential termination
ADATE320		-250		mV	C <sub>T</sub>	
ADATE320-1		-400		mV	C <sub>T</sub>	
Differential Mode Voltage						Measured differentially
100 $\Omega$ Differential Termination						
ADATE320		250		mV	C <sub>T</sub>	
ADATE320-1		400		mV	C <sub>T</sub>	
No External Termination						
ADATE320	450	500	550	mV	P	
ADATE320-1	700	800	900	mV	P	
AC SPECIFICATIONS						
Rise/Fall Times, 20% to 80%		100		ps	C <sub>B</sub>	Unless otherwise specified, all ac tests are performed after dc levels calibration; input transition time: 50 ps 20% to 80%; outputs terminated 50 $\Omega$ to 0.0 V; comparator CLC set to 1/4 scale (010) Measured with 50 $\Omega$ to 0.0 V
Propagation Delay		580		ps	C <sub>B</sub>	$V_{DUTx} = 0.0$ V to 1.0 V swing, drive term mode, VIT = 0.0 V, comparator threshold = 0.5 V
Propagation Delay TC		1		ps/ $^{\circ}$ C	C <sub>T</sub>	$V_{DUTx} = 0.0$ V to 1.0 V swing, drive term mode, VIT = 0.0 V, comparator threshold = 0.5 V
Propagation Delay Matching High Transition to Low Transition		10		ps	C <sub>B</sub>	$V_{DUTx} = 0.0$ V to 1.0 V swing, drive term mode, VIT = 0.0 V, comparator threshold = 0.5 V
Propagation Delay Matching High to Low Comparator		10		ps	C <sub>B</sub>	$V_{DUTx} = 0.0$ V to 1.0 V swing, drive term mode, VIT = 0.0 V, comparator threshold = 0.5 V
Propagation Delay Dispersion						Drive term mode, VIT = 0.0 V
Slew Rate: 400 ps vs. 1.0 ns (20% to 80%)		20		ps	C <sub>B</sub>	$V_{DUTx} = 0.0$ V to 0.5 V swing, comparator threshold = 0.25 V
Overdrive: 250 mV vs. 1.0 V		25		ps	C <sub>B</sub>	For 250 mV: $V_{DUTx}$ : 0.0 V to 0.50 V swing; for 1.0 V: $V_{DUTx}$ : 0.0 V to 1.25 V swing, comparator threshold = 0.25 V
1.0 V Pulse Width: 0.4 ns, 0.5 ns, 1 ns, 5 ns, 10 ns		25		ps	C <sub>B</sub>	$V_{DUTx} = 0.0$ V to 1.0 V swing, 32 MHz, comparator threshold = 0.5 V
0.5 V Pulse Width: 0.4 ns, 0.5 ns, 1 ns, 5 ns, 10 ns		25		ps	C <sub>B</sub>	$V_{DUTx} = 0.0$ V to 0.5 V swing, 32 MHz, comparator threshold = 0.25 V
Duty Cycle: 5% to 95%		10		ps	C <sub>B</sub>	$V_{DUTx} = 0.0$ V to 1.0 V swing, 32 MHz, comparator threshold = 0.5 V
Minimum Detectable Pulse Width		200		ps	C <sub>B</sub>	$V_{DUTx} = 0.0$ V to 1.0 V swing, 32 MHz, greater than 50% output differential amplitude
Input Equivalent Rise/Fall Time, 1.0 V, Terminated		110		ps	C <sub>B</sub>	$V_{DUTx} = 0.0$ V to 1.0 V swing, drive term mode, VIT = 0.0 V, CLC = 010, measured from digitized plot, 20% to 80% transition time of digitized plot is root-sum square (RSS) of input equivalent rise/fall and 50 ps input stimulus
Input Equivalent Rise/Fall Time, 2.0 V, Unterminated		500		ps	C <sub>B</sub>	$V_{DUTx} = 0.0$ V to 2.0 V swing, drive high-Z, measured from digitized plot, 20% to 80% transition time of digitized plot is root-sum square (RSS) of input equivalent rise/fall and 50 ps input stimulus
Cable Loss Compensation (CLC)						$V_{DUTx} = 0.0$ V to 1.0 V swing, drive term mode, VIT = 0.0 V, maximum CLC setting
CLC Amplitude		20		%	C <sub>B</sub>	
CLC Resolution		3		Bits	D	
CLC Time Constant 1		280		ps	S	
CLC Time Constant 2		4.8		ns	S	

## SPECIFICATIONS

## DIFFERENTIAL MODE COMPARATOR (DMC) SPECIFICATIONS

Table 5.

Parameter	Min	Typ	Max	Unit	Test Level	Test Conditions/Comments
<b>DC SPECIFICATIONS</b>						
Input Voltage Range	-1.5		+4.5	V	D	VOHx tests at VOLx = -1.5 V, VOLx tests at VOHx = 1.5 V
Functional Differential Range	±0.05		±1.1	V	D	
Maximum Differential Input			±6.0	V	D	
Input Offset Voltage	-250		+250	mV	P	Offset interpolated from measurements at DAC Code 0x2666 (-1.0 V) and DAC Code 0x5999 (1.0 V), with V <sub>CM</sub> = 0.0 V
Input Offset Voltage TC		±150		µV/°C	C <sub>T</sub>	
Gain	1.0		1.1	V/V	P	Gain derived from measurements at DAC Code 0x2666 (-1.0 V) and DAC Code 0x5999 (1.0 V); based on an ideal DAC transfer function (see Table 24)
Gain TC		±40		ppm/°C	C <sub>T</sub>	
VOHx, VOLx Resolution		153		µV	D	
VOHx, VOLx DNL		±250		µV	C <sub>T</sub>	After two-point gain/offset calibration; V <sub>CM</sub> = 0.0 V; calibration points at DAC Code 0x2666 (-1.0 V) and DAC Code 0x5999 (1.0 V)
VOHx, VOLx INL	-8		+8	mV	P	After two-point gain/offset calibration; V <sub>CM</sub> = 0.0 V; calibration points DAC Code 0x2666 (-1.0 V) and DAC Code 0x5999 (1.0 V), measured over VOHx/VOLx range of -1.1 V to +1.1 V
Uncertainty Band		11		mV	C <sub>B</sub>	V <sub>DUTx</sub> = 0.0 V, sweep comparator threshold to determine the uncertainty band
Programmable Hysteresis		200		mV	C <sub>B</sub>	
Hysteresis Resolution		4		Bits	D	
Common-Mode Rejection Ratio (CMRR)	-1.0		+1.0	mV/V	P	ΔOffset measured at V <sub>CM</sub> = -1.5 V and +4.5 V, V <sub>DM</sub> = 0.0 V
DC PSR		±5		mV/V	C <sub>T</sub>	ΔOffset measured at V <sub>CM</sub> = 0.0 V, V <sub>DM</sub> = calibration points DAC Code 0x2666 (-1.0 V) and DAC Code 0x5999 (1.0 V)
<b>AC SPECIFICATIONS</b>						
Propagation Delay		580		ps	C <sub>B</sub>	All ac tests are performed after dc levels calibration; input transition time = 50 ps 20% to 80%; outputs terminated 50 Ω to VTTCx, comparator CLC set to ¼ scale (010) V <sub>DUT0</sub> = 0.0 V, V <sub>DUT1</sub> = -0.5 V to +0.5 V swing, drive termination mode, VIT = 0.0 V, comparator threshold = 0.0 V, repeat with VDUTx inputs reversed
Propagation Delay TC		2		ps/°C	C <sub>T</sub>	V <sub>DUT0</sub> = 0.0 V, V <sub>DUT1</sub> = -0.5 V to +0.5 V swing, drive termination mode, VIT = 0.0 V, comparator threshold = 0.0 V, repeat with VDUTx inputs reversed
Propagation Delay Matching High Transition to Low Transition		15		ps	C <sub>B</sub>	V <sub>DUT0</sub> = 0.0 V, V <sub>DUT1</sub> = -0.5 V to +0.5 V swing, drive termination mode, VIT = 0.0 V, comparator threshold = 0.0 V, repeat with VDUTx inputs reversed
Propagation Delay Matching High to Low Comparator		15		ps	C <sub>B</sub>	V <sub>DUT0</sub> = 0.0 V, V <sub>DUT1</sub> = -0.5 V to +0.5 V swing, drive termination mode, VIT = 0.0 V, comparator threshold = 0.0 V, repeat with VDUTx inputs reversed
Propagation Delay Dispersion						V <sub>DUT0</sub> = 0.0 V, VIT = 0.0 V, drive termination mode, repeat with VDUTx inputs reversed
Slew Rate: 400 ps vs. 1 ns (20% to 80%)		30		ps	C <sub>B</sub>	V <sub>DUT1</sub> = -0.5 V to +0.5 V swing, comparator threshold = 0.0 V
Overdrive: 250 mV vs. 750 mV		25		ps	C <sub>B</sub>	For 250 mV: V <sub>DUT1</sub> = 0.0 V to 0.5 V swing; for 750 mV: V <sub>DUT1</sub> = 0.0 V to 1.0 V swing, comparator threshold = -0.25 V, repeat with VDUTx inputs reversed with comparator threshold = +0.25 V
1.0 V Pulse Width: 0.7 ns, 1.0 ns, 5.0 ns, 10 ns		25		ps	C <sub>B</sub>	V <sub>DUT1</sub> = -0.5 V to +0.5 V swing, 32 MHz, comparator threshold = 0.0 V

## SPECIFICATIONS

Table 5. (Continued)

Parameter	Min	Typ	Max	Unit	Test Level	Test Conditions/Comments
0.5 V Pulse Width: 0.6 ns, 1.0 ns, 5.0 ns, 10 ns		25		ps	C <sub>B</sub>	V <sub>DUT1</sub> = -0.25 V to +0.25 V swing, 32 MHz, comparator threshold = 0.0 V
Duty Cycle: 5% to 95%		5		ps	C <sub>B</sub>	V <sub>DUT1</sub> = -0.5 V to +0.5 V swing, 32 MHz, comparator threshold = 0.0 V
Minimum Detectable Pulse Width		200		ps	C <sub>B</sub>	V <sub>DUT0</sub> = 0.0 V, V <sub>DUT1</sub> = -0.5 V to +0.5 V swing, 32 MHz, drive term mode, VIT = 0.0 V, comparator threshold = 0.0 V, greater than 50% output differential amplitude, repeat with VDUTx inputs reversed
Input Equivalent Rise/Fall Time		110		ps	C <sub>B</sub>	V <sub>DUT0</sub> = 0.0 V, V <sub>DUT1</sub> = -0.5 V to +0.5 V swing, drive term mode, VIT = 0.0 V, comparator threshold = 0.0 V, CLC = ¼ scale, measured from digitized plot, $t = \sqrt{(t_{CMP}^2 - t_N^2)}$
Cable Loss Compensation (CLC)						V <sub>DUT0</sub> = 0.0 V, V <sub>DUT1</sub> = -0.8 V to +0.8 V swing, drive term mode, VIT = 0.0 V, comparator threshold = 0.0 V, comparator CLC set to maximum CLC setting, repeat with VDUTx inputs reversed
CLC Amplitude		20		%	C <sub>B</sub>	
CLC Resolution		3		Bits	D	
CLC Time Constant 1		280		ps	S	
CLC Time Constant 2		4.8		ns	S	

## ACTIVE LOAD SPECIFICATIONS

Table 6.

Parameter	Min	Typ	Max	Unit	Test Level	Test Conditions/Comments
DC SPECIFICATIONS						Load in active on state, RCVx active
Input Characteristics						
Active Load Commutation Voltage (VCOMx) Range	-1.5		+4.5	V	D	IOHx = IOLx = 1 mA, VDUTx open circuit
VCOMx Offset	-200		+200	mV	P	Measured at DAC Code 0x4000 (0.0 V), uncalibrated
VCOMx Offset TC		±100		µV/°C	C <sub>T</sub>	
VCOMx Gain	1.0		1.1	V/V	P	Gain derived from measurements at DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V)
VCOMx Gain TC		±20		ppm/°C	C <sub>T</sub>	
VCOMx Resolution		153		µV	D	
VCOMx DNL		±250		µV	C <sub>T</sub>	IOHx = IOLx = 12.5 mA, after two-point gain/offset calibration; measured over VCOMx range -1.5 V to +4.5 V; calibration points DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V)
VCOMx INL						IOHx = IOLx = 12.5 mA; after two-point gain/offset calibration; calibration points DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V)
Focused Range	-5		+5	mV	P	Measured over VCOMx range of -0.5 V to +3.5 V
Full Range	-10		+10	mV	P	Measured over VCOMx range of -1.5 V to +4.5 V
DUTGND Voltage Accuracy	-5	±1	+5	mV	P	Over ±0.1 V range; measured over -0.5 V to +3.5 V focused VCOMx range
Output Characteristics						
Maximum Source Current (IOLx)	25			mA	D	V <sub>DUTx</sub> ≤ 3.5 V (a compliance limit is set by a 50 Ω internal resistor as illustrated in Figure 141)
IOLx Offset	-600		+600	µA	P	IOHx = -2.5 mA, VCOMx = 1.5 V, V <sub>DUTx</sub> = 0.0 V; offset extrapolated from measurements at DAC Code 0x451F (1 mA) and DAC Code 0xA666 (20 mA)
IOLx Offset TC		±1		µA/°C	C <sub>T</sub>	
IOLx Gain Error	0		+25	%	P	IOHx = -2.5 mA, VCOMx = 1.5 V, V <sub>DUTx</sub> = 0.0 V; gain derived from measurements at DAC Code 0x451F (1 mA) and DAC Code 0xA666 (20 mA); based on an ideal dc transfer function

## SPECIFICATIONS

Table 6. (Continued)

Parameter	Min	Typ	Max	Unit	Test Level	Test Conditions/Comments
IOLx Gain TC		±100		ppm/°C	C <sub>T</sub>	
IOLx Resolution		763		nA	D	
IOLx DNL		±1.25		μA	C <sub>T</sub>	IOLx = -2.5 mA, VCOMx = 1.5 V, V <sub>DUTx</sub> = 0.0 V; after two-point gain/offset calibration; measured over IOLx range 0 mA to 25 mA; calibrated at DAC Code 0x451F (1 mA) and DAC Code 0xA666 (20 mA)
IOLx INL	-100		+100	μA	P	IOLx = -2.5 mA, VCOM = 1.5 V, V <sub>DUTx</sub> = 0.0 V, after two-point gain/offset calibration
IOLx 90% Commutation Voltage		0.25	0.4	V	P	IOLx = IOLx = 25 mA, VCOM = 2.0 V, measure IOLx reference at V <sub>DUTx</sub> = -1.0 V, measure IOLx current at V <sub>DUTx</sub> = 1.6 V, check > 90% of reference current
		0.1		V	C <sub>T</sub>	IOLx = IOLx = 1 mA, VCOM = 2.0 V, measure IOLx reference at V <sub>DUTx</sub> = -1.0 V, measure IOLx current at V <sub>DUTx</sub> = 1.9 V, check > 90% of reference current
Maximum Sink Current (IOHx)	25			mA	D	V <sub>DUTx</sub> ≥ -0.5 V (a compliance limit is set by a 50 Ω internal resistor as illustrated in <a href="#">Figure 141</a> )
IOHx Offset	-600		+600	μA	P	IOLx = -2.5 mA, VCOM = 1.5 V, V <sub>DUTx</sub> = 3.0 V, offset extrapolated from measurements at DAC Code 0x451F (1 mA) and DAC Code 0xA666 (20 mA)
IOHx Offset TC		±1		μA/°C	C <sub>T</sub>	
IOHx Gain Error	0		+25	%	P	IOLx = -2.5 mA, VCOM = 1.5 V, V <sub>DUTx</sub> = 3.0 V, gain derived from measurements at DAC Code 0x451F (1 mA) and DAC Code 0xA666 (20 mA); based on an ideal DAC transfer function
IOHx Gain TC		±100		ppm/°C	C <sub>T</sub>	
IOHx Resolution		763		nA	D	
IOHx DNL		±1.25		μA	C <sub>T</sub>	IOLx = -2.5 mA, VCOM = 1.5 V, V <sub>DUTx</sub> = 3.0 V, after two-point gain/offset calibration; measured over IOHx range of 0 mA to 25 mA; calibrated at DAC Code 0x451F (1 mA) and DAC Code 0xA666 (20 mA)
IOHx INL	-100		+100	μA	P	IOLx = -2.5 mA, VCOM = 1.5 V, V <sub>DUTx</sub> = 3.0 V, after two-point gain/offset calibration
IOHx 90% Commutation Voltage		0.25	0.4	V	P	IOHx = IOLx = 25 mA, VCOM = 2.0 V, measure IOHx reference at V <sub>DUTx</sub> = 4.0 V, measure IOHx current at V <sub>DUTx</sub> = 2.4 V, ensure > 90% of reference current
		0.1		V	C <sub>T</sub>	IOHx = IOLx = 1 mA, VCOM = 2.0 V, measure IOHx reference at V <sub>DUTx</sub> = 4.0 V, measure IOHx current at V <sub>DUTx</sub> = 2.1 V <sub>DUTx</sub> , ensure > 90% of reference current
AC SPECIFICATIONS						
All ac measurements are performed after dc calibration unless noted, load active on						
Dynamic Performance						
Toggle RCVx; DUTx terminated 50 Ω to 0.0 V; IOLx = IOHx = 20 mA, VIH = VIL = 0.0 V; VCOM = +1.5 V for IOLx and -1.5 V for IOHx						
Propagation Delay, Load Active On to Load Active Off		1.7		ns	C <sub>B</sub>	Measured from zero crossing of RCVx - $\overline{RCVx}$ to 50% of final output value; repeat for drive low and drive high
Propagation Delay, Load Active Off to Load Active On		2.9		ns	C <sub>B</sub>	Measured from zero crossing of RCVx - $\overline{RCVx}$ to 50% of final output value; repeat for drive low and drive high
Propagation Delay Matching		1.2		ns	C <sub>B</sub>	Active on vs. active off; repeat for drive low and drive high
Load Spike		140		mV	C <sub>B</sub>	Repeat for drive low and drive high
Settling Time to Within 5%		2.5		ns	C <sub>B</sub>	Measured from output crossing 50% final value to output within 5% final value

## PPMU SPECIFICATIONS

PPMU enabled in force voltage mode unless noted.

## SPECIFICATIONS

Table 7.

Parameter	Min	Typ	Max	Unit	Test Level	Test Conditions/Comments
<b>FORCE VOLTAGE (FV)</b>						
Current Range A	-40		+40	mA	D	
Current Range B	-1		+1	mA	D	
Current Range C	-100		+100	μA	D	
Current Range D	-10		+10	μA	D	
Current Range E	-2		+2	μA	D	
FV Range at Output, Range A	-1.0		+4.0	V	D	Output range for full-scale source/sink
	-1.5		+4.5	V	D	Output range for ±25 mA or less
FV Range at Output, Range B, Range C, Range D, and Range E	-1.5		+4.5	V	D	Output range for full-scale source/sink
FV Offset, Range C	-100		+100	mV	P	Measured at DAC Code 0x4000 (0.0 V) in Range C
FV Offset, All Ranges		±30		mV	C <sub>T</sub>	Measured at DAC Code 0x4000 (0.0 V) applies to all other ranges
FV Offset TC, All Ranges		±100		μV/°C	C <sub>T</sub>	Measured at DAC Code 0x4000 (0.0 V)
FV Gain, Range C	1.0		1.1	V/V	P	Gain derived from measurements at DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V); based on an ideal DAC transfer function
FV Gain, All Ranges		1.05		V/V	C <sub>T</sub>	Gain derived from measurements at DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V); based on an ideal DAC transfer function
FV Gain TC, All Ranges		±10		ppm/°C	C <sub>T</sub>	Gain derived from measurements at DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V)
<b>FV INL</b>						
Range A		±1.5		mV	C <sub>T</sub>	After two-point gain/offset calibration, output range of -1.5 V to +4.5 V; calibration points DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V), PPMU Current Range A
Range C, Focused Range	-1.7		+1.7	mV	P	After two-point gain/offset calibration, output range of -0.5 V to +3.5 V; calibration points DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V)
Range C, Full Range	-5		+5	mV	P	After two-point gain/offset calibration, output range of -1.5 V to +4.5 V; calibration points DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V)
Range B, Range D, and Range E		±1.0		mV	C <sub>T</sub>	After two-point gain/offset calibration, output range of -1.5 V to +4.5 V; calibration points DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V)
FV Compliance vs. Source/Sink Current, Range A (±40 mA)		±1		mV	C <sub>T</sub>	Force -1.0 V; measure voltage while sinking 0.0 mA and full-scale current; measure ΔV; force 4.0 V; measure voltage while sourcing 0.0 mA and full-scale current; measure ΔV
FV Compliance vs. Source/Sink Current, Range A (±25 mA)		±1		mV	C <sub>T</sub>	Force -1.5 V; measure voltage while sinking 0.0 mA and 25 mA; measure ΔV; force 4.5 V; measure voltage while sourcing 0.0 mA and 25 mA; measure ΔV
FV Compliance vs. Source/Sink Current, Range B, Range C, Range D, and Range E		±1		mV	C <sub>T</sub>	Force -1.5 V; measure voltage while sinking 0.0 mA and full-scale current; measure ΔV; force 4.5 V; measure voltage while sourcing 0.0 mA and full-scale current; measure ΔV
DUTGND Voltage Accuracy	-5	±1	+5	mV	P	Over ±0.1 V range; measured over -0.5 V to +3.5 V focused PPMU output range
<b>FORCE CURRENT (FI)</b>						
DUTx Pin Voltage Range, Range A	-1.0		+4.0	V	D	PPMU enabled in force current/measure current (FIMI) Full-scale source and sink current
	-1.5		+4.5	V	D	DUTx pin source and sink 25 mA or less

## SPECIFICATIONS

Table 7. (Continued)

Parameter	Min	Typ	Max	Unit	Test Level	Test Conditions/Comments
DUTx Pin Voltage Range, Range B, Range C, Range D, and Range E	-1.5		+4.5	V	D	Full-scale source and sink current
Zero-Current Offset, All Ranges	-14.5		+14.5	% FSR	P	Interpolated from measurements at PPMU DAC Code 0x4CCC (-80% FS) and DAC Code 0xB333 (80% FS) for each range
Zero-Current Offset TC		±0.02		% FSR/°C	C <sub>T</sub>	
Gain Error, All Ranges	0		30	%	P	Derived from measurements at PPMU DAC Code 0x4CCC (-80% FS) and DAC Code 0xB333 (80% FS) for each range
Gain Drift						
Range A		±50		ppm/°C	C <sub>T</sub>	PPMU self heating effects in Range A can influence gain drift measurements
Range B		±50		ppm/°C	C <sub>T</sub>	
Range C, Range D, and Range E		±50		ppm/°C	C <sub>T</sub>	
INL						After two-point gain/offset calibration
Range A	-0.12		+0.12	% FSR	P	Measured over FSR output of Range A (±40 mA)
Range B, Range C, and Range D	-0.04		+0.04	% FSR	P	Measured over FSR output of Range B (±1 mA), Range C (±100 µA), and Range D (±10 µA)
Range E	-0.045		+0.045	% FSR	P	Measured over FSR output of Range E (±2 µA)
FI Compliance vs. Voltage Load						
Range A	-0.3		+0.3	% FSR	P	Force positive full-scale current driving -1.0 V and +4.0 V, measure ΔI at DUTx pin; force negative full-scale current driving -1.0 V and +4.0 V, measure ΔI at DUTx pin
	-0.1		+0.1	% FSR	P	Force positive full-scale current driving 0.0 V and 3.0 V, measure ΔI at DUTx pin; force negative full-scale current driving 0.0 V and 3.0 V, measure ΔI at DUTx pin
Range B and Range C	-0.3		+0.3	% FSR	P	Force positive full-scale current driving -1.5 V and +4.5 V, measure ΔI at DUTx pin; force negative full-scale current driving -1.5 V and +4.5 V, measure ΔI at DUTx pin
	-0.06		+0.06	% FSR	P	Force positive full-scale current driving 0.0 V and 3.0 V, measure ΔI at DUTx pin; force negative full-scale current driving 0.0 V and 3.0 V, measure ΔI at DUTx pin
Range D	-0.3		+0.3	% FSR	P	Force positive full-scale current driving -1.5 V and +4.5 V, measure ΔI at DUTx pin; force negative full-scale current driving -1.5 V and +4.5 V, measure ΔI at DUTx pin
Range E	-0.85		+0.85	% FSR	P	Force positive full-scale current driving -1.5 V and +4.5 V, measure ΔI at DUTx pin; force negative full-scale current driving -1.5 V and +4.5 V, measure ΔI at DUTx pin; allows 10 nA DUTx pin leakage
MEASURE VOLTAGE (MV)						PPMU enabled, force voltage/measure voltage (FVMV)
Range	-1.5		+4.5	V	D	
Offset	-25		+25	mV	P	Range B, V <sub>DUTx</sub> = 0.0 V, offset = (PPMU_Mx - V <sub>DUTx</sub> )
Offset TC		±50		µV/°C	C <sub>T</sub>	
Gain	0.98		1.02	V/V	P	Range B, derived from measurements at V <sub>DUTx</sub> = 0.0 V and 3.0 V
Gain TC		±5		ppm/°C	C <sub>T</sub>	
INL	-1.7		+1.7	mV	P	Range B, measured over -1.5 V to +4.5 V
MEASURE CURRENT (MI)						PPMU enabled in FIMI
DUTx Pin Voltage Range						Full-scale source and sink current

## SPECIFICATIONS

Table 7. (Continued)

Parameter	Min	Typ	Max	Unit	Test Level	Test Conditions/Comments
Range A	-1.0		+4.0	V	D	
Range B, Range C, Range D, and Range E	-1.5		+4.5	V	D	
Zero-Current Offset						
Range B	-4		+4	%FSR	P	Interpolated from measurements sourcing and sinking 80% FS current each range; for example, 2% FSR is 40 $\mu$ A in Range B
All Ranges		$\pm 0.5$		%FSR	C <sub>T</sub>	
Zero-Current Offset TC						
Range A		$\pm 0.01$		%FSR/ $^{\circ}$ C	C <sub>T</sub>	
Range B, Range C, and Range D		$\pm 0.01$		%FSR/ $^{\circ}$ C	C <sub>T</sub>	
Range E		$\pm 0.02$		%FSR/ $^{\circ}$ C	C <sub>T</sub>	
Gain Error						Derived from measurements sourcing and sinking 80% FS current
Range B	-30		+5	%	P	
All Ranges		-10		%	C <sub>T</sub>	
Gain TC						
Range A		$\pm 50$		ppm/ $^{\circ}$ C	C <sub>T</sub>	
Range B, Range C, and Range D		$\pm 50$		ppm/ $^{\circ}$ C	C <sub>T</sub>	
Range E		$\pm 50$		ppm/ $^{\circ}$ C	C <sub>T</sub>	
INL						After two-point gain/offset calibration at $\pm 80\%$ FS current
Range A		$\pm 0.02$		%FSR	C <sub>T</sub>	Measured over FSR output of -40 mA to +40 mA
Range B	-0.02		+0.02	%FSR	P	Measured over FSR output of -1 mA to +1 mA
Range C, Range D, and Range E		$\pm 0.01$		%FSR	C <sub>T</sub>	Measured over FSR output of Range C, Range D, and Range E
DUTx Pin Voltage Rejection	-1.3		+1.3	$\mu$ A	P	Range B, FVM1, force -1.0 V and +4.0 V into 0.5 mA load, measure $\Delta I$ reported at PPMU_Mx pin
DUTGND Voltage Accuracy	-5	$\pm 1$	+5	mV	P	Over $\pm 0.1$ V range
MEASURE PIN DC CHARACTERISTICS						
Output Range	-1.5		+5.0	V	D	
Output Impedance			200	$\Omega$	P	PPMU enabled in FVMV, source resistance: PPMU force 4.5 V into 0.0 mA, -1.0 mA, sink resistance: PPMU force -1.5 V into 0.0 mA, 1.0 mA, resistance = $\Delta V/\Delta I$ at PPMU_Mx pin
Output Leakage Current When Tristated	-1		+1	$\mu$ A	P	Tested at -1.7 V and +5.2 V
Output Short Circuit Current	-10		+10	mA	P	PPMU enabled in FVMV, source: PPMU force +4.5 V, PPMU_Mx = -1.5 V, sink: PPMU force -1.5 V, PPMU_Mx = 5.0 V
PPMU_Mx Pin, Parasitic Output Capacitance			2	pF	S	Parasitic capacitance contributed by pin
PPMU_Mx Pin, External Load Capacitance	100			pF	S	External capacitance tolerated by pin (exceeding this value may cause instability)
PPMU VOLTAGE CLAMPS (FI)						PPMU enabled in FIMI, PPMU clamps enabled; clamp accuracy applies only when $ PCHx - PCLx  \geq 1.0$ V
Low Voltage Clamp Range (PCLx)	-1.5		+3.5	V	D	
High Voltage Clamp Range (PCHx)	-0.5		+4.5	V	D	
Offset, Voltage Clamps (PCHx/PCLx)	-300		+300	mV	P	Range B, PPMU force $\pm 0.5$ mA into open; PCHx measured at DAC Code 0x4000 (0.0 V) with PCLx at DAC Code 0x0000 (-2.5 V); PCLx measured at DAC

## SPECIFICATIONS

Table 7. (Continued)

Parameter	Min	Typ	Max	Unit	Test Level	Test Conditions/Comments
Offset TC, Voltage Clamps (PCHx/PCLx)		±0.5		mV/°C	C <sub>T</sub>	Code 0x4000 (0.0 V) with PCHx at DAC Code 0xFFFF (+7.5 V)
Gain, Voltage Clamps (PCHx/PCLx)	1.0		1.1	V/V	P	Range B, PPMU force ±0.5 mA into open; PCHx gain derived from measurements at DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V) with PCLx at DAC Code 0x0000 (-2.5 V); PCLx gain derived from measurements at DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V) with PCHx at DAC Code 0xFFFF (7.5 V)
Gain TC, Voltage Clamps (PCHx/PCLx)		±25		ppm/°C	C <sub>T</sub>	
INL, Voltage Clamps (PCHx/PCLx)	-20		+20	mV	P	Range B, PPMU force ±0.5 mA into open after two-point gain/offset calibration; measured over PPMU clamp functional range
Positive Voltage Clamp, Voltage Droop (Source)	-50		+50	mV	P	ΔV at DUTx pin, Range A, PCHx = +4.0 V, PCLx = -1.0 V, PPMU force 5.0 mA and 40 mA into open circuit, calibrated
Negative Voltage Clamp, Voltage Droop (Sink)	-50		+50	mV	P	ΔV at DUTx pin, Range A, PCHx = +4.0 V, PCLx = -1.0 V, PPMU force -5.0 mA and -40 mA into open circuit, calibrated
DUTGND Voltage Accuracy	-5	±1	+5	mV	P	Over ±0.1 V range; measured at end points of clamp functional range
PPMU CURRENT CLAMPS (FV)						PPMU enabled in FVMV, dc accuracy of the current clamps only applies over the following conditions: 30% FS ≤ PCHx ≤ 100% FS or -100% FS ≤ PCLx ≤ -30% FS
Functional Range						
Low Current Clamp (PCLx)	-120		-20	%FS	S	For example, -120% FS in Range A is -48 mA and -20% FS in Range A is -8 mA
High Current Clamp (PCHx)	20		120	%FS	S	For example, 20% FS in Range A is 8 mA and 120% FS in Range A is 48 mA
DC Accuracy Range						
Low Current Clamp (PCLx)	-100		-30	%FS	D	For example, -100% FS in Range A is -40 mA and -30% FS in Range A is -12 mA
High Current Clamp (PCHx)	30		100	%FS	D	For example, 30% FS in Range A is 12 mA and 100% FS in Range A is 40 mA
Static Current Limit, Source and Sink, All Ranges	±120	±140	±160	%FS	P	PCLx at DAC Code 0x0000 (-2.5 V), PCHx at DAC Code 0xFFFF (7.5 V), sink: force -1.5 V, short DUTx to 4.5 V, source: force 4.5 V, short DUTx to -1.5 V
Offset, Current Clamps (PCHx/PCLx)	-10		+10	%FSR	P	All ranges; PPMU force ±1.0 V into 0.0 V <sup>1</sup>
Offset TC, Current Clamps (PCHx/PCLx)		±0.02		%FSR/°C	C <sub>T</sub>	All ranges
Gain Error, Current Clamps (PCHx/PCLx)	0		30	%	P	All ranges; PPMU force ±1.0 V into 0.0 V <sup>2</sup>
Gain TC, Current Clamps (PCHx/PCLx)		±50		ppm/°C	C <sub>T</sub>	All ranges
INL, Current Clamps (PCHx/PCLx)	-0.15		+0.15	%FSR	P	All ranges; PPMU force ±1.0 V into 0.0 V, after two-point gain/offset calibration; PCHx calibration at DAC Code 0xA000 (3.75 V or 50% FS) and DAC Code 0xB333 (4.50 V or 80% FS); PCLx calibration at DAC Code 0x6000 (1.25 V or -50% FS) and DAC Code 0x4CCC (0.50 V or -80% FS); measured over dc accuracy range



## SPECIFICATIONS

Table 7. (Continued)

Parameter	Min	Typ	Max	Unit	Test Level	Test Conditions/Comments
Current Droop						
Low Current Clamp (PCLx), Sink	-2		+2	%FSR	P	PCLx = 0.5 V (-80% FS), PCHx = 4.5 V (80% FS), PPMU force -0.5 V and +3.5 V into $V_{DUTx} = 4.5$ V, measure $\Delta I$ at the DUTx pin in Range A
High Current Clamp (PCHx), Source	-2		+2	%FSR	P	PCLx = 0.5 V (-80% FS), PCHx = 4.5 V (80% FS), PPMU force -0.5 V and +3.5 V into $V_{DUTx} = -1.5$ V, measure $\Delta I$ at the DUTx pin in Range A
<b>SETTLING/SWITCHING TIMES</b>						
FV Settling Time to 0.1% of Final Value						
Range A, 200 pF and 2000 pF Load		20		$\mu$ s	S	PPMU enabled in FV, Range A, step from 0.0 V to 4.0 V
Range B, 200 pF and 2000 pF Load		25		$\mu$ s	S	PPMU enabled in FV, Range B, DCL disabled, step from 0.0 V to 4.0 V
Range C, 200 pF Load		25		$\mu$ s	S	PPMU enabled in FV, Range C, DCL disabled, step from 0.0 V to 4.0 V
Range C, 2000 pF Load		65		$\mu$ s	S	PPMU enabled in FV, Range C, DCL disabled, step from 0.0 V to 4.0 V
FV Settling Time to 1.0% of Final Value						
Range A, 200 pF and 2000 pF Load		16		$\mu$ s	C <sub>B</sub>	PPMU enabled in FV, Range A, DCL disabled, step from 0.0 V to 4.0 V
Range B, 200 pF and 2000 pF Load		14		$\mu$ s	C <sub>B</sub>	PPMU enabled in FV, Range B, DCL disabled, step from 0.0 V to 4.0 V
Range C, 200 pF and 2000 pF Load		18		$\mu$ s	C <sub>B</sub>	PPMU enabled in FV, Range C, DCL disabled enabled, step from 0.0 V to 4.0 V
FI Settling Time to 0.1% of Final Value						
Range A, 200 pF in Parallel with 120 $\Omega$		16		$\mu$ s	S	PPMU enabled in FI, Range A, DCL disabled, step from 0.0 mA to 40 mA
Range B, 200 pF in Parallel with 1.5 k $\Omega$		10		$\mu$ s	S	PPMU enabled in FI, Range B, DCL disabled, step from 0.0 mA to 1 mA
Range C, 200 pF in Parallel with 15.0 k $\Omega$		40		$\mu$ s	S	PPMU enabled in FI, Range C, DCL disabled, step from 0.0 mA to 100 $\mu$ A
FI Settling Time to 1.0% of Final Value						
Range A, 200 pF in Parallel with 120 $\Omega$		8		$\mu$ s	C <sub>B</sub>	PPMU enabled in FI, Range A, DCL disabled, step from 0.0 mA to 40 mA
Range B, 200 pF in Parallel with 1.5 k $\Omega$		8		$\mu$ s	C <sub>B</sub>	PPMU enabled in FI, Range B, DCL disabled, step from 0.0 mA to 1 mA
Range C, 200 pF in Parallel with 15.0 k $\Omega$		8		$\mu$ s	C <sub>B</sub>	PMU enabled in FI, Range C, DCL disabled, step from 0.0 mA to 100 $\mu$ A
<b>INTERACTION AND CROSSTALK</b>						
Measure Voltage Channel to Channel Crosstalk		10		$\mu$ V	C <sub>T</sub>	PPMU enabled in FIMV, Range B, channel under test: force 0.0 mA into 0.0 V; other channel: force 0.0 mA into $V_{DUTx}$ ; sweep $V_{DUTx}$ from -1.5 V to +4.5 V; measure $\Delta V$ at PPMU_Mx under test
Measure Current Channel to Channel Crosstalk		0.0001		%FSR	C <sub>T</sub>	PPMU enabled in FVMI, Range B; channel under test: force 0.0 V into open circuit; other channel: force 0.0 V into $I_{DUTx}$ ; sweep $I_{DUTx}$ from -1.0 mA to +1.0 mA; measure $\Delta V$ at PPMU_Mx under test

## SPECIFICATIONS

- <sup>1</sup> PCHx offset is derived from measurements at DAC Code 0xA000 (3.75 V or 50% FS) and DAC Code 0xB333 (4.50 V or 80% FS), with PCLx at DAC Code 0x0000 (-2.5 V). PCLx offset is derived from measurements at DAC Code 0x6000 (1.25 V or -50% FS) and DAC Code 0x4CCC (0.50 V or -80% FS), with PCHx at DAC Code 0xFFFF (7.5 V).
- <sup>2</sup> PCHx gain is derived from the measurements at DAC Code 0xA000 (3.75 V or 50% FS) and DAC Code 0xB333 (4.50 V or 80% FS), with PCLx at DAC Code 0x0000 (-2.5 V). PCLx gain is derived from measurements at DAC Code 0x6000 (1.25 V or -50% FS) and DAC Code 0x4CCC (0.50 V or -80% FS), with PCHx at DAC Code 0xFFFF (7.5 V). For example, the ideal gain is  $\pm$ FS per 2.5 V in all ranges; in Range B, the ideal gain is  $\pm$ 400  $\mu$ A/V; therefore, 30% error is  $\pm$ 520  $\mu$ A/V.

## PPMU GO/NO-GO COMPARATORS SPECIFICATIONS

Table 8.

Parameter	Min	Typ	Max	Unit	Test Level	Test Conditions/Comments
DC SPECIFICATIONS						
Compare Voltage Range	-1.5		+5.0	V	D	Measured at DAC Code 0x4000 (0 V)
Input Offset Voltage	-250		+250	mV	P	
Input Offset Voltage TC		$\pm$ 100		$\mu$ V/ $^{\circ}$ C	C <sub>T</sub>	Gain derived from measurements at DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V)
Gain	1.0		1.1	V/V	P	
Gain TC		$\pm$ 10		ppm/ $^{\circ}$ C	C <sub>T</sub>	Gain derived from measurements at DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V)
Comparator Threshold Resolution		153		$\mu$ V	D	After two-point calibration; measured over POHx/POLx range -1.5 V to +5.0 V; calibration points DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V)
Comparator Threshold DNL		$\pm$ 250		$\mu$ V	C <sub>T</sub>	
Comparator Threshold INL	-7		+7	mV	P	After two-point calibration; measured over POHx/POLx range -1.5 V to +5.0 V; calibration points DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V)
DUTGND Voltage Accuracy	-5	$\pm$ 1	+5	mV	P	Over $\pm$ 0.1 V range

## PPMU EXTERNAL SENSE PINS SPECIFICATIONS

Table 9.

Parameter	Min	Typ	Max	Unit	Test Level	Test Conditions/Comments
DC SPECIFICATIONS						
Voltage Range	-1.5		+4.5	V	D	PPMU input select, in all states
Leakage	-2	0.0	+2	nA	P	Tested at -1.5 V and +4.5 V
Maximum Load Capacitance	2000			pF	S	Capacitive load tolerated at DUTx sense pins

## VREF, VREFGND, AND DUTGND REFERENCE INPUT PINS SPECIFICATIONS

Table 10.

Parameter	Min	Typ	Max	Unit	Test Level	Test Conditions/Comments
DC SPECIFICATIONS						
VREF Input Voltage Range	2.475	2.500	2.525	V	D	Provided externally, V <sub>REF</sub> = 2.500 V, V <sub>REFGND</sub> = 0.000 V
VREF Input Bias Current			10	$\mu$ A	P	Tested with 2.500 V applied
DUTGND Input Voltage Range, Referenced to AGND	-0.1		+0.1	V	D	
DUTGND Input Bias Current	-10		+10	$\mu$ A	P	Tested at -100 mV and +100 mV

## SPECIFICATIONS

## TEMPERATURE MONITOR SPECIFICATIONS

Table 11.

Parameter	Min	Typ	Max	Unit	Test Level	Test Conditions/Comments
DC SPECIFICATIONS						
Temperature Sensor Gain		10		mV/K	D	3.00 V at room temperature, 300 K (27°C)
Temperature Sensor Accuracy		±10		°C	C <sub>T</sub>	20°C < T <sub>C</sub> < 80°C, V <sub>CC</sub> THERM only (T <sub>J</sub> = T <sub>C</sub> )

## ALARM FUNCTIONS SPECIFICATIONS

Table 12.

Parameter	Min	Typ	Max	Unit	Test Level	Test Conditions/Comments
DC SPECIFICATIONS						
Overvoltage Alarm High, OVDH						
Functional Voltage Range	-1.0		+5.0	V	D	OVDL DAC set to DAC Code 0x0000 (-2.5 V)
Uncalibrated Error at -1.0 V	-300		+200	mV	P	Includes 5% uncalibrated gain ±250 mV offset
Uncalibrated Error at 5.0 V	0		500	mV	P	Includes 5% uncalibrated gain ±250 mV offset
Offset Voltage TC		±0.5		mV/°C	C <sub>T</sub>	
Gain		1.05		V/V	C <sub>T</sub>	Gain derived from measurements at DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V); based on an ideal DAC transfer function (see Table 24)
Hysteresis		140		mV	C <sub>T</sub>	Hysteresis is only applied coming out of alarm
Overvoltage Alarm Low, OVDL						
Functional Voltage Range	-2.0		+4.0	V	D	OVDH DAC set to DAC Code 0xFFFF (7.5 V)
Uncalibrated Error at -2.0 V	-350		+150	mV	P	Includes 5% uncalibrated gain ±250 mV offset
Uncalibrated Error at 4.0 V	-50		+450	mV	P	Includes 5% uncalibrated gain ±250 mV offset
Offset Voltage TC		±0.5		mV/°C	C <sub>T</sub>	
Gain		1.05		V/V	C <sub>T</sub>	Gain derived from measurements at DAC Code 0x4000 (0.0 V) and DAC Code 0x8CCC (3.0 V); based on an ideal DAC transfer function (see Table 24)
Hysteresis		140		mV	C <sub>T</sub>	Hysteresis is only applied coming out of alarm
Thermal Alarm						
Setpoint Error		±10		°C	C <sub>T</sub>	Relative to default alarm value, T <sub>J</sub> = 100°C
Thermal Hysteresis		15		°C	C <sub>T</sub>	
ALARM Output Characteristics						
Off State Leakage		10	500	nA	P	Disable alarm, apply V <sub>DD</sub> to $\overline{\text{ALARM}}$ pin, and measure leakage current
Maximum On Voltage at 200 $\mu$ A		0.1	0.7	V	P	$\overline{\text{ALARM}}$ pin asserted, force 200 $\mu$ A into pin and measure voltage
AC SPECIFICATIONS						
Propagation Delay		0.5		$\mu$ s	C <sub>B</sub>	For OVDH: V <sub>DUTX</sub> = 0.0 V to 4.5 V step, OVDH = 4.0 V, OVDL = -1.0 V; for OVDL: V <sub>DUTX</sub> = 0.0 V to -1.5 V step, OVDH = 4.0 V, OVDL = -1.0 V

## SERIAL PROGRAMMABLE INTERFACE (SPI) SPECIFICATIONS

Table 13.

Parameter	Min	Typ	Max	Unit	Test Level	Test Conditions/Comments
DC SPECIFICATIONS						
Input Voltage						$\overline{\text{RST}}$ , $\overline{\text{CS}}$ , SCLK, SDI
Logic High	V <sub>DD</sub> - 0.7		V <sub>DD</sub>	V	P <sub>F</sub>	
Logic Low	0.0		0.7	V	P <sub>F</sub>	

## SPECIFICATIONS

Table 13. (Continued)

Parameter	Min	Typ	Max	Unit	Test Level	Test Conditions/Comments
Input Bias Current	-10	1	+10	μA	P	Tested at 0.0 V and $V_{DD}$ ; $\overline{RST}$ tested at $V_{DD}$ ; $\overline{RST}$ has an internal 50 kΩ pull-up to $V_{DD}$
SCLK Crosstalk on DUTx Pin		1		mV	$C_B$	DCL disabled, PPMU forcing 0.0 V
Serial Output						
Logic High	$V_{DD} - 0.5$		$V_{DD}$	V	$P_F$	SDO, sourcing 2 mA
Logic Low	0.0		0.5	V	$P_F$	Sinking 2 mA
$\overline{BUSY}$ Output Characteristics						Open-drain output
Off State Leakage		10	500	nA	P	$\overline{BUSY}$ pin not asserted, apply $V_{DD}$ to pin and measure leakage current
Maximum On Voltage at 2 mA		0.01	0.7	V	P	$\overline{BUSY}$ pin asserted, force 2 mA into pin and measure voltage

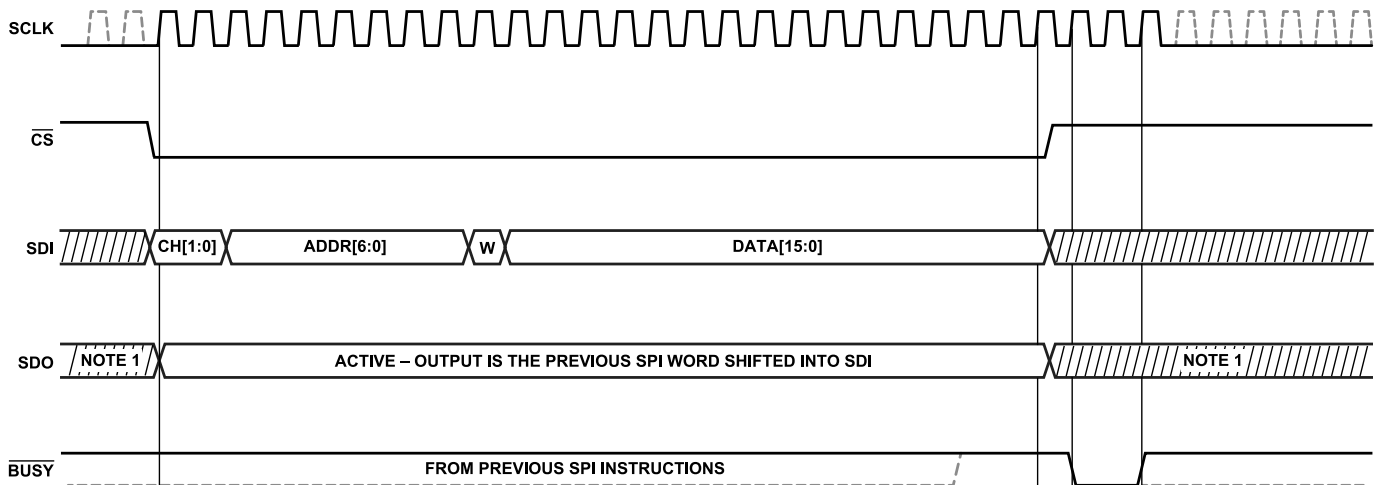
## SPI TIMING SPECIFICATIONS

Table 14.

Parameter	Symbol	Min	Typ	Max	Unit	Test Level	Description
SCLK Operating Frequency	$f_{CLK}$		50		MHz	$P_F$	
		0.5		100	MHz	S	
SCLK High Time	$t_{CH}$	4.5			ns	S	
SCLK Low Time	$t_{CL}$	4.5			ns	S	
$\overline{CS}$ to SCLK Setup at Assert	$t_{CSAS}$	1.5			ns	S	Setup time of $\overline{CS}$ assert to next rising edge of SCLK.
$\overline{CS}$ to SCLK Hold at Assert	$t_{CSAH}$	1.5			ns	S	Hold time of $\overline{CS}$ assert to next rising edge of SCLK.
$\overline{CS}$ to SCLK Setup at Release	$t_{CSRS}$	1.5			ns	S	Setup time of $\overline{CS}$ release to next rising edge of SCLK.
$\overline{CS}$ to SCLK Hold at Release	$t_{CSRH}$	1.5			ns	S	Hold time of $\overline{CS}$ release to next rising edge of SCLK. This parameter is only critical if the number of SCLK cycles from previous release of $\overline{CS}$ is the minimum specified by the $t_{CSAM}$ parameter.
$\overline{CS}$ Assert to SDO Active	$t_{CSO}$	0		4	ns	S	Delay time from $\overline{CS}$ assert to SDO active state.
$\overline{CS}$ Release to SDO High-Z	$t_{CSZ}$	0		11	ns	S	Delay from $\overline{CS}$ release to SDO high-Z state, strongly influenced by external SDO pin loading.
$\overline{CS}$ Release to Next Assert	$t_{CSAM}$	3			Cycles	D	Minimum release time of $\overline{CS}$ between consecutive assertions of $\overline{CS}$ . This parameter is specified in units of SCLK cycles, more specifically in terms of rising edges of the SCLK input.
SDI to SCLK Setup	$t_{DS}$	3			ns	S	Setup time of SDI data prior to next rising edge of SCLK.
SDI to SCLK Hold	$t_{DH}$	4			ns	S	Hold time of SDI data following previous rising edge of SCLK.
SCLK to Valid SDO	$t_{DO}$	0		6	ns	S	Propagation delay from rising edge of SCLK to valid SDO data.
$\overline{BUSY}$ Assert from $\overline{CS}/\overline{RST}$	$t_{BUSA}$	0		6	ns	S	Propagation delay from first rising SCLK following valid $\overline{CS}$ release (or $\overline{RST}$ release in the case of hardware reset) to $\overline{BUSY}$ assert.
$\overline{BUSY}$ Width	$t_{BUSW}$						
Following $\overline{CS}$		3		21	Cycles	D	Delay time from first rising SCLK after valid $\overline{CS}$ release to $\overline{BUSY}$ release. Satisfies the requirements detailed in the <a href="#">SPI Clock Cycles</a> and the <a href="#">BUSY Pin</a> section, except following $\overline{RST}$ or software reset.
Following $\overline{RST}$		744			Cycles	D	Delay time from first rising SCLK after $\overline{RST}$ release (or valid $\overline{CS}$ release in the case of software reset) to $\overline{BUSY}$ release. Satisfies the requirement of synchronous reset sequence detailed in the <a href="#">SPI Clock Cycles</a> and the <a href="#">BUSY Pin</a> section.



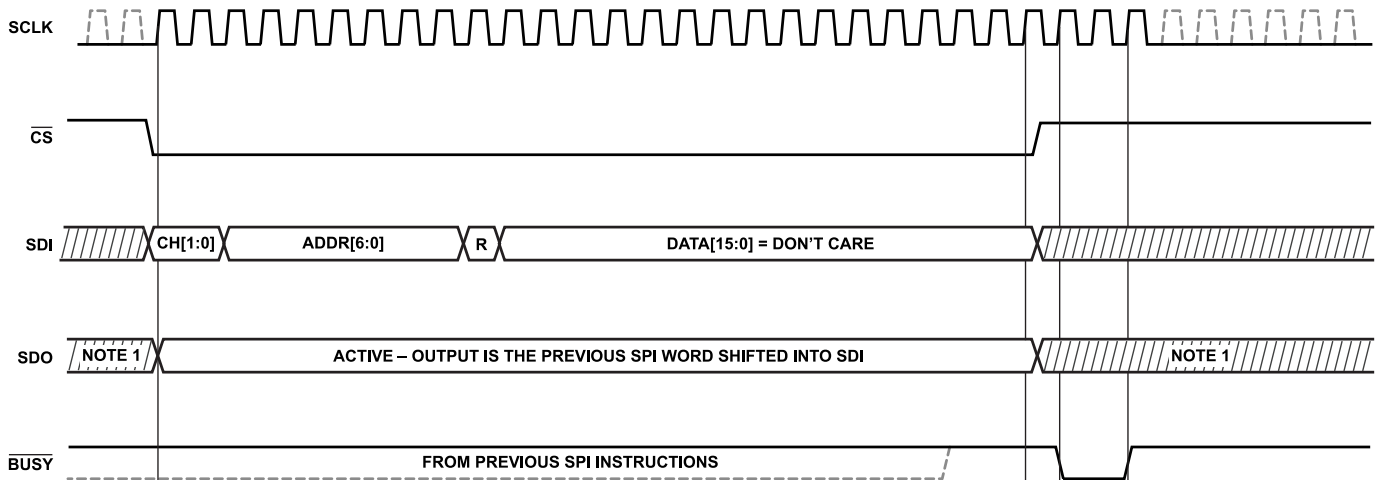
SPECIFICATIONS



NOTES

- 1. IF THE SPI\_SDO\_HIZ CONTROL BIT (ADDR 0x12 [1]) IS HIGH, THE SDO PIN BECOMES ACTIVE FOLLOWING THE ASSERTION OF CS. IT BECOMES HIGH-Z FOLLOWING RELEASE OF CS. IF THE SPI\_SDO\_HIZ CONTROL BIT IS LOW, THE SDO PIN REMAINS ACTIVE INDEPENDENT OF CS.

Figure 3. SPI Write Instruction Timing Diagram

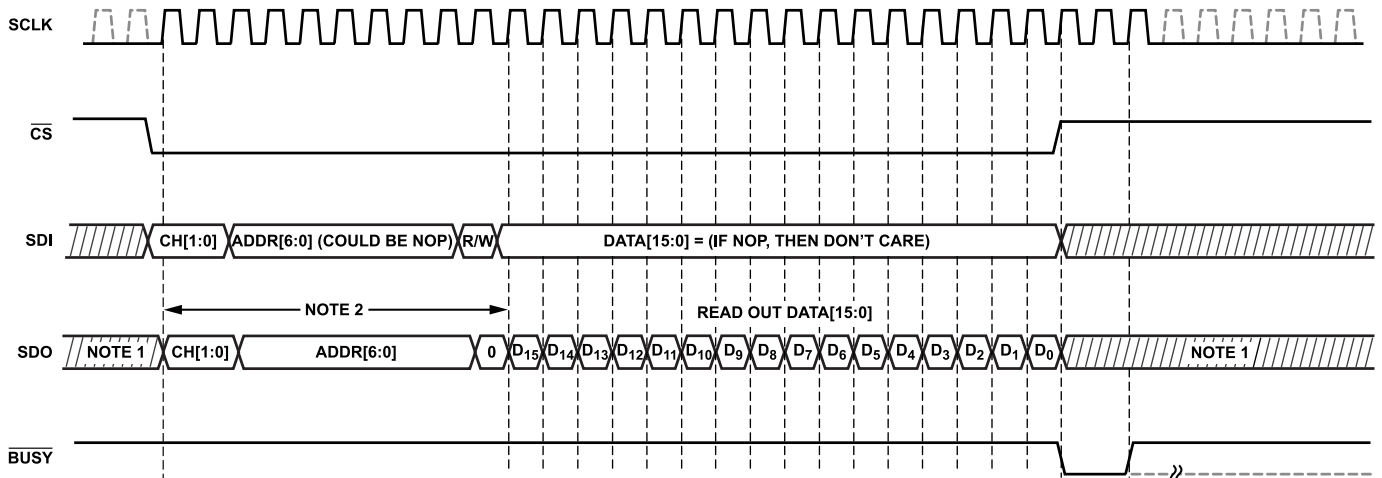


NOTES

- 1. IF THE SPI\_SDO\_HIZ CONTROL BIT (ADDR 0x12 [1]) IS HIGH, THE SDO PIN BECOMES ACTIVE FOLLOWING THE ASSERTION OF CS. IT BECOMES HIGH-Z FOLLOWING RELEASE OF CS. IF THE SPI\_SDO\_HIZ CONTROL BIT IS LOW, THE SDO PIN ALWAYS REMAINS ACTIVE INDEPENDENT OF CS.

Figure 4. SPI Read Request Instruction Timing Diagram (Prior to Readout Instruction)

SPECIFICATIONS



NOTES

1. IF THE SPI\_SDO\_HIZ CONTROL BIT (ADDR 0x12 [1]) IS HIGH, THE SDO PIN BECOMES ACTIVE FOLLOWING THE ASSERTION OF CS. IT BECOMES HIGH-Z FOLLOWING RELEASE OF CS. IF THE SPI\_SDO\_HIZ CONTROL BIT IS LOW, THE SDO PIN REMAINS ACTIVE INDEPENDENT OF CS.
2. THE FIRST 10 BITS OF SDO FOLLOWING A READ REQUEST ECHO ADDRESS AND CHANNEL BITS OF THE PRECEDING REQUEST. THE R/W BIT POSITION IS SET LOW. THE FOLLOWING 16 BITS CONTAIN DATA FROM THE REQUESTED ADDRESS AND CHANNEL.

Figure 5. SPI Readout Instruction Timing Diagram (Subsequent to Read Request Instruction)

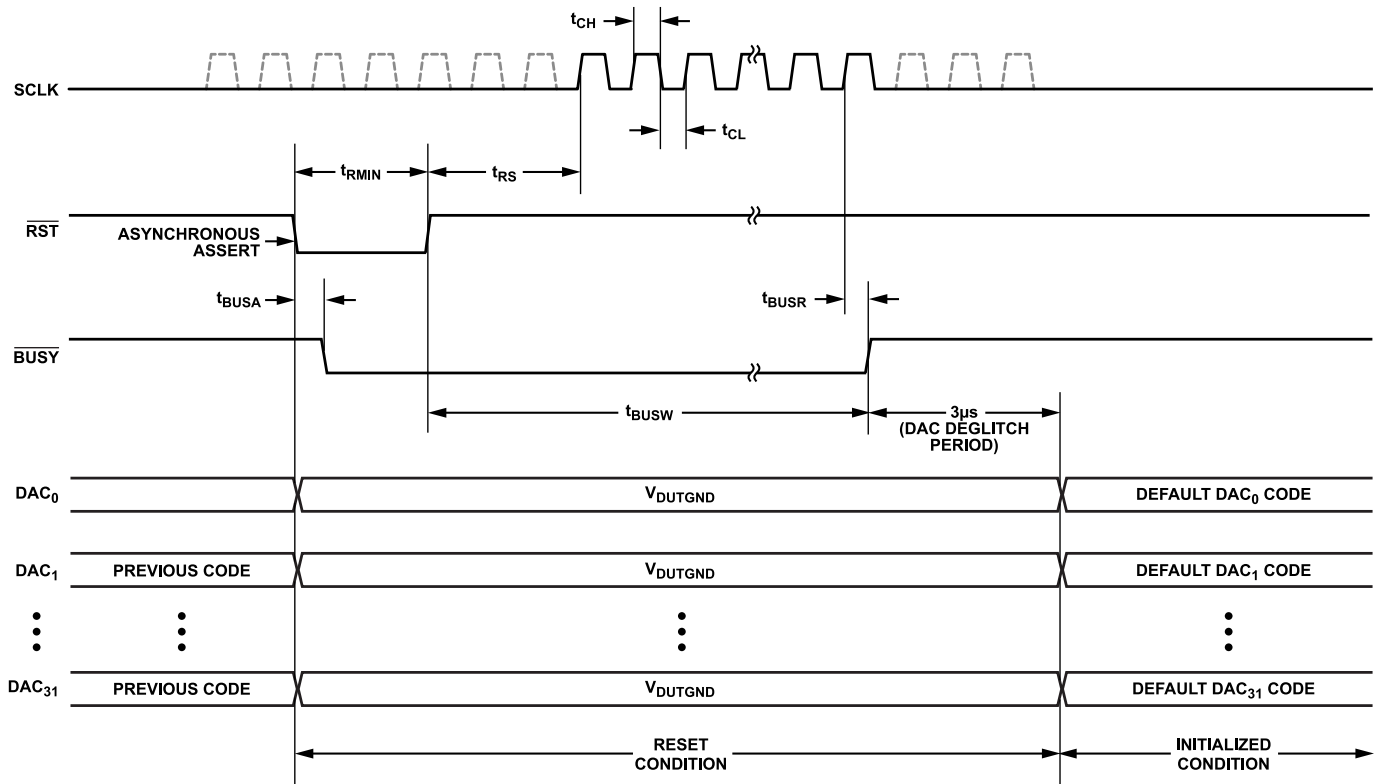


Figure 6. SPI Detailed Hardware Reset Timing Diagram

SPECIFICATIONS

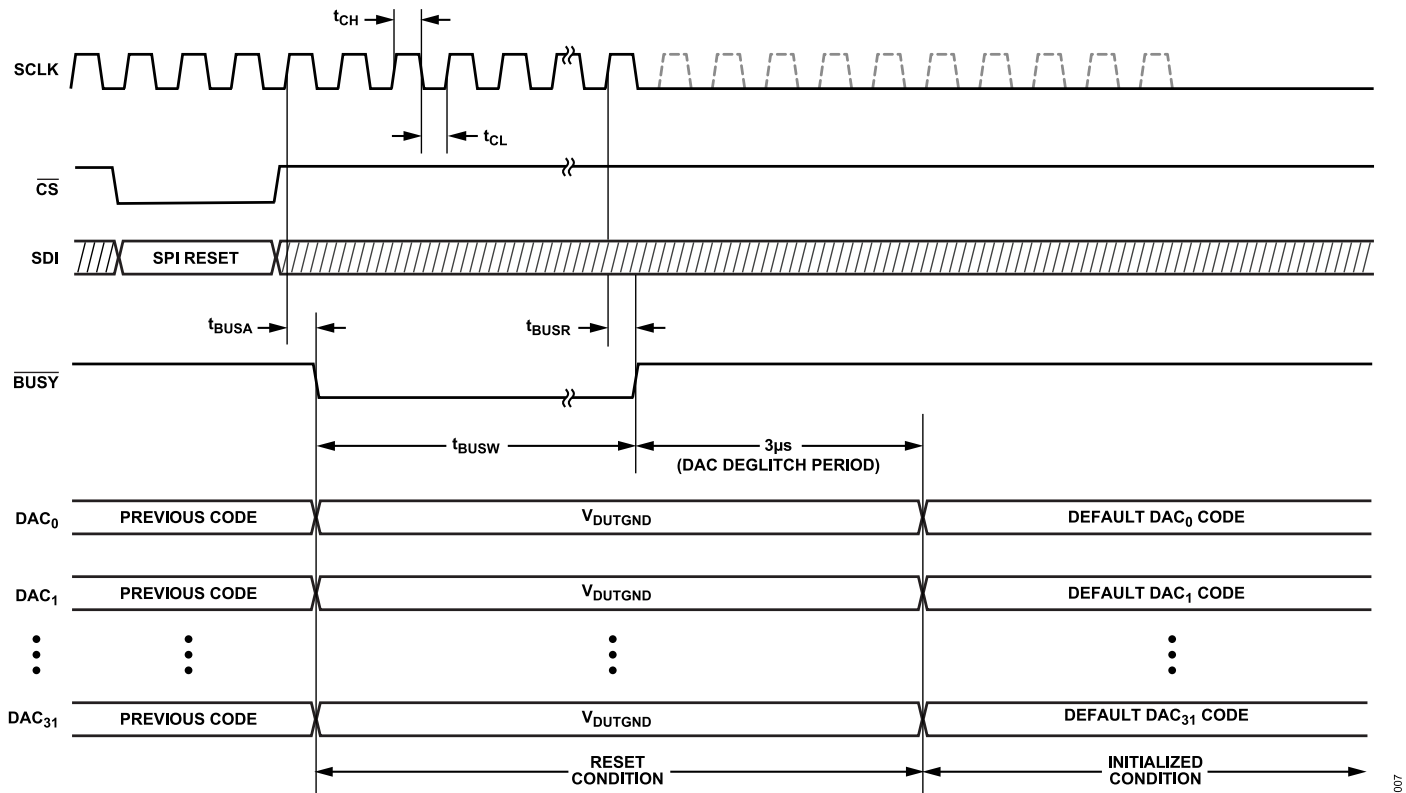


Figure 7. SPI Detailed Software Reset Timing Diagram



## ABSOLUTE MAXIMUM RATINGS

Table 15.

Parameter	Rating
Supply Voltages	
Positive Supply Voltage ( $V_{CC}$ to PGND)	-0.5 V to +9.0 V
Positive Supply Voltage ( $V_{DD}$ to DGND)	-0.5 V to +2.2 V
Negative Supply Voltage ( $V_{EE}$ to PGND)	-6.0 V to +0.5 V
Supply Voltage Difference ( $V_{CC}$ to $V_{EE}$ )	-1.0 V to +15.0 V
Reference Ground (DUTGND to AGND)	-0.5 V to +0.5 V
Supply Sequence or Dropout Condition	No limitations
Input/Output Voltages	
Digital Input Voltage Range	-0.5 V to $V_{DD} + 0.5$ V
VREF Input Voltage Range	-0.5 V to +3.5 V
VREFGND, DUTGND Input Voltage Range	-0.5 V to +0.5 V
DUTx Output Short-Circuit Voltage <sup>1</sup>	-3.0 V to +6.0 V
High Speed Termination (VTTCx, VTDDx) Input Voltage Range	-0.5 V to +2.2 V
High Speed DATx/RCVx Common-Mode Input Voltage Range <sup>2</sup>	-0.5 V to +2.2 V
High Speed DATx/RCVx Differential Mode Input Voltage Range <sup>2</sup>	-1.0 V to +1.0 V
High Speed CMPHx/CHPLx, PPMU_CMPHx/PPMU_CMPLx Absolute Output Voltage Range	-0.5 V to +2.2 V
DUTx Input/Output Pin Current Limit	
DCL Maximum Short-Circuit Current <sup>3</sup>	$\pm 120$ mA
Operating Temperature, Junction	125°C
Storage Temperature Range	-65°C to +150°C

<sup>1</sup>  $R_L = 0 \Omega$ ,  $V_{DUTx}$  continuous short-circuit condition ( $V_{IH}$ ,  $V_{IL}$ ,  $V_{IT}$ ), high-Z, VCOM, and all clamp modes.

<sup>2</sup> DATx,  $\overline{DATx}$ , RCVx,  $\overline{RCVx}$ ,  $R_{SOURCE} = 0 \Omega$ , no pin to exceed either maximum common-mode input range or differential mode input range.

<sup>3</sup>  $R_L = 0 \Omega$ ,  $V_{DUTx} = -3$  V to +6 V; DCL current limit. Continuous short-circuit condition. The ADATE320 is designed to withstand continuous short-circuit fault.

## USER INFORMATION AND TRUTH TABLES

Table 17. Driver Truth Table

DRIVE_ENABLE_x, Address 0x19, Bit 0	DRV Control Register <sup>1</sup>			High Speed Inputs <sup>2</sup>		Driver State
	DRIVE_FORCE_x, Address 0x19, Bit 1	DRIVE_FORCE_STATE_x, Address 0x19, Bits[3:2]	DRIVE_VT_HIZ_x, Address 0x19, Bit 4	DATx	RCVx	
0	X	XX	X	X	X	Low leakage
1	1	00	X	X	X	Active VIL
1	1	01	X	X	X	Active VIH
1	1	10	X	X	X	Active high-Z
1	1	11	X	X	X	Active VIT
1	0	XX	0	X	1	Active high-Z
1	0	XX	1	X	1	Active VIT
1	0	XX	X	0	0	Active VIL
1	0	XX	X	1	0	Active VIH

<sup>1</sup> X means don't care.

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

## THERMAL CHARACTERISTICS

$\theta_{JA}$  is specified for the worst case conditions, that is, a device soldered in a circuit board for surface-mount packages.

Table 16. Thermal Resistance

Package Type	Airflow Velocity (m/sec)	$\theta_{JA}$	$\theta_{JC}$	Unit
84-Lead LFCSP	N/A <sup>1</sup>	N/A <sup>1</sup>	3.2	°C/W
	0	45	N/A <sup>1</sup>	°C/W
	1	40	N/A <sup>1</sup>	°C/W
	2	37	N/A <sup>1</sup>	°C/W

<sup>1</sup> N/A means not applicable.

## EXPLANATION OF TEST LEVELS

D Definition.

S Design verification simulation.

P 100% production tested.

P<sub>F</sub> Functionally checked during production test.

C<sub>T</sub> Characterized on tester.

C<sub>B</sub> Characterized on bench.

## ABSOLUTE MAXIMUM RATINGS

<sup>2</sup> See Figure 138 for more detailed information about high speed DATx/RCVx input multiplexing.

Table 18. Comparator Truth Table

DMC_ENABLE, Address 0x1A, Bit 0	Comparator State							
	CMPH0	State	CMPL0	State	CMPH1	State	CMPL1	State
0	$V_{DUT0} < VOH0$	0	$V_{DUT0} < VOL0$	0	$V_{DUT1} < VOH1$	0	$V_{DUT1} < VOL1$	0
	$VOH0 < V_{DUT0}$	1	$VOL0 < V_{DUT0}$	1	$VOH1 < V_{DUT1}$	1	$VOL1 < V_{DUT1}$	1
1 <sup>1</sup>	$V_{DUT0} - V_{DUT1} < VOH0$	0	$V_{DUT0} - V_{DUT1} < VOL0$	0	$V_{DUT1} < VOH1$	0	$V_{DUT1} < VOL1$	0
	$VOH0 < V_{DUT0} - V_{DUT1}$	1	$VOL0 < V_{DUT0} - V_{DUT1}$	1	$VOH1 < V_{DUT1}$	1	$VOL1 < V_{DUT1}$	1

<sup>1</sup> Note that the Channel 1 normal window comparator continues to function while the device is in differential compare mode, but at a greatly reduced bandwidth.

Table 19. Active Load Truth Table

LOAD_ENABLE_x, Address 0x1B, Bit 0	LOAD/DRV Control Registers <sup>1</sup>			High Speed Inputs <sup>2</sup>		Load State
	LOAD_FORCE_x, Address 0x1B, Bit 1	DRIVE_VT_HIZ_x, Address 0x19, Bit 4	DATx	RCVx		
0	X	X	X	X	X	Low leakage
1	1	X	X	X	X	Active on
1	0	X	X	X	0	Active off
1	0	0	X	X	1	Active on
1	0	1	X	X	1	Active off

<sup>1</sup> X means don't care.

<sup>2</sup> See Figure 138 for more detailed information about high speed DATx/RCVx input multiplexing.

Table 20. PPMU Go/No-Go Comparator Truth Table

PPMU_ENABLE_x, Address 0x1C, Bit 0	PPMU Control Register <sup>1</sup>		PPMU Go/No-Go Comparator State <sup>2</sup>		
	PPMU_STANDBY_x, Address 0x1C, Bit 1	PPMU_CMPHx	State	PPMU_CMPLx	State
0	X	X	0	X	0
1	X	PPMUx MV/MI < POHx	0	PPMUx MV/MI < POLx	0
1	X	POHx < PPMUx MV/MI	1	POLx < PPMUx MV/MI	1

<sup>1</sup> X means don't care.

<sup>2</sup> The PPMUx MV/MI inputs to the PPMU go/no-go comparators always come directly from the respective internal PPMU instrumentation amplifiers, not from the PPMU\_Mx output pins (see Figure 143). The internal instrumentation amplifiers are independently configured for either measure voltage (MV) or measure current (MI), depending on the settings of the PPMU\_MEAS\_VI\_x control bit, as described in Figure 150. When PPMU power is not enabled, the respective go/no-go comparator outputs are locked to a static low state (see Table 21).

Table 21. PPMU Measure Pin Truth Table

PPMU_ENABLE_x, Address 0x1C, Bit 0	PPMU Control Register <sup>1</sup>					PPMU_Mx, Pin State
	PPMU_STANDBY_x, Address 0x1C, Bit 1	PPMU_MEAS_ENABLE_x, Address 0x1C, Bit 13	PPMU_MEAS_SEL_x, Address 0x1C, Bit 14	PPMU_MEAS_VI_x, Address 0x1C, Bit 6		
X	X	0	X	X	High-Z	
0	X	1	0	X	Active MV	
0	X	1	1	X	Active VTHERM <sup>2</sup>	
1	X	1	0	0	Active MV	
1	X	1	0	1	Active MI	
1	X	1	1	X	Active VTHERM <sup>2</sup>	

<sup>1</sup> X means don't care.

## ABSOLUTE MAXIMUM RATINGS

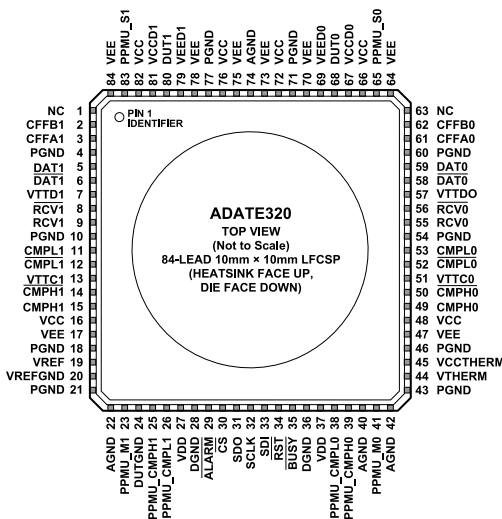
<sup>2</sup> When applicable, PPMU\_M0 is connected to the internal temperature sensor node (VTHERM), and PPMU\_M1 is connected to the internal temperature sensor reference ground node (AGND) (see [Figure 143](#)).

## ESD CAUTION



**ESD (electrostatic discharge) sensitive device.** Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS



NOTES  
 1. NC = NO CONNECT.  
 2. THE EXPOSED PAD IS INTERNALLY CONNECTED VIA A HIGH IMPEDANCE DIE ATTACHED TO VEE (SUBSTRATE).

Figure 8. Pin Configuration

Table 22. Pin Function Descriptions

Pin No.	Mnemonic	Description
59	DAT0	Driver High Speed Data Input, Channel 0.
58	$\overline{\text{DAT0}}$	Driver High Speed Data Input Complement, Channel 0.
57	VTDD0	Driver High Speed Input Termination, Channel 0.
55	RCV0	Driver High Speed Receive Input, Channel 0.
56	$\overline{\text{RCV0}}$	Driver High Speed Receive Input Complement, Channel 0.
5	DAT1	Driver High Speed Data Input, Channel 1.
6	$\overline{\text{DAT1}}$	Driver High Speed Data Input Complement, Channel 1.
7	VTDD1	Driver High Speed Input Termination, Channel 1.
9	RCV1	Driver High Speed Receive Input, Channel 1.
8	$\overline{\text{RCV1}}$	Driver High Speed Receive Input Complement, Channel 1.
53	Cmpl0	Comparator High Speed Output Low, Channel 0.
52	$\overline{\text{Cmpl0}}$	Comparator High Speed Output Low Complement, Channel 0.
51	VTTC0	Comparator High Speed Output Termination, Channel 0.
49	CMPH0	Comparator High Speed Output High, Channel 0.
50	$\overline{\text{CMPH0}}$	Comparator High Speed Output High Complement, Channel 0.
11	Cmpl1	Comparator High Speed Output Low, Channel 1.
12	$\overline{\text{Cmpl1}}$	Comparator High Speed Output Low Complement, Channel 1.
13	VTTC1	Comparator High Speed Output Termination, Channel 1.
15	CMPH1	Comparator High Speed Output High, Channel 1.
14	$\overline{\text{CMPH1}}$	Comparator High Speed Output High Complement, Channel 1.
61	CFFA0	PPMU External Compensation Capacitor Pin A, Channel 0.
62	CFFB0	PPMU External Compensation Capacitor Pin B, Channel 0.
65	PPMU_S0	PPMU External Sense Connect, Channel 0.
41	PPMU_M0	PPMU Analog Measure Output, Channel 0.
38	PPMU_Cmpl0	PPMU Go/No-Go Comparator Output Low, Channel 0.
39	PPMU_CMPH0	PPMU Go/No-Go Comparator Output High, Channel 0.
3	CFFA1	PPMU External Compensation Capacitor Pin A, Channel 1.
2	CFFB1	PPMU External Compensation Capacitor Pin B, Channel 1.
83	PPMU_S1	PPMU External Sense Connect, Channel 1.

## PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

Table 22. Pin Function Descriptions (Continued)

Pin No.	Mnemonic	Description
23	PPMU_M1	PPMU Analog Measure Output, Channel 1.
26	PPMU_CMPL1	PPMU Go/No-Go Comparator Output Low, Channel 1.
25	PPMU_CMPH1	PPMU Go/No-Go Comparator Output High, Channel 1.
34	$\overline{\text{RST}}$	Reset Input (Active Low).
32	SCLK	Serial Programmable Interface (SPI) Clock Input.
30	$\overline{\text{CS}}$	Serial Programmable Interface (SPI) Chip Select Input (Active Low).
33	SDI	Serial Programmable Interface (SPI) Serial Data Input.
31	SDO	Serial Programmable Interface (SPI) Serial Data Output.
29	$\overline{\text{ALARM}}$	Fault Alarm Open-Drain Output (Open-Collector, Active Low).
35	$\overline{\text{BUSY}}$	Serial Programmable Interface (SPI) Busy Output (Open-Collector, Active Low).
19	VREF	DAC Precision 2.500 V Reference Input.
20	VREFGND	DAC Precision 0.000 V Reference Input.
24	DUTGND	DUT Ground Sense Input.
68	DUT0	DUT Pin, Channel 0.
80	DUT1	DUT Pin, Channel 1.
45	VCCTHERM	Temperature Sensor VCC Supply (8.0 V).
44	VTHERM	Temperature Sensor Analog Output.
16, 48, 66, 67, 72, 76, 81, 82	VCC, VCCD0, VCCD1	Analog Supply (8.0 V).
27, 37	VDD	Digital Supply (1.8 V).
22, 40, 42, 74	AGND	Analog Ground (Quiet).
28, 36	DGND	Digital Ground.
4, 10, 18, 21, 43, 46, 54, 60, 71, 77	PGND	Power Ground.
17, 47, 64, 69, 70, 73, 75, 78, 79, 84	VEE, VEED0, VEED1	Analog Supply (-5.0 V).
1, 63	NC	No Connect. These pins can be grounded or left floating.
	EP	Exposed Pad. The exposed pad is internally connected via a high impedance die attached to VEE (substrate).

TYPICAL PERFORMANCE CHARACTERISTICS

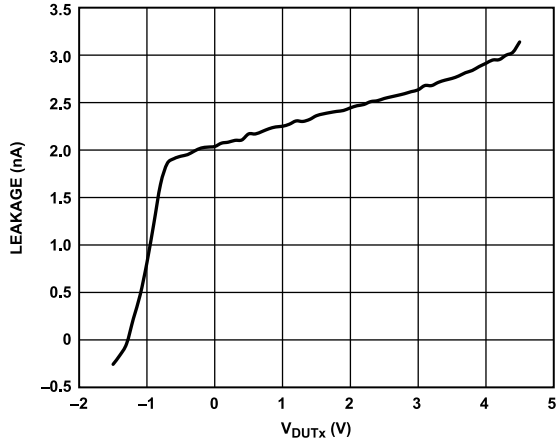


Figure 9. DUTx Pin Leakage in High-Z Mode

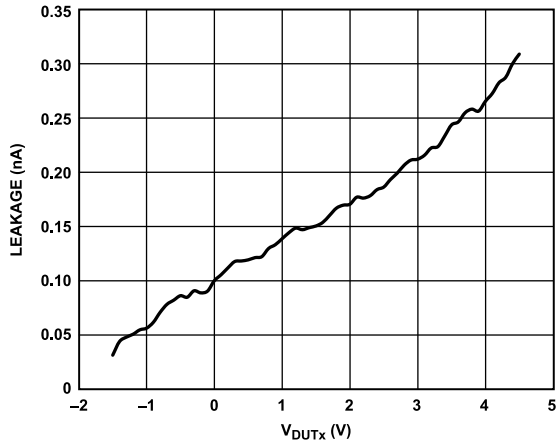


Figure 10. DUTx Pin Leakage in Low Leakage Mode

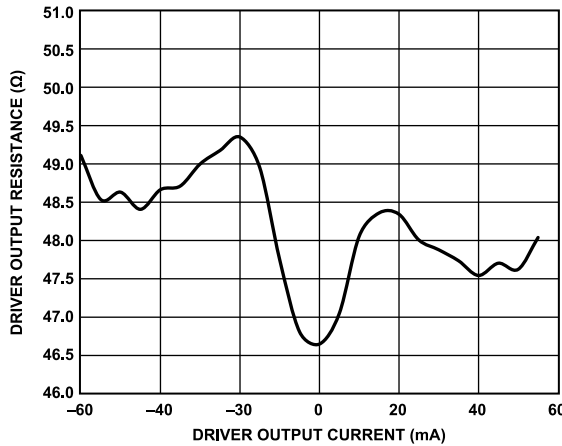


Figure 11. Driver Output Resistance vs. Driver Output Current

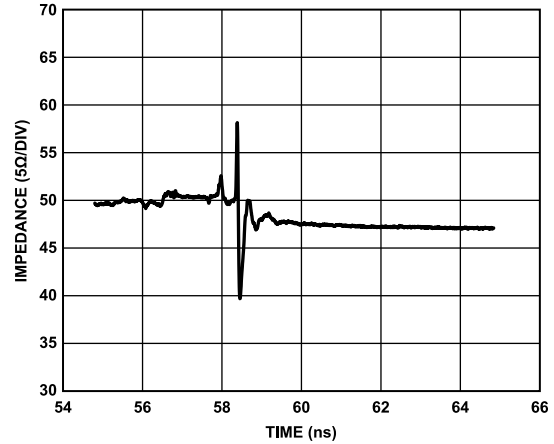


Figure 12. DUTx Pin Time-Domain Reflectometry (TDR) Response

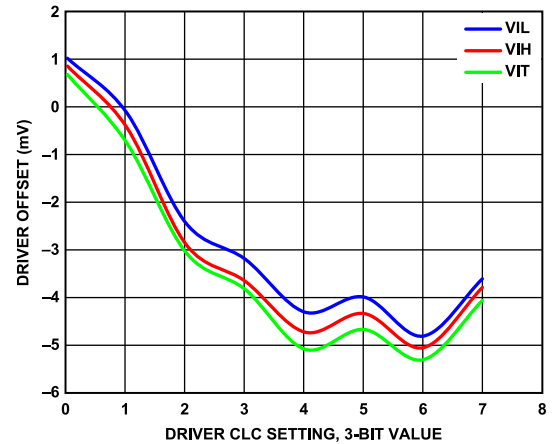


Figure 13. Driver Offset vs. Driver CLC Setting, 3-Bit Value

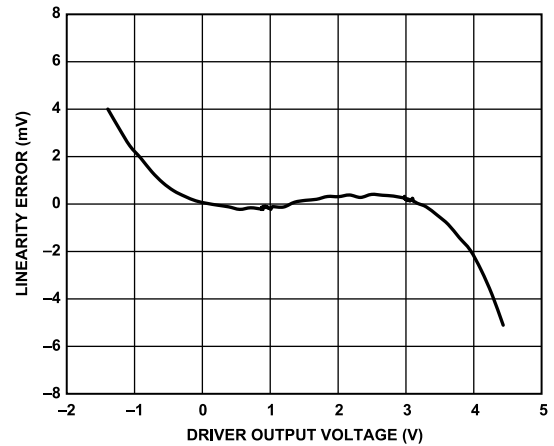


Figure 14. Driver VIH INL

TYPICAL PERFORMANCE CHARACTERISTICS

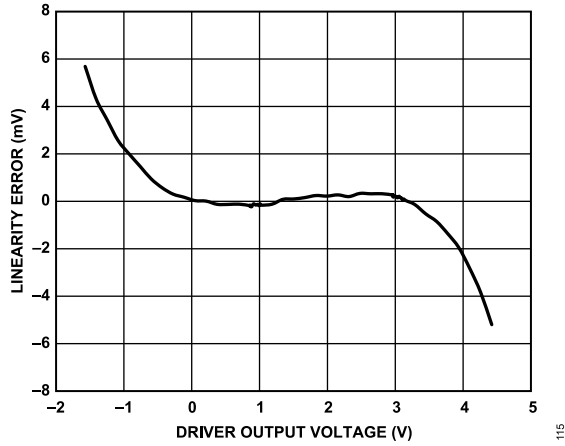


Figure 15. Driver VIL INL

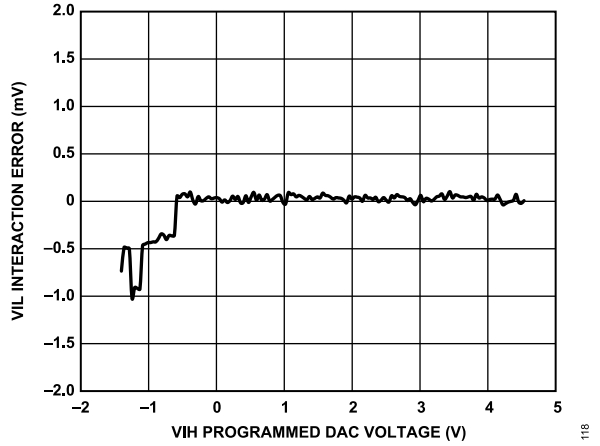


Figure 18. Driver VIL Interaction Error vs. VIH Programmed DAC Voltage

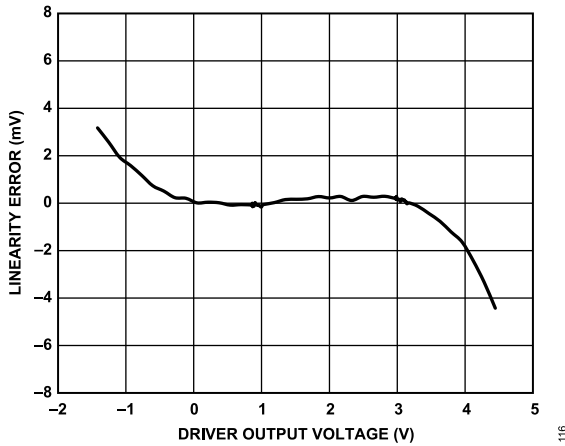


Figure 16. Driver VIT INL

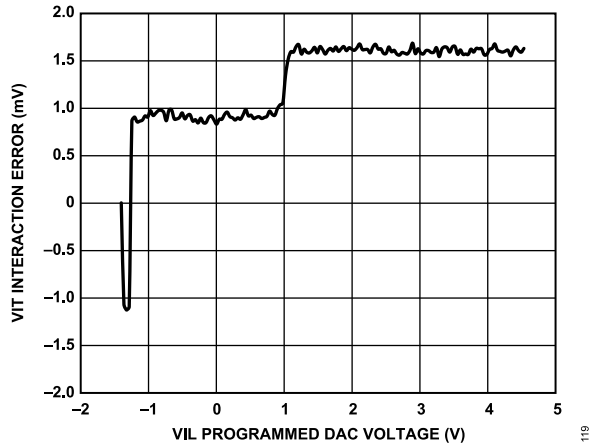


Figure 19. Driver VIT Interaction Error vs. VIH Programmed DAC Voltage

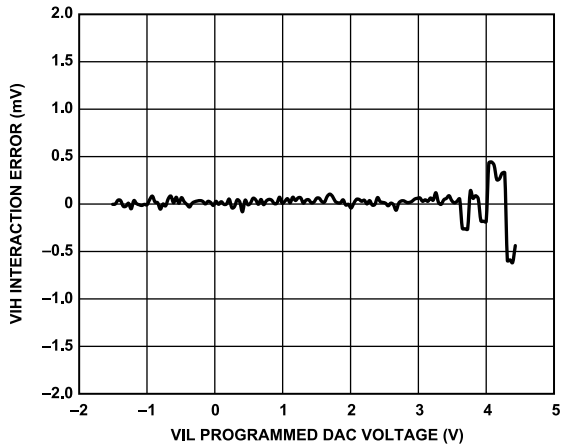


Figure 17. Driver VIH Interaction Error vs. VIL Programmed DAC Voltage

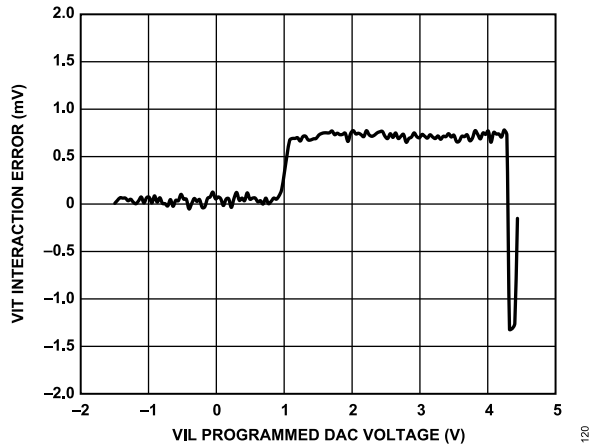


Figure 20. Driver VIT Interaction Error vs. VIL Programmed DAC Voltage

TYPICAL PERFORMANCE CHARACTERISTICS

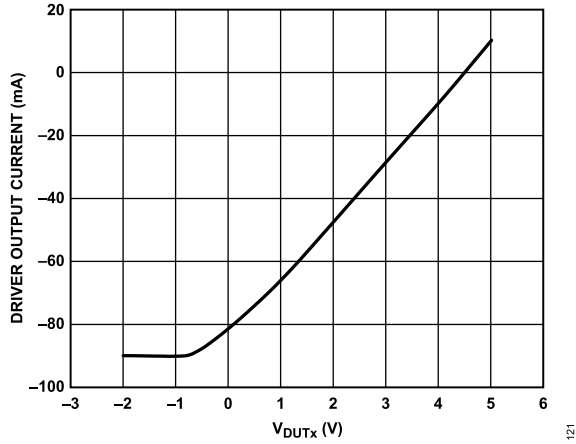


Figure 21. Driver Output Current Limit, Sink

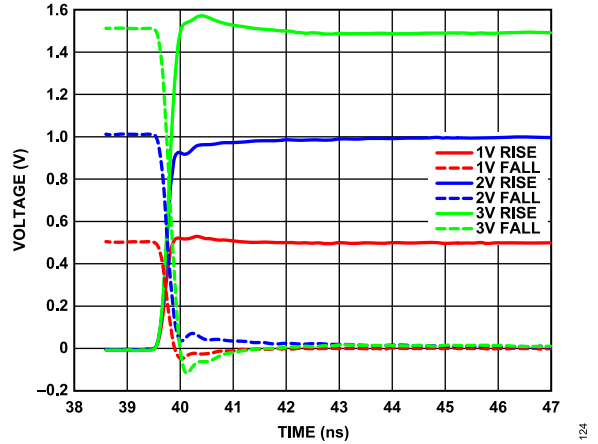


Figure 24. Driver Large Swing Response

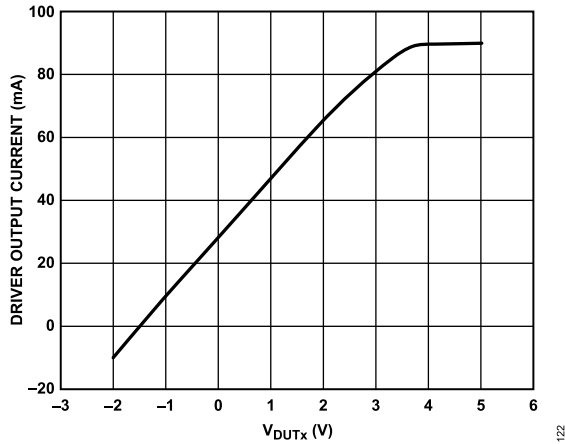


Figure 22. Driver Output Current Limit, Source

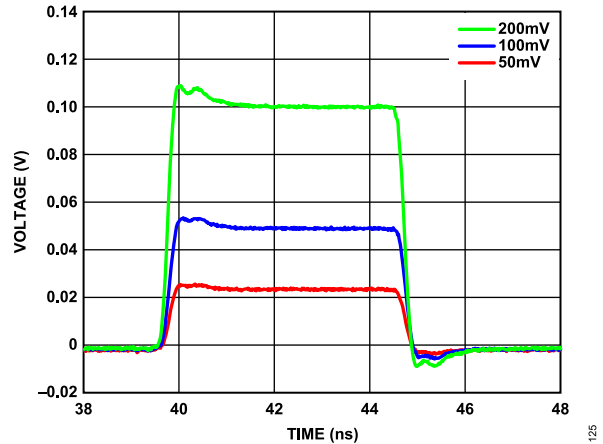


Figure 25. Driver 100 MHz Response, Small Swing

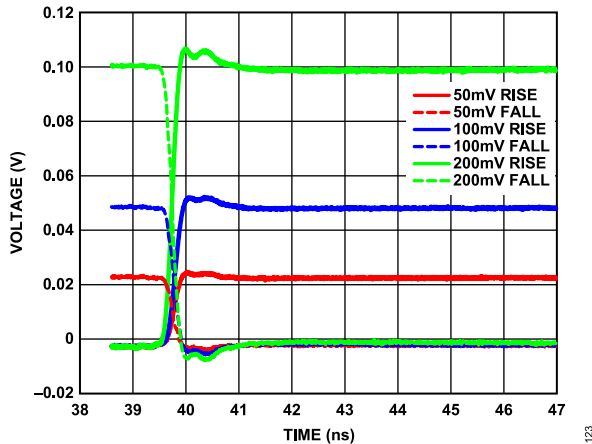


Figure 23. Driver Small Swing Response

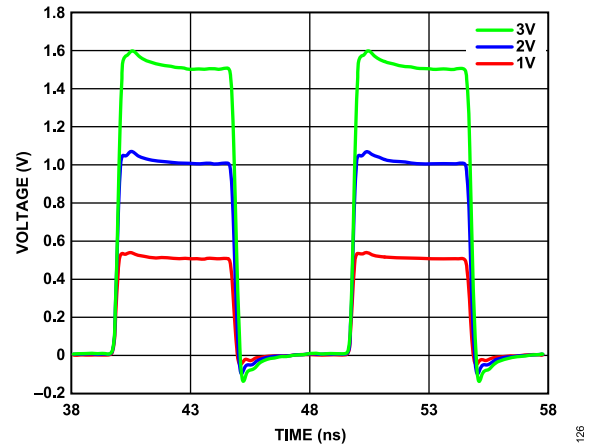


Figure 26. Driver 100 MHz Response, Large Swing



TYPICAL PERFORMANCE CHARACTERISTICS

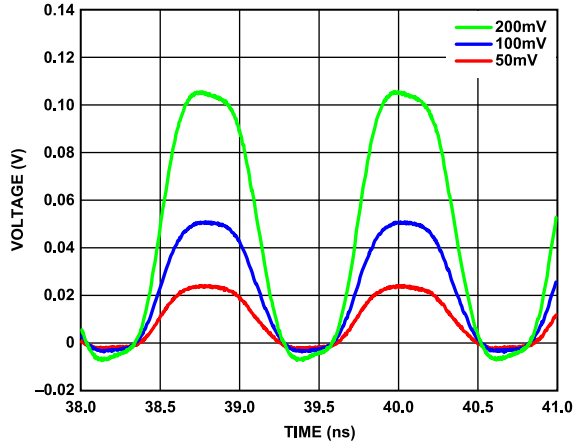


Figure 27. Driver 800 MHz Response, Small Swing

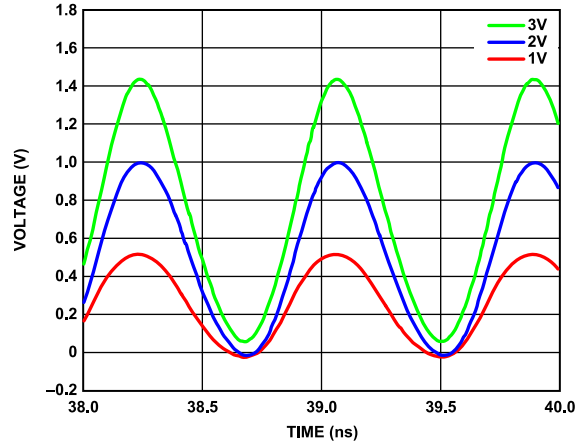


Figure 30. Driver 1.25 GHz Response, Large Swing

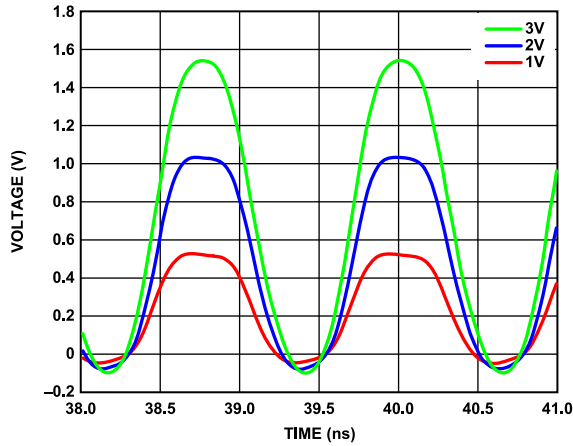


Figure 28. Driver 800 MHz Response, Large Swing

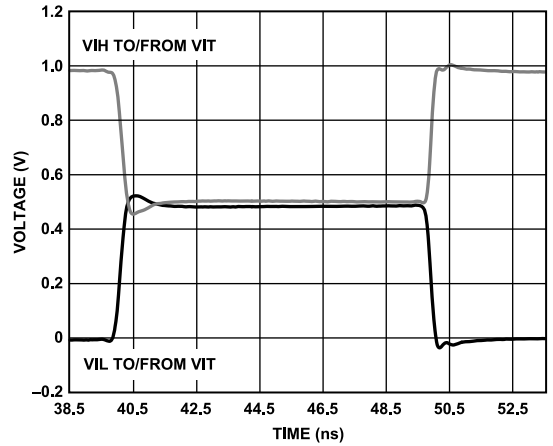


Figure 31. Driver VIL/VIH to/from VIT, VIH = 2.0 V, VIL = 0.0 V, VIT = 1.0 V; 50 Ω Terminated

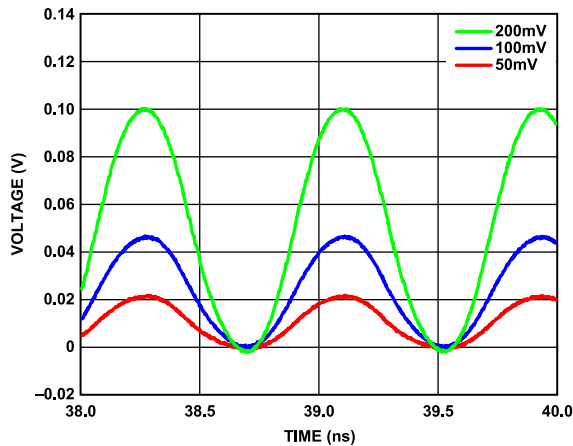


Figure 29. Driver 1.25 GHz Response, Small Swing

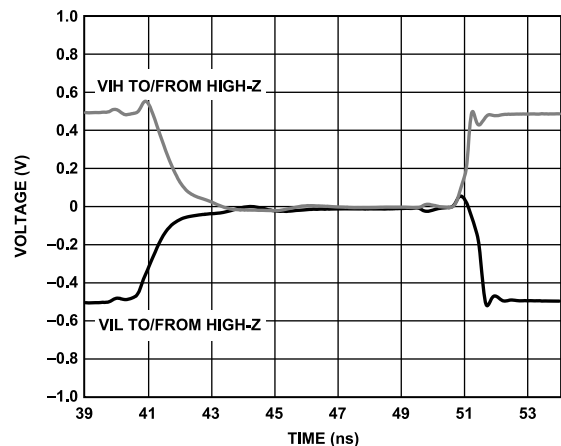


Figure 32. Driver VIL/VIH to/from High-Z, VIH = 1.0 V, VIL = -1.0 V; 50 Ω Terminated

TYPICAL PERFORMANCE CHARACTERISTICS

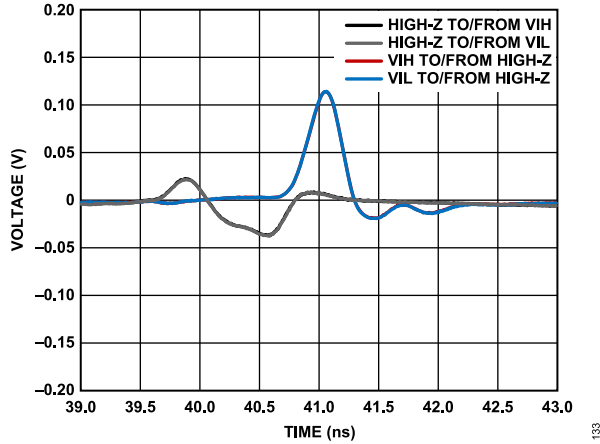


Figure 33. Driver to/from High-Z Transient Spike,  $V_{IH} = V_{IL} = 0.0\text{ V}$ ;  $50\ \Omega$  Terminated

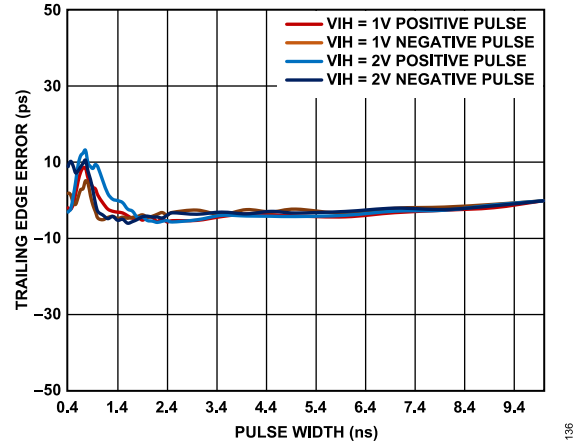


Figure 36. Driver Pulse Width (Positive/Negative) Trailing Edge Timing Error,  $V_{IH} = 1.0\text{ V}, 2.0\text{ V}$ ;  $V_{IL} = 0.0\text{ V}$ ; CLC = Midscale;  $50\ \Omega$  Terminated

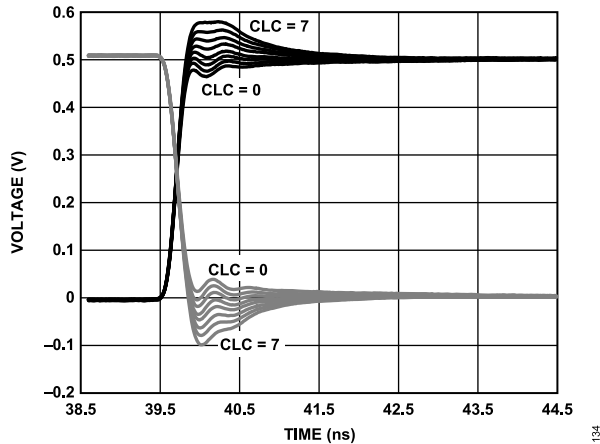


Figure 34. Driver Transition vs. CLC,  $V_{IH} = 1.0\text{ V}$ ,  $V_{IL} = 0.0\text{ V}$ ;  $50\ \Omega$  Terminated

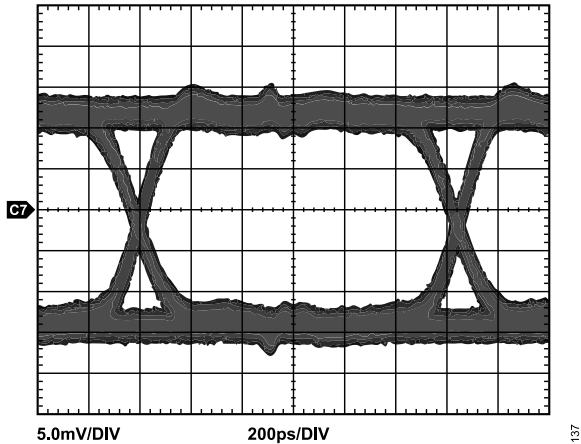


Figure 37. Driver Eye Diagram, 800 Mbps, PRBS31,  $V_{IH} = 50\text{ mV}$ ,  $V_{IL} = 0.0\text{ V}$ ;  $50\ \Omega$  Terminated

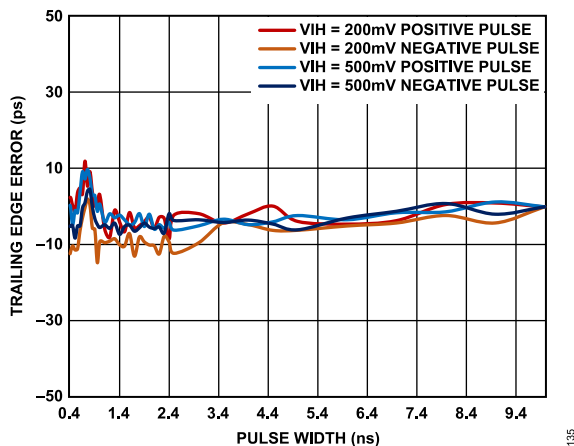


Figure 35. Driver Pulse Width (Positive/Negative) Trailing Edge Timing Error,  $V_{IH} = 0.2\text{ V}, 0.5\text{ V}$ ;  $V_{IL} = 0.0\text{ V}$ ; CLC = Midscale;  $50\ \Omega$  Terminated

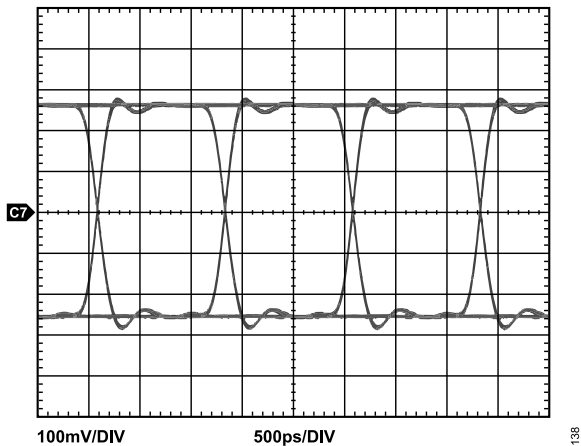


Figure 38. Driver Eye Diagram, 800 Mbps, PRBS31,  $V_{IH} = 1.0\text{ V}$ ,  $V_{IL} = 0.0\text{ V}$ ;  $50\ \Omega$  Terminated

TYPICAL PERFORMANCE CHARACTERISTICS

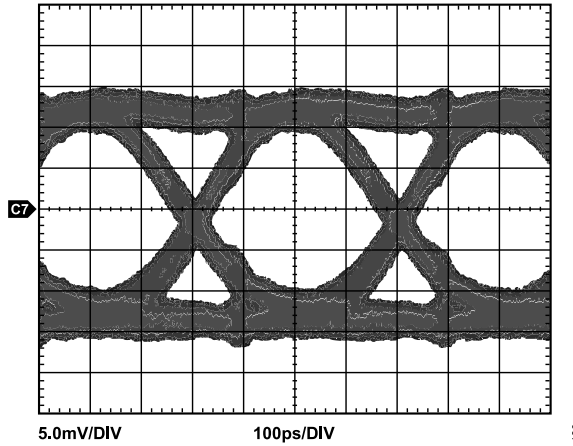


Figure 39. Driver Eye Diagram, 2.5 Gbps, PRBS31,  $V_{IH} = 50\text{ mV}$ ,  $V_{IL} = 0.0\text{ V}$ ;  $50\ \Omega$  Terminated

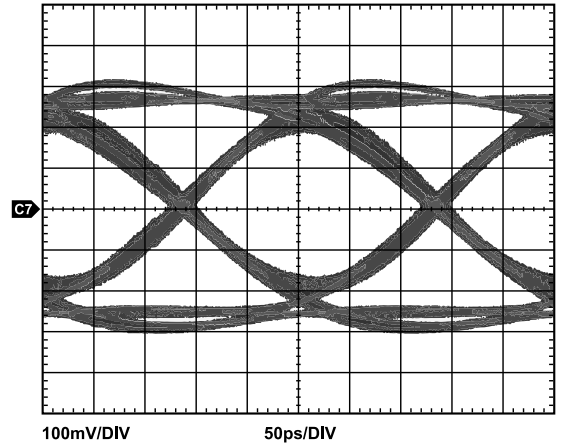


Figure 42. Driver Eye Diagram, 4.0 Gbps, PRBS31,  $V_{IH} = 1.0\text{ V}$ ,  $V_{IL} = 0.0\text{ V}$ ;  $50\ \Omega$  Terminated

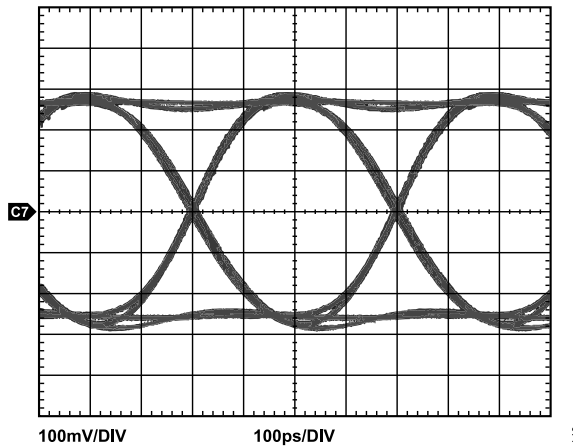


Figure 40. Driver Eye Diagram, 2.5 Gbps, PRBS31,  $V_{IH} = 1.0\text{ V}$ ,  $V_{IL} = 0.0\text{ V}$ ;  $50\ \Omega$  Terminated

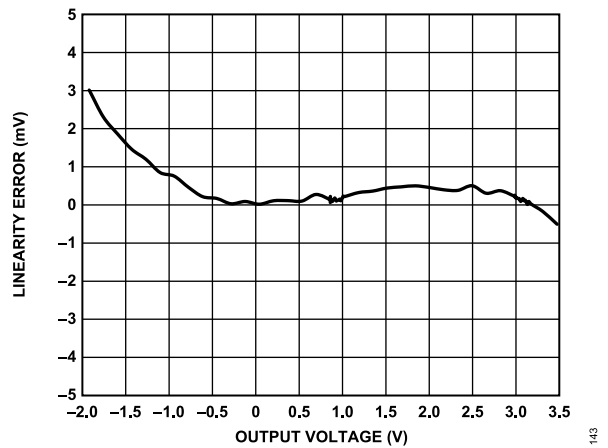


Figure 43. Reflection Clamp VCLx INL

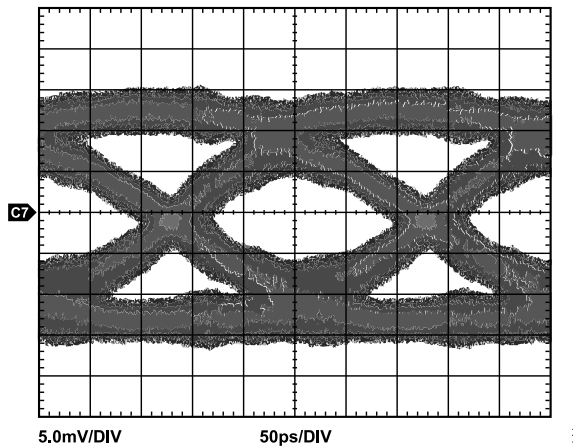


Figure 41. Driver Eye Diagram, 4.0 Gbps, PRBS31,  $V_{IH} = 50\text{ mV}$ ,  $V_{IL} = 0.0\text{ V}$ ;  $50\ \Omega$  Terminated

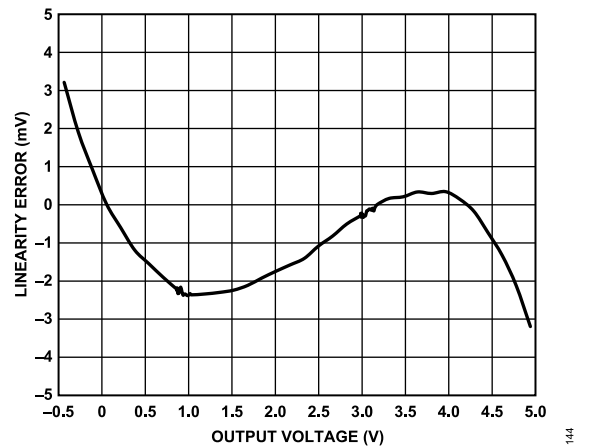


Figure 44. Reflection Clamp VCHx INL

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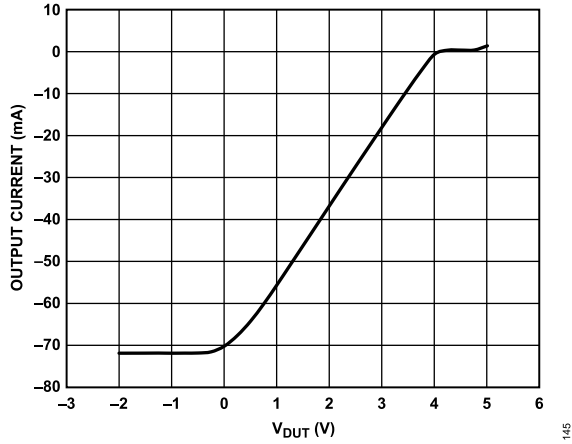


Figure 45. Reflection Clamp Current Limit,  $V_{CHx} = 5.0\text{ V}$ ,  $V_{CLx} = 4.0\text{ V}$ ;  $V_{DUTx}$  Swept from  $-2.0\text{ V}$  to  $+5.0\text{ V}$

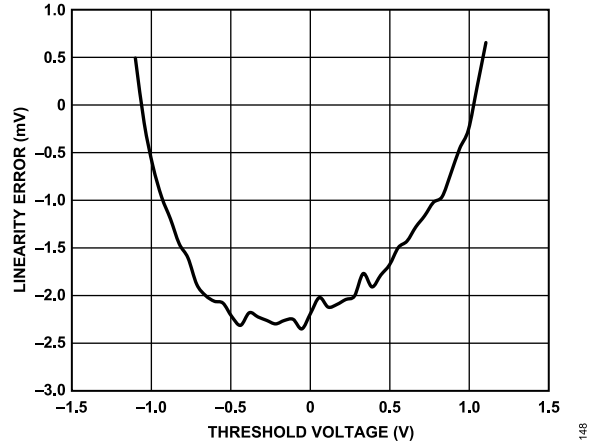


Figure 48. Differential Mode Comparator Threshold INL

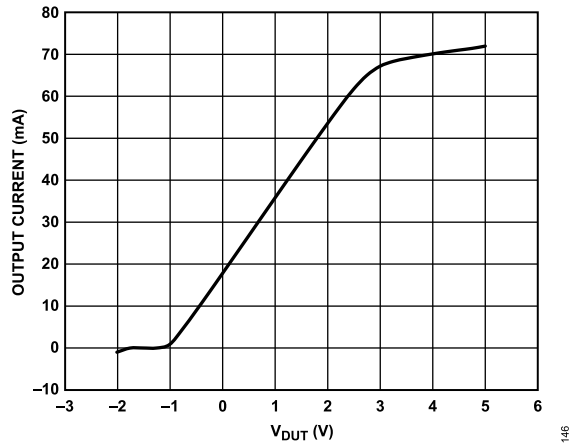


Figure 46. Reflection Clamp Current Limit,  $V_{CHx} = -1.0\text{ V}$ ,  $V_{CLx} = -2.0\text{ V}$ ;  $V_{DUTx}$  Swept from  $-2.0\text{ V}$  to  $+5.0\text{ V}$

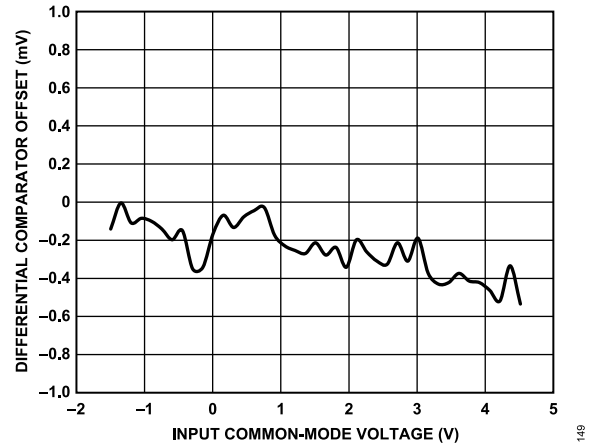


Figure 49. Differential Mode Comparator Common-Mode Rejection Error

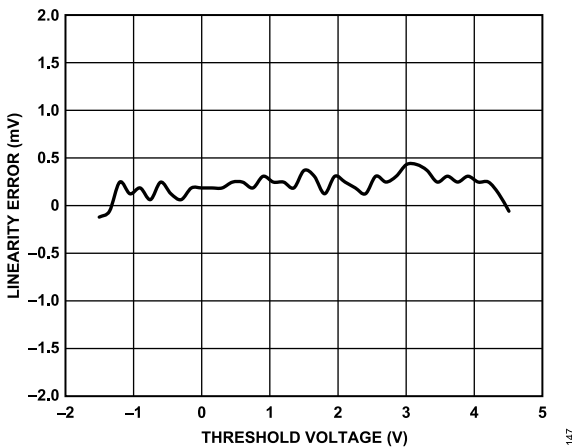


Figure 47. Normal Window Comparator Threshold INL

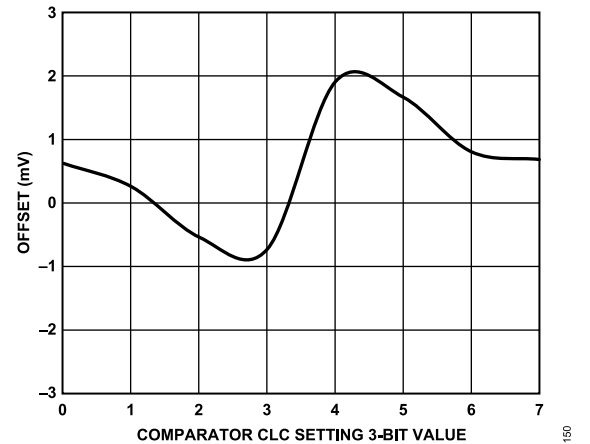


Figure 50. Normal Window Comparator Offset Error vs. CLC Setting

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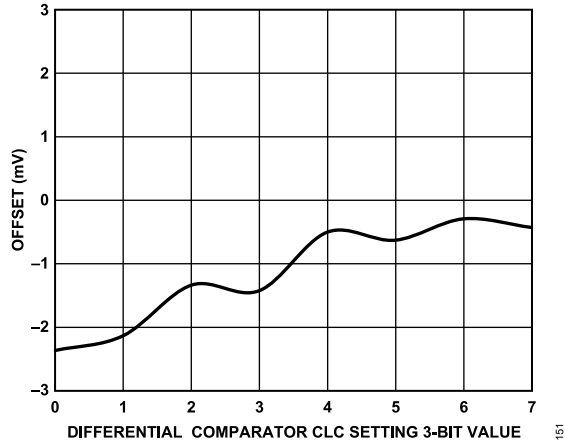


Figure 51. Differential Mode Comparator Offset Error vs. CLC Setting

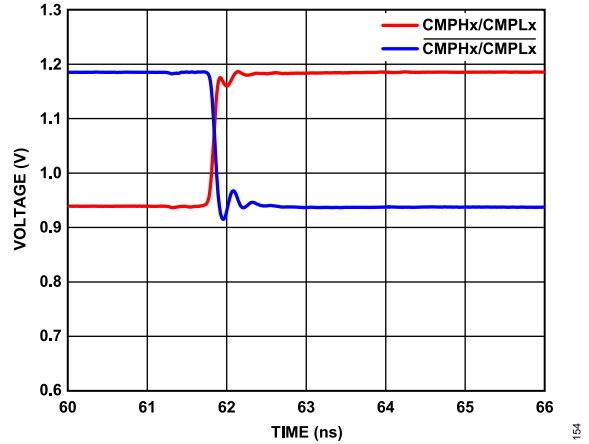


Figure 54. Comparator CML Output Waveform (ADATE320)

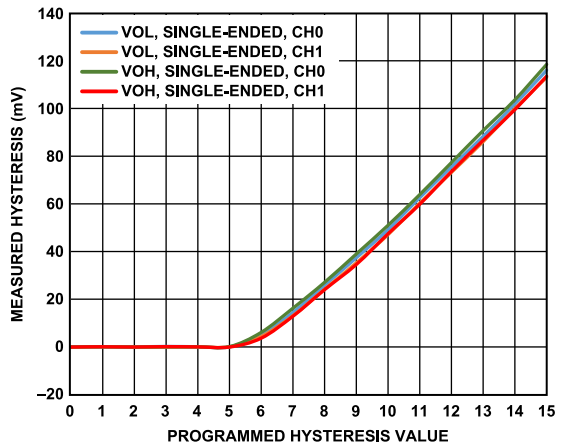


Figure 52. Normal Window Comparator Hysteresis vs. Programmed Hysteresis Value

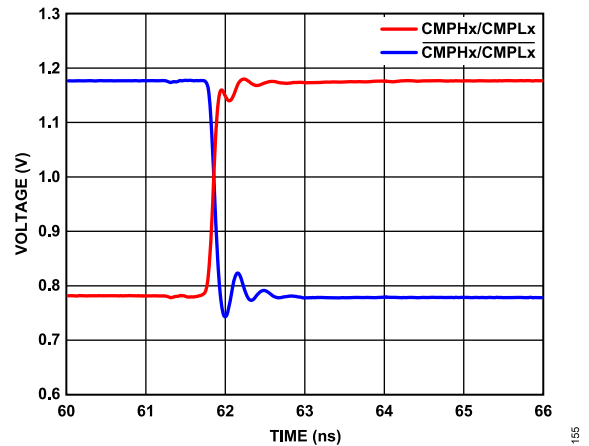


Figure 55. Comparator CML Output Waveform (ADATE320-1)

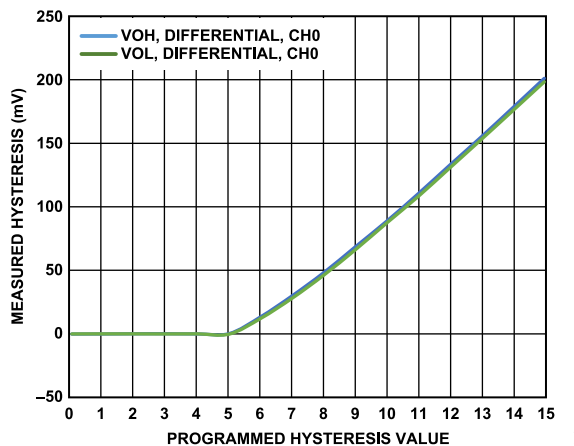


Figure 53. Differential Mode Comparator Hysteresis vs. Programmed Hysteresis Value

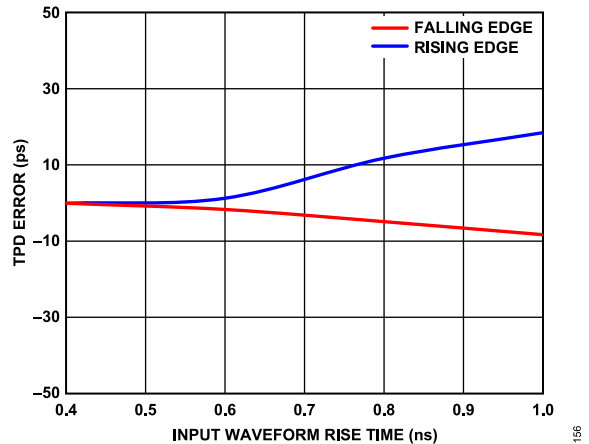


Figure 56. Normal Window Comparator Propagation Delay vs. Input Rise Time, 1.0 V Input Swing

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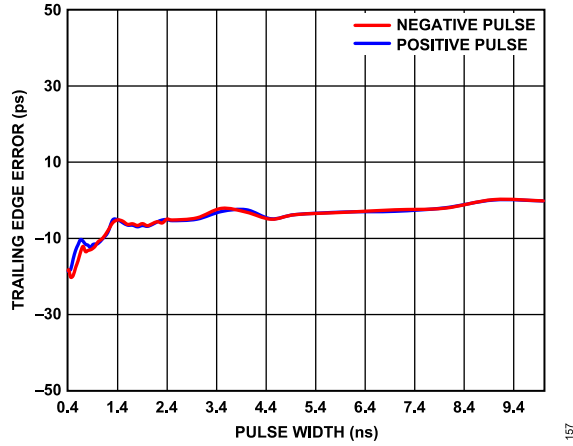


Figure 57. Normal Window Comparator Pulse Width (Positive/Negative) Trailing Edge Timing Error, 1.0 V Input Swing

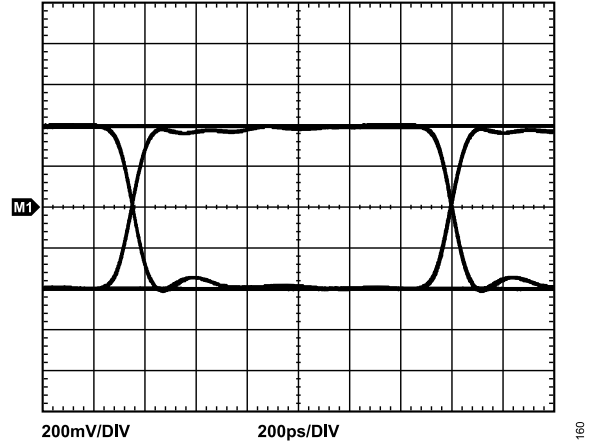


Figure 60. Normal Window Comparator Eye Diagram, 800 Mbps, PRBS31, 1.0 V Input Swing; 50 Ω Terminated

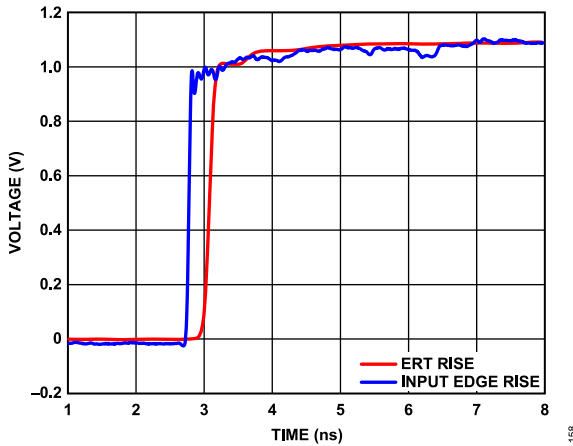


Figure 58. Normal Window Comparator Equivalent Rise Time (ERT), 1.0 V Input Swing, 50 ps 20% to 80%; 50 Ω Terminated

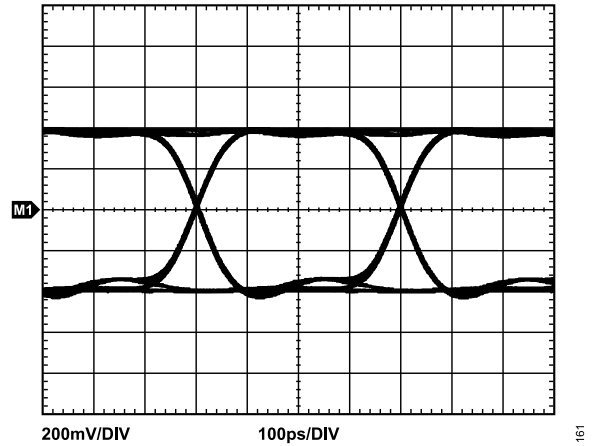


Figure 61. Normal Window Comparator Eye Diagram, 2.5 Gbps, PRBS31, 1.0 V Input Swing; 50 Ω Terminated

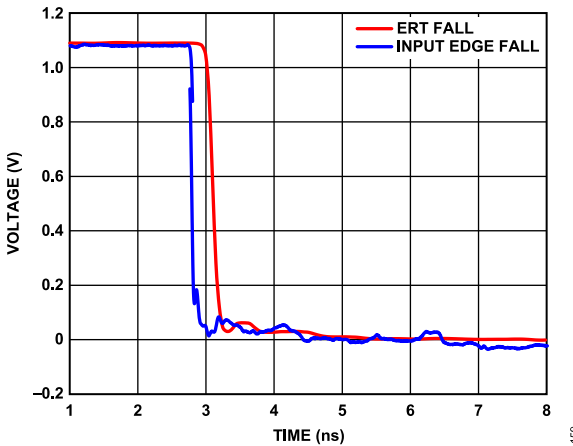


Figure 59. Normal Window Comparator Equivalent Fall Time (EFT), 1.0 V Input Swing, 50 ps 20% to 80%; 50 Ω Terminated

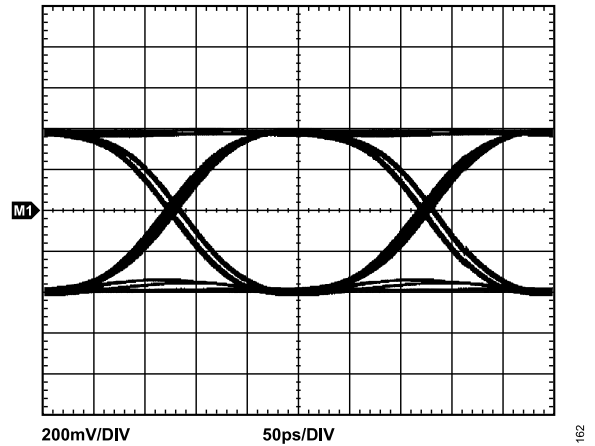


Figure 62. Normal Window Comparator Eye Diagram, 4.0 Gbps, PRBS31, 1.0 V Input Swing; 50 Ω Terminated

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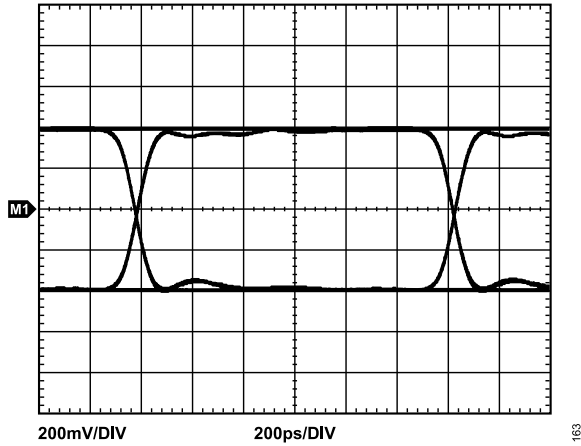


Figure 63. Differential Mode Comparator Eye Diagram, 800 Mbps, PRBS31, 1.0 V Input Swing; 50 Ω Terminated

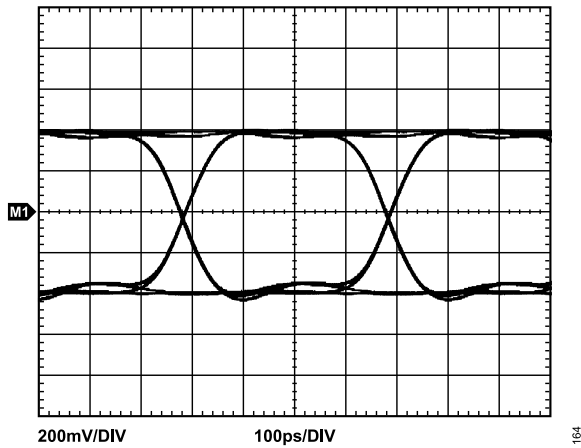


Figure 64. Differential Mode Comparator Eye Diagram, 2.5 Gbps, PRBS31, 1.0 V Input Swing; 50 Ω Terminated

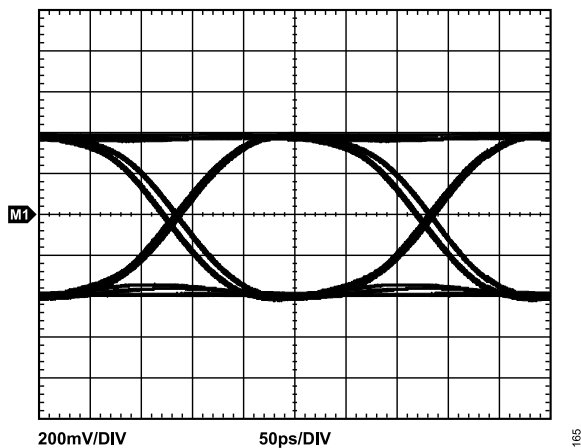


Figure 65. Differential Mode Comparator Eye Diagram, 4.0 Gbps, PRBS31, 1.0 V Input Swing; 50 Ω Terminated

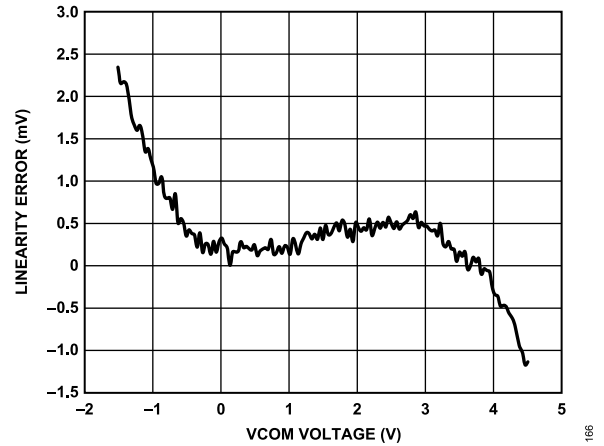


Figure 66. Active Load VCOM INL

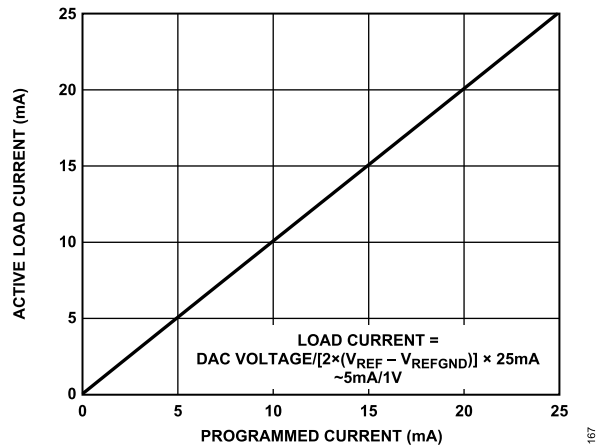


Figure 67. Active Load IOHx/IOLx Transfer Function

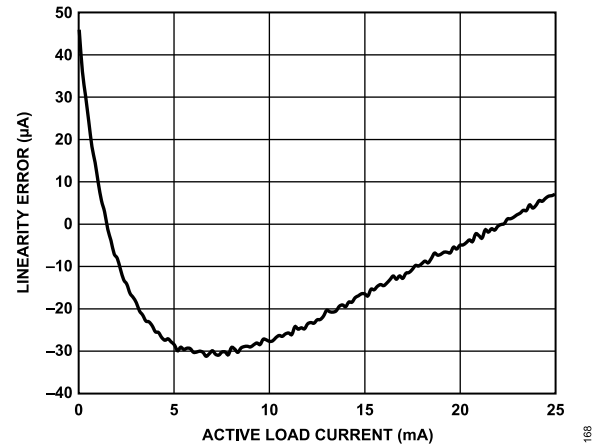


Figure 68. Active Load IOHx INL

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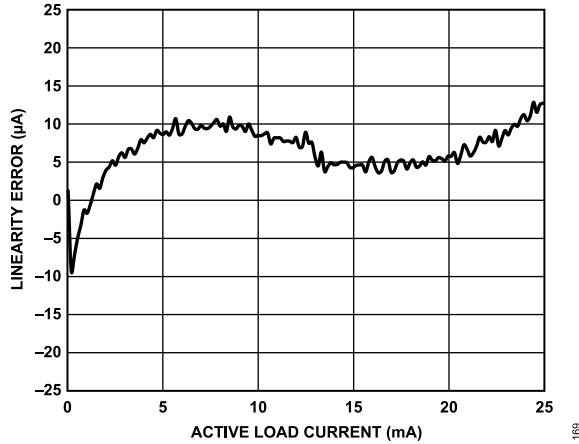


Figure 69. Active Load IOLx INL

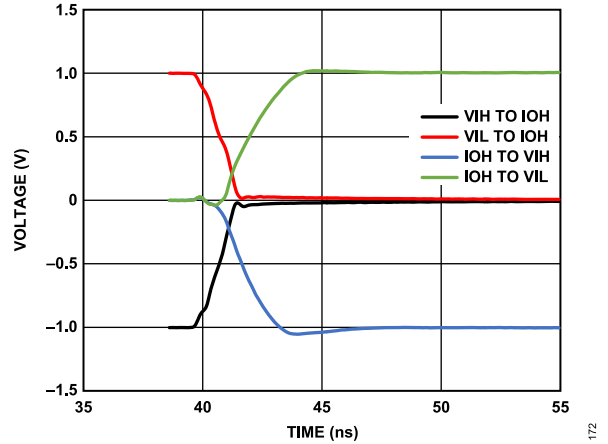


Figure 72. Active Load IOHx to/from Driver Transient Response,  $V_{IH} = V_{IL} = 0.0\text{ V}$ ,  $IOH_x = IOL_x = 20\text{ mA}$ ;  $50\ \Omega$  Terminated

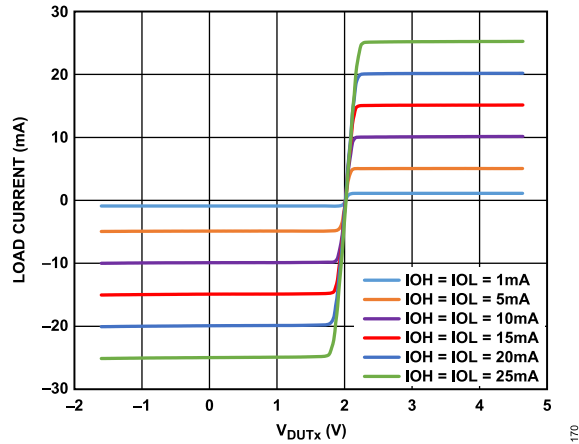


Figure 70. Active Load Commutation Response,  $V_{COM} = 2.0\text{ V}$

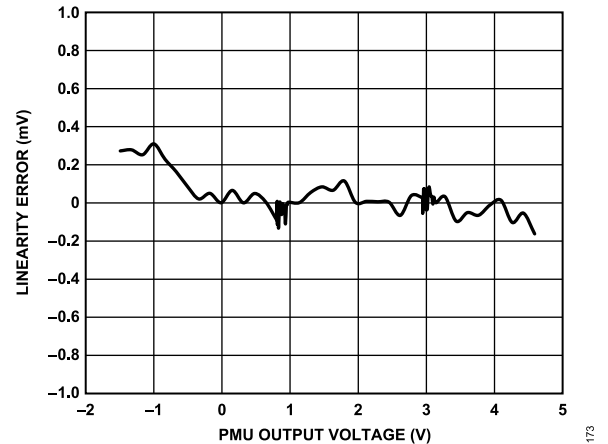


Figure 73. PPMU Force Voltage INL, All Ranges

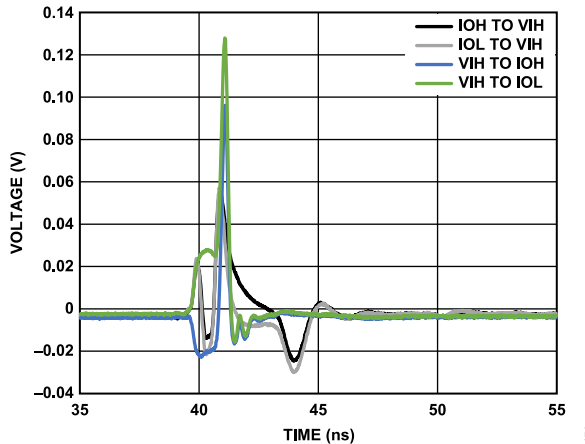


Figure 71. Active Load to/from Driver Input/Output Spike,  $V_{IH} = V_{IL} = 0.0\text{ V}$ ,  $IOH_x = IOL_x = 0.0\text{ mA}$ ;  $50\ \Omega$  Terminated

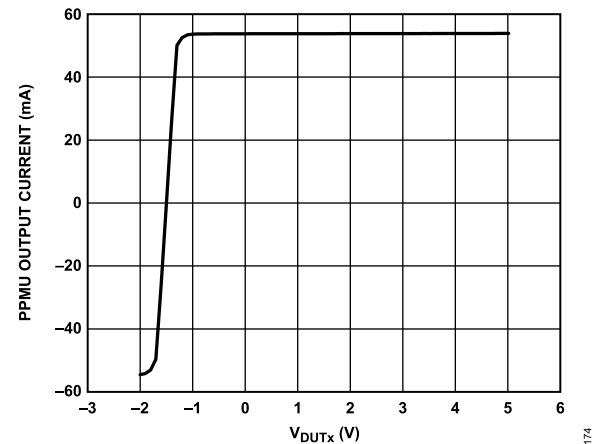


Figure 74. PPMU Force Voltage Output Current Limit, Range A,  $FV = -1.5\text{ V}$ ,  $V_{DUTx}$  Swept  $-2.0\text{ V}$  to  $+5.0\text{ V}$



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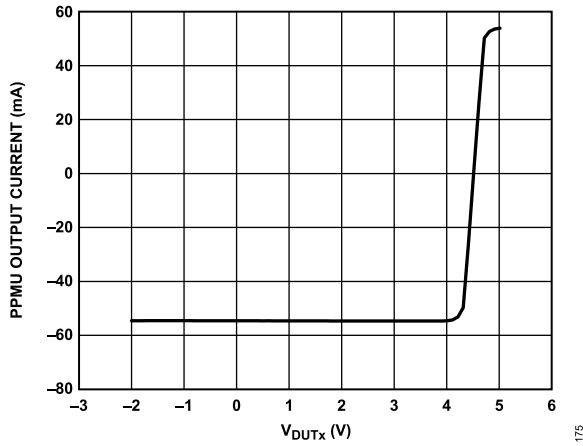


Figure 75. PPMU Force Voltage Output Current Limit, Range A, FV = 4.5 V,  $V_{DUTx}$  Swept -2.0 V to +5.0 V

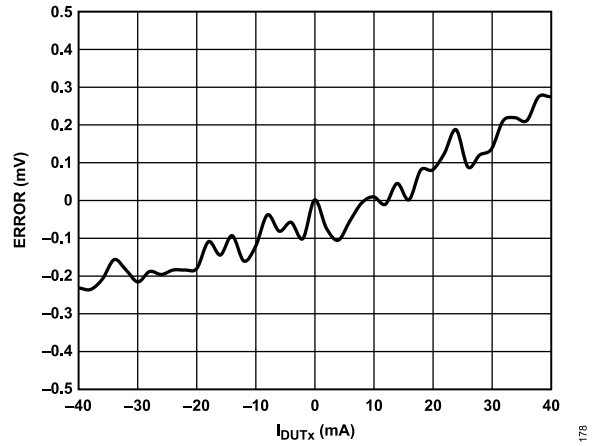


Figure 78. PPMU Force Voltage Compliance Error, Range A, FV = -1.0 V vs. Output Current, Internal Sense

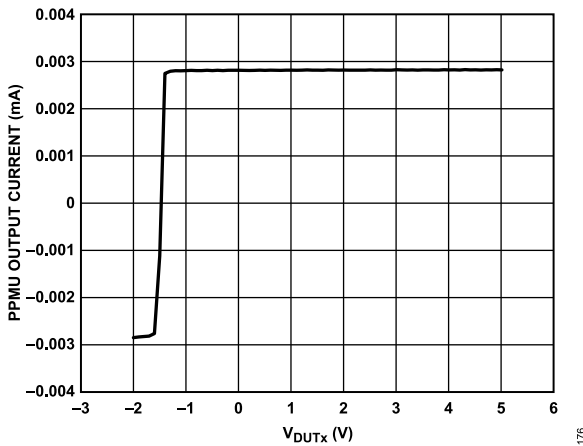


Figure 76. PPMU Force Voltage Output Current Limit, Range E, FV = -1.5 V,  $V_{DUTx}$  Swept -2.0 V to +5.0 V

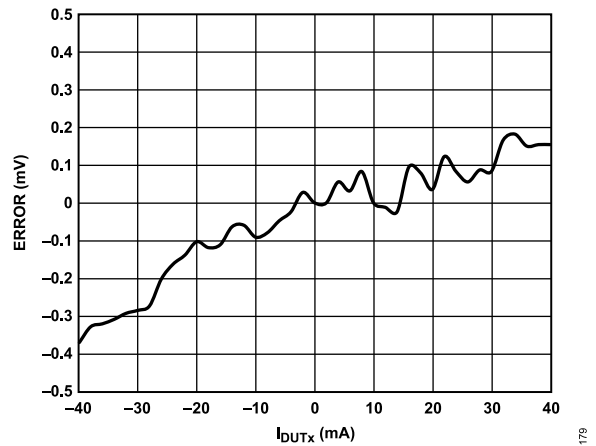


Figure 79. PPMU Force Voltage Compliance Error, Range A, FV = 4.0 V vs. Output Current ( $I_{DUTx}$ ), Internal Sense

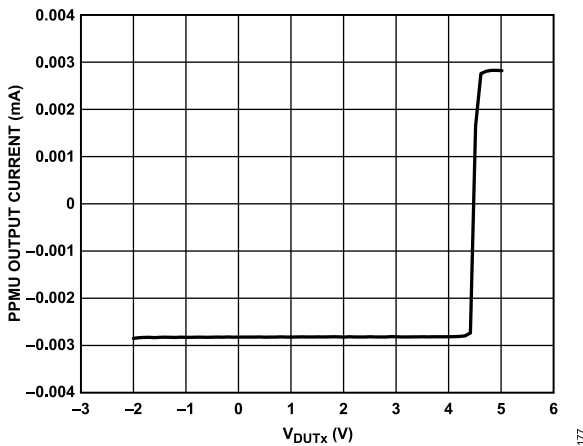


Figure 77. PPMU Force Voltage Output Current Limit, Range E, FV = 4.5 V,  $V_{DUTx}$  Swept -2.0 V to +5.0 V

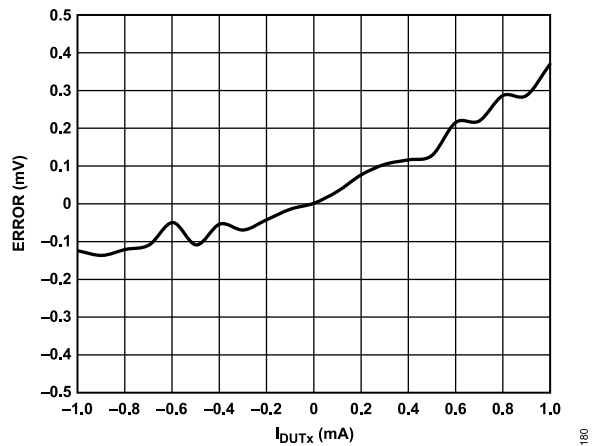


Figure 80. PPMU Force Voltage Compliance Error, Range B, FV = -1.5 V vs. Output Current ( $I_{DUTx}$ ), Internal Sense

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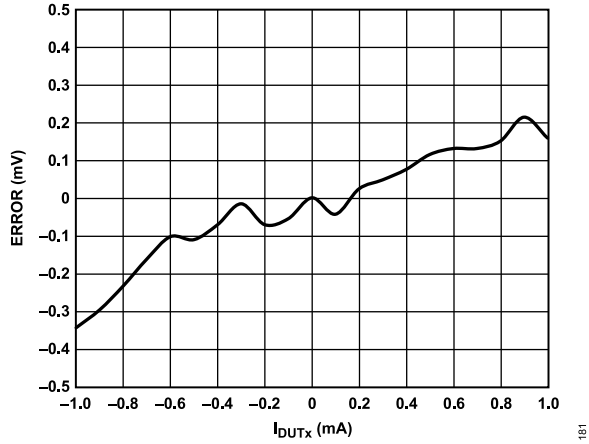


Figure 81. PPMU Force Voltage Compliance Error, Range B, FV = 4.5 V vs. Output Current ( $I_{DUTx}$ ), Internal Sense

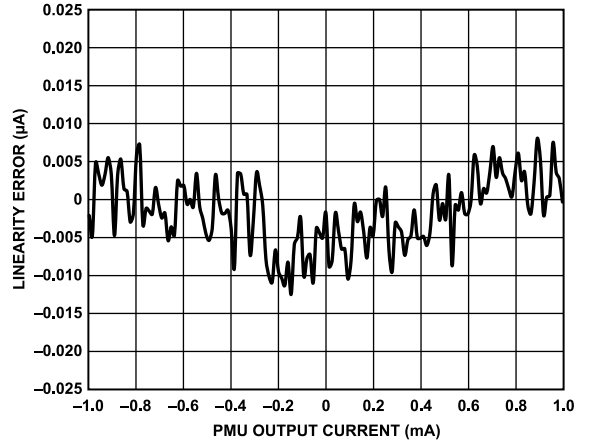


Figure 84. PPMU Force Current INL, Range C

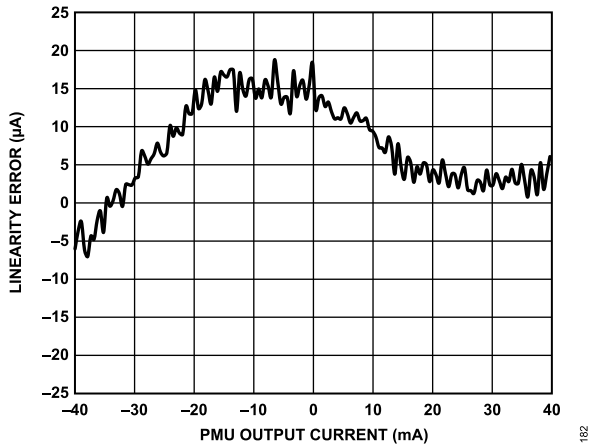


Figure 82. PPMU Force Current INL, Range A

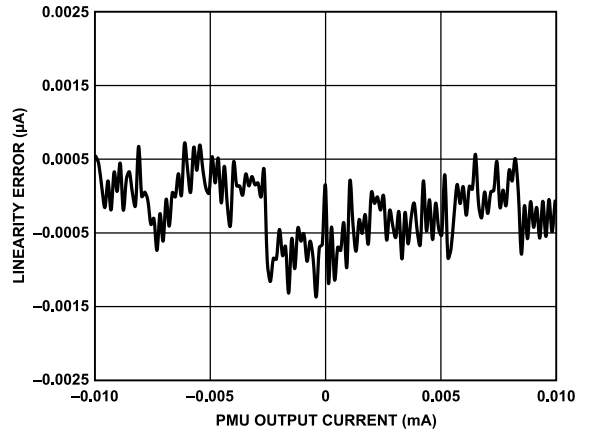


Figure 85. PPMU Force Current INL, Range D

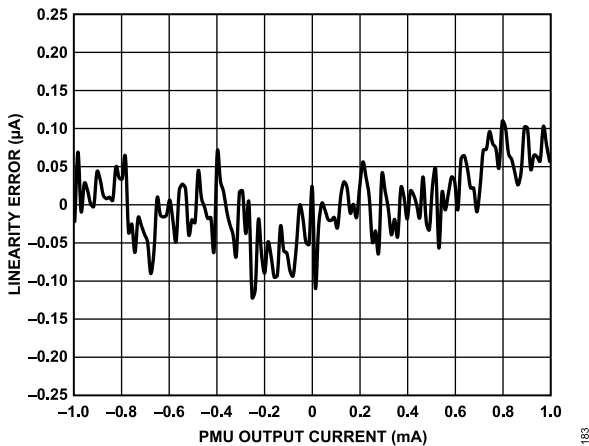


Figure 83. PPMU Force Current INL, Range B

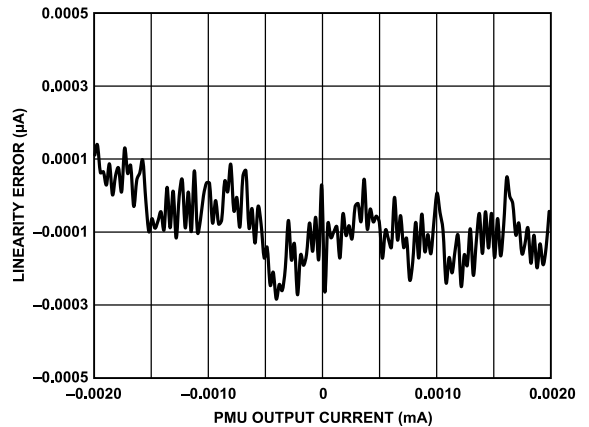


Figure 86. PPMU Force Current INL, Range E

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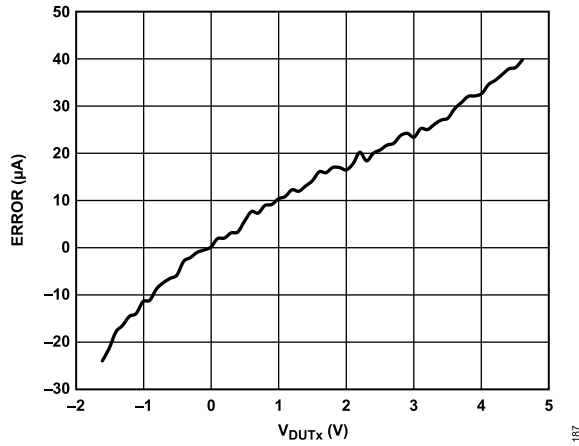


Figure 87. PPMU Force Current Compliance Error, Range A, FI = -40 mA vs. Output Voltage ( $V_{DUTx}$ )

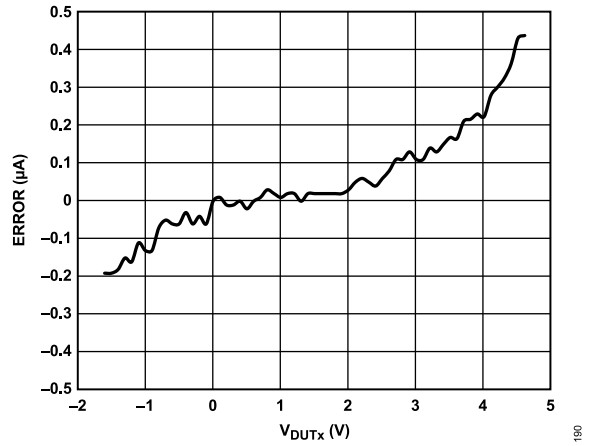


Figure 90. PPMU Force Current Compliance Error, Range B, FI = 1 mA vs. Output Voltage ( $V_{DUTx}$ )

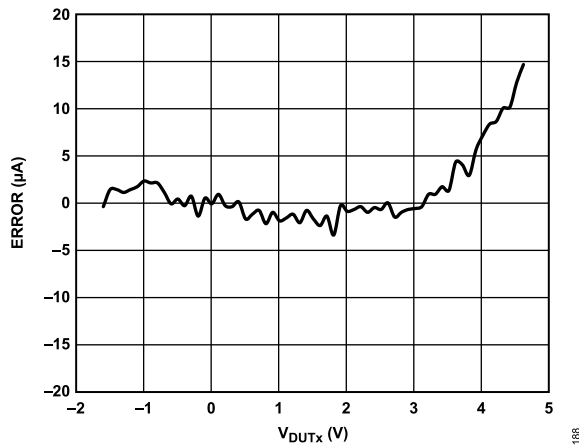


Figure 88. PPMU Force Current Compliance Error, Range A, FI = 40 mA vs. Output Voltage ( $V_{DUTx}$ )

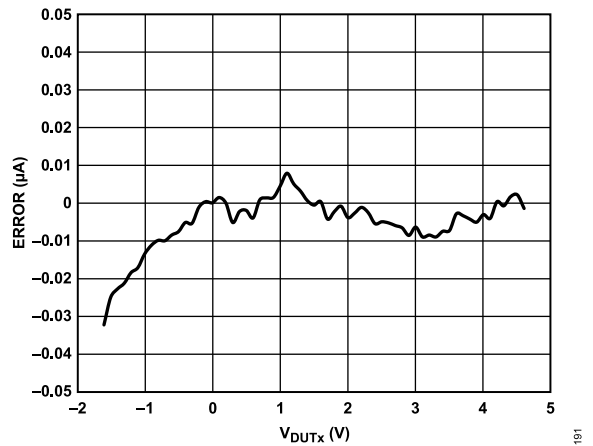


Figure 91. PPMU Force Current Compliance Error, Range C, FI = -100 µA vs. Output Voltage ( $V_{DUTx}$ )

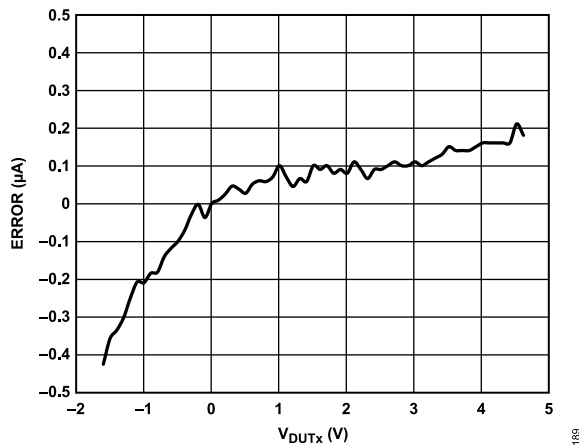


Figure 89. PPMU Force Current Compliance Error, Range B, FI = -1 mA vs. Output Voltage ( $V_{DUTx}$ )

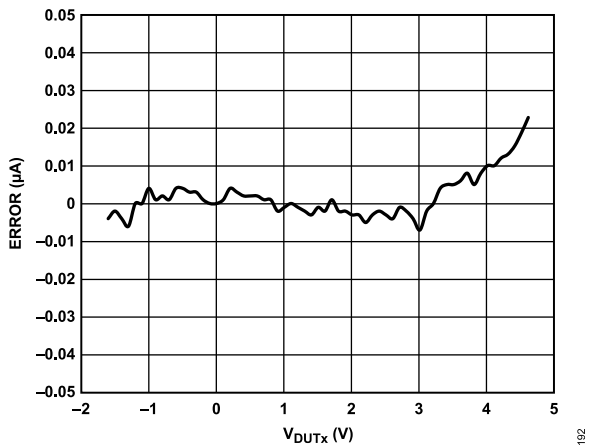


Figure 92. PPMU Force Current Compliance Error, Range C, FI = 100 µA vs. Output Voltage ( $V_{DUTx}$ )

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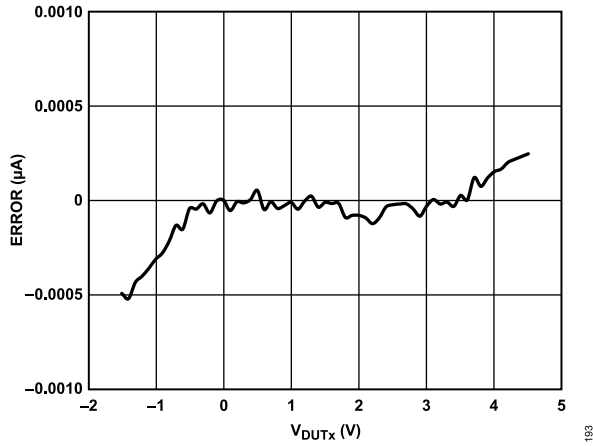


Figure 93. PPMU Force Current Compliance Error, Range E, FI = -2  $\mu$ A vs. Output Voltage ( $V_{DUTx}$ )

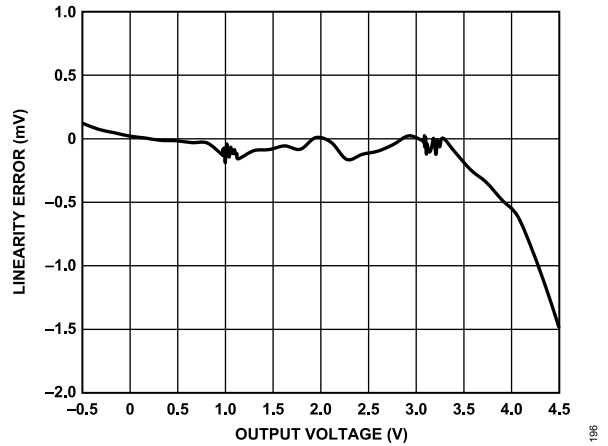


Figure 96. PPMU Voltage Clamp PCHx INL

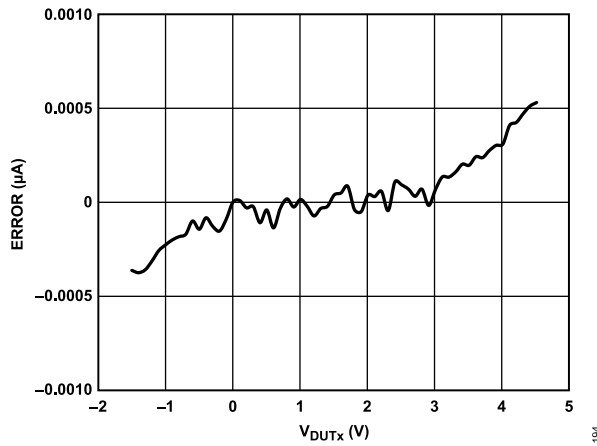


Figure 94. PPMU Force Current Compliance Error, Range E, FI = 2  $\mu$ A vs. Output Voltage ( $V_{DUTx}$ )

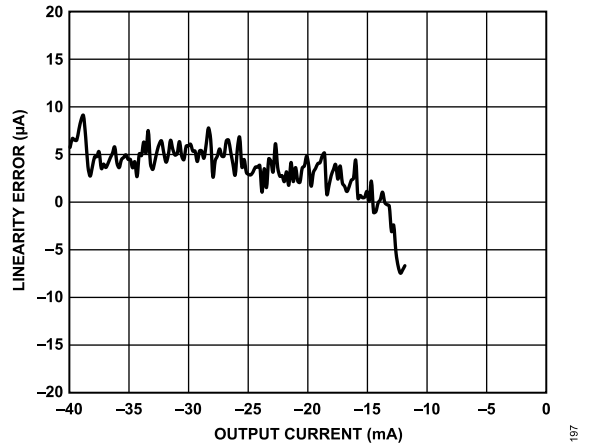


Figure 97. PPMU Current Clamp PCLx INL, Range A

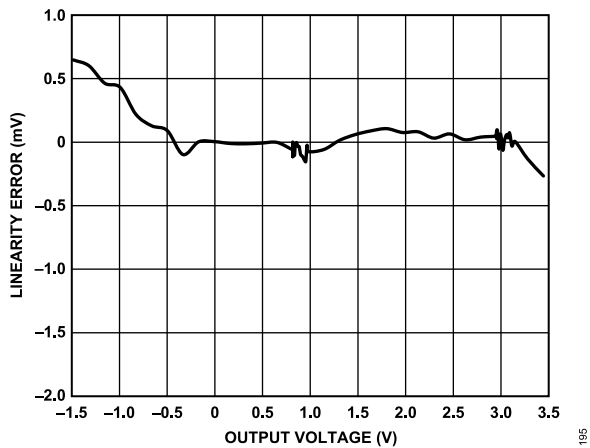


Figure 95. PPMU Voltage Clamp PCLx INL

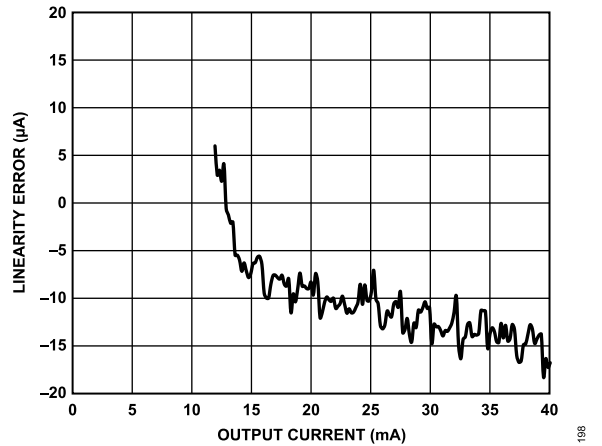


Figure 98. PPMU Current Clamp PCHx INL, Range A

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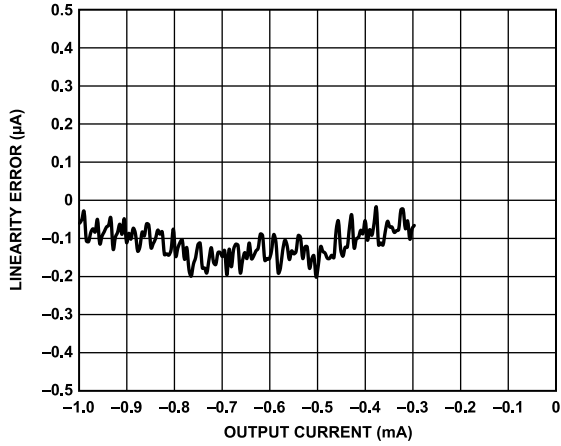


Figure 99. PPMU Current Clamp PCLx INL, Range B

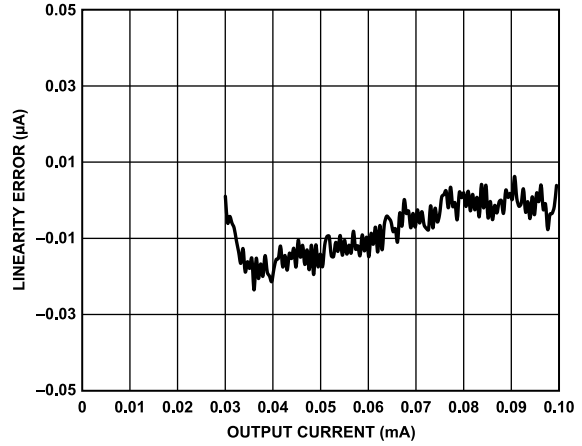


Figure 102. PPMU Current Clamp PCHx INL, Range C

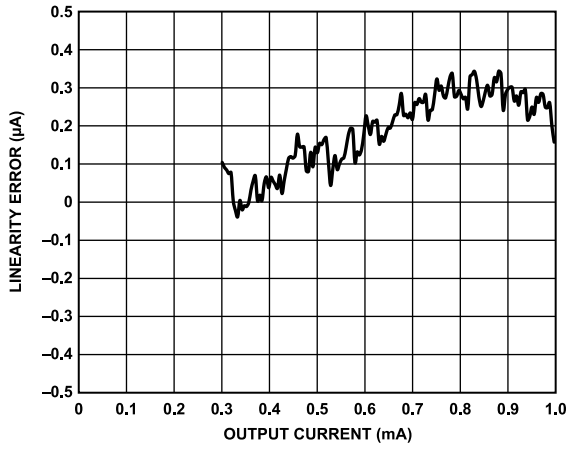


Figure 100. PPMU Current Clamp PCHx INL, Range B

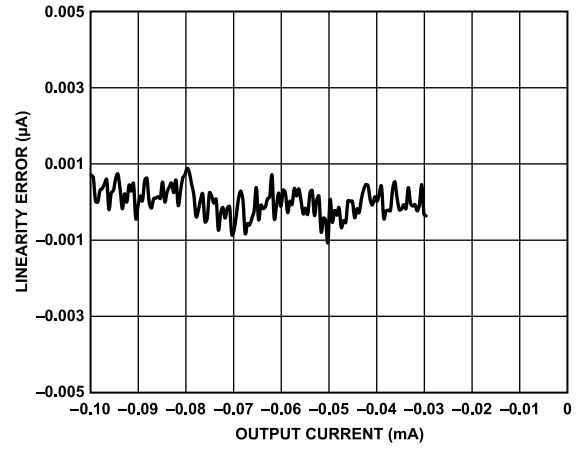


Figure 103. PPMU Current Clamp PCLx INL, Range D

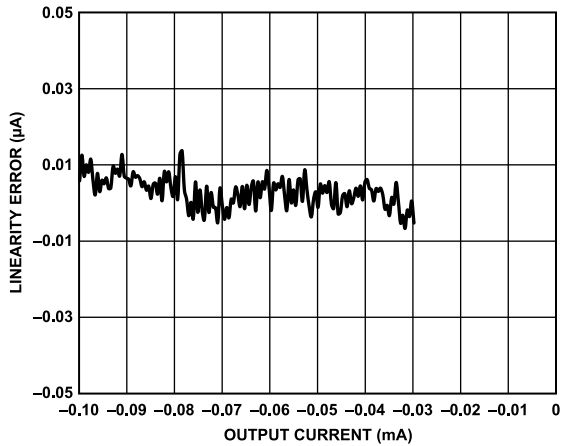


Figure 101. PPMU Current Clamp PCLx INL, Range C

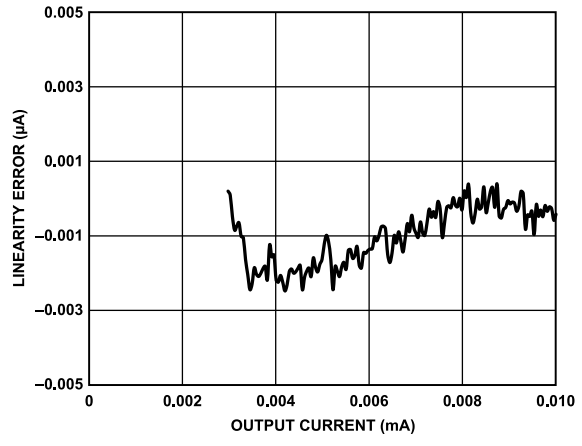


Figure 104. PPMU Current Clamp PCHx INL, Range D

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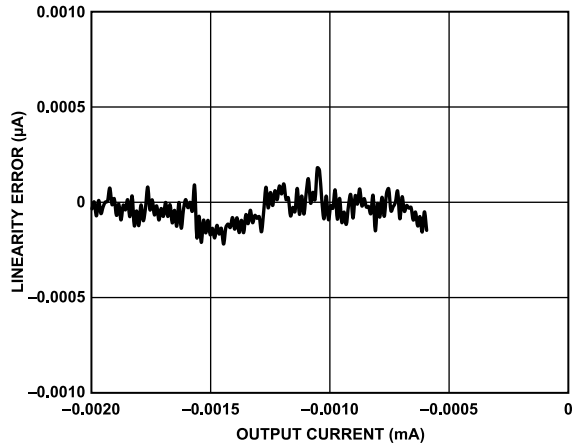


Figure 105. PPMU Current Clamp PCLx INL, Range E

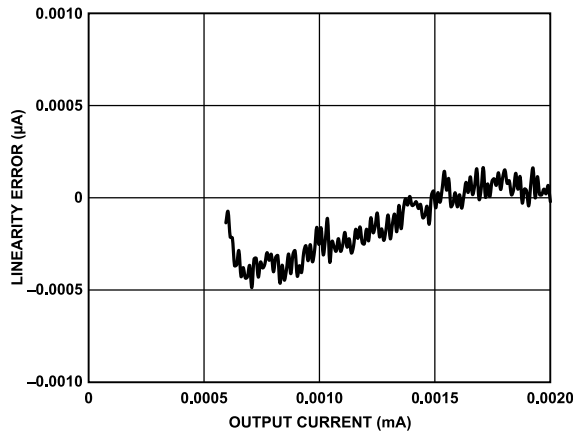


Figure 106. PPMU Current Clamp PCHx INL, Range E

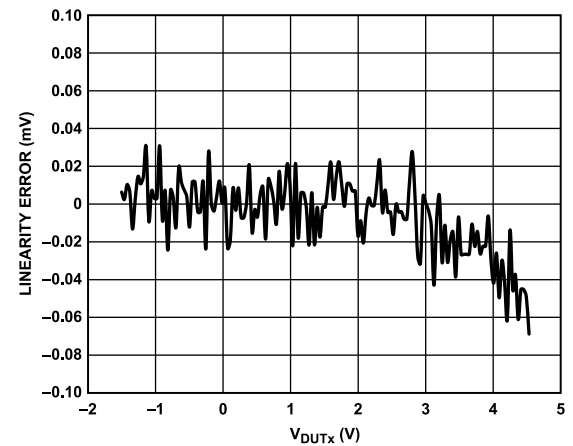


Figure 107. PPMU Measure Voltage INL, Range B

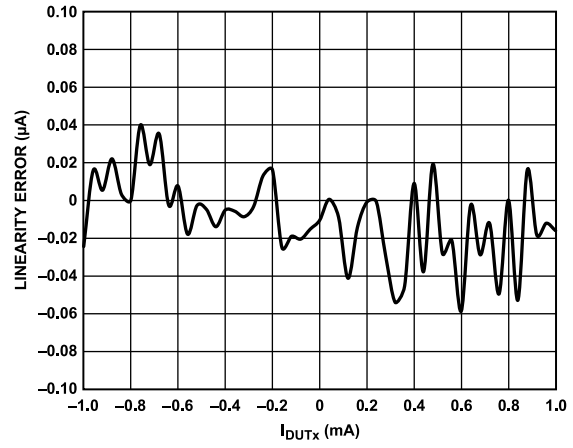


Figure 108. PPMU Measure Current INL, Range B

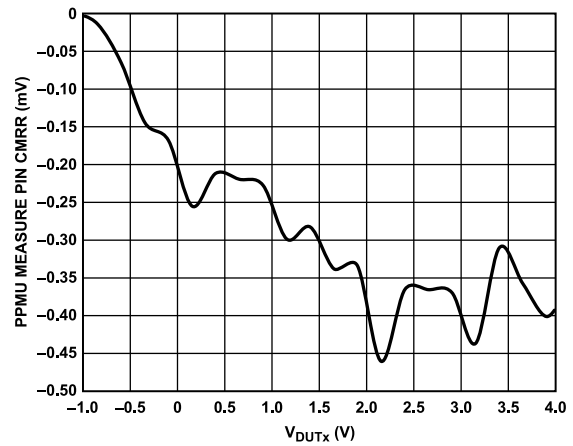


Figure 109. PPMU Measure Current Common-Mode Rejection Error, Force Voltage Measure Current (FVMI), Source 0.5 mA

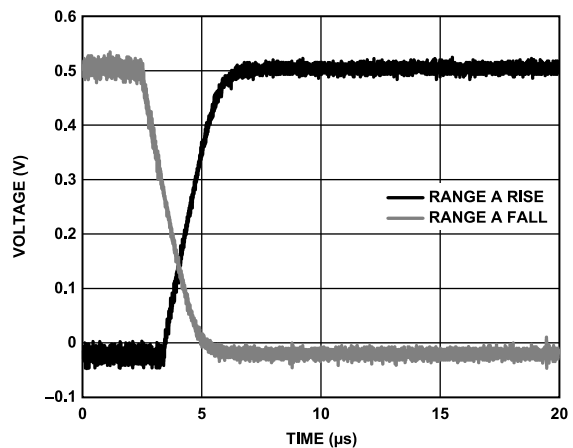


Figure 110. PPMU Force Voltage Transient Response, Range A, 0.0 V to 0.5 V, Uncalibrated,  $C_{LOAD} = 200 \text{ pF}$

TYPICAL PERFORMANCE CHARACTERISTICS

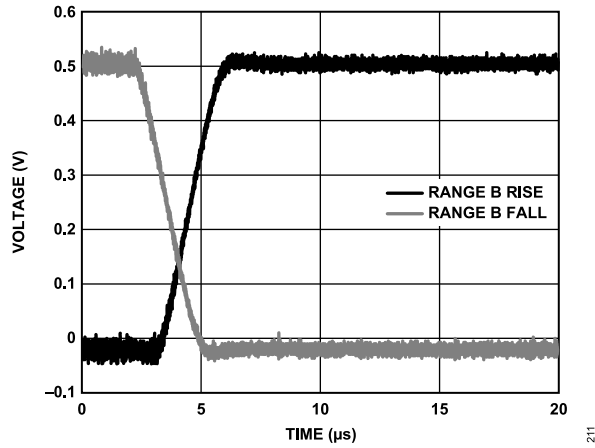


Figure 111. PPMU Force Voltage Transient Response, Range B, 0.0 V to 0.5 V, Uncalibrated,  $C_{LOAD} = 200 \text{ pF}$

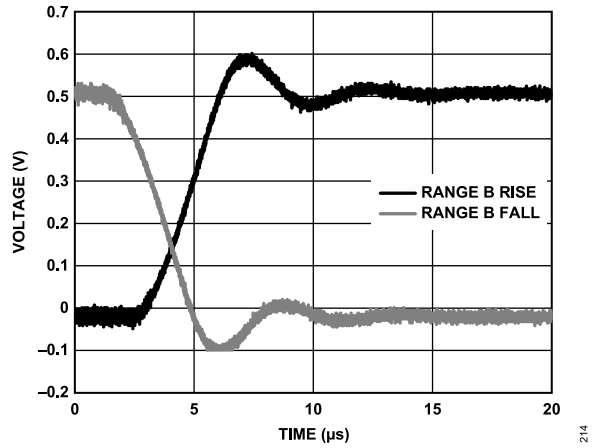


Figure 114. PPMU Force Voltage Transient Response, Range B, 0.0 V to 0.5 V, Uncalibrated,  $C_{LOAD} = 2000 \text{ pF}$

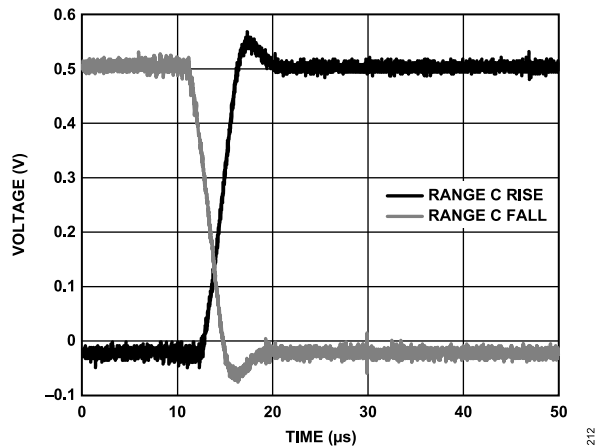


Figure 112. PPMU Force Voltage Transient Response, Range C, 0.0 V to 0.5 V, Uncalibrated,  $C_{LOAD} = 200 \text{ pF}$

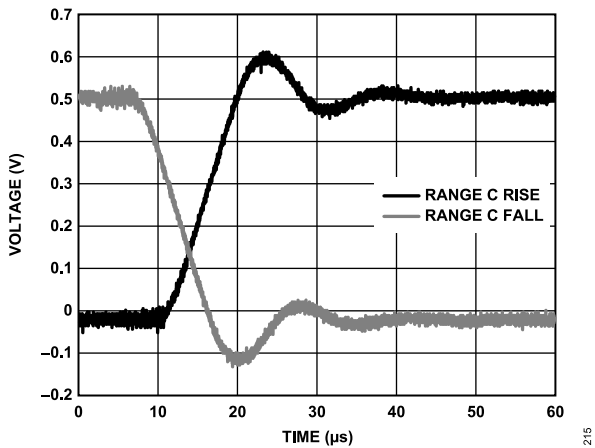


Figure 115. PPMU Force Voltage Transient Response, Range C, 0.0 V to 0.5 V, Uncalibrated,  $C_{LOAD} = 2000 \text{ pF}$

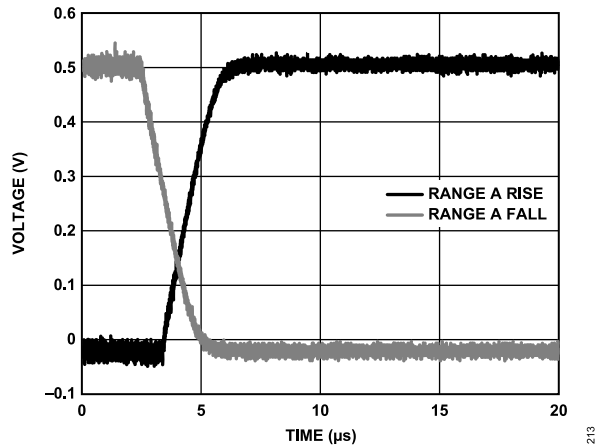


Figure 113. PPMU Force Voltage Transient Response, Range A, 0.0 V to 0.5 V, Uncalibrated,  $C_{LOAD} = 2000 \text{ pF}$

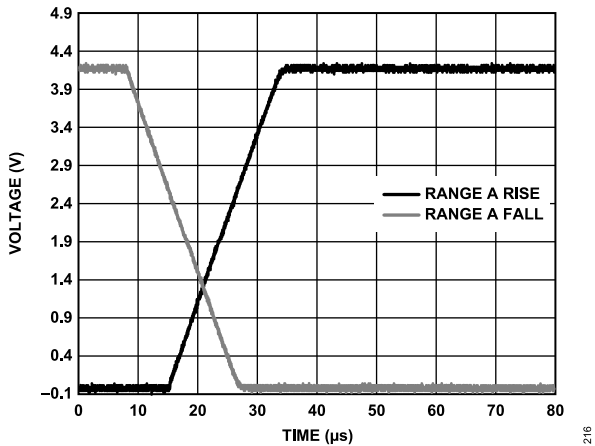


Figure 116. PPMU Force Voltage Transient Response, Range A, 0.0 V to 4.0 V, Uncalibrated,  $C_{LOAD} = 200 \text{ pF}$

TYPICAL PERFORMANCE CHARACTERISTICS

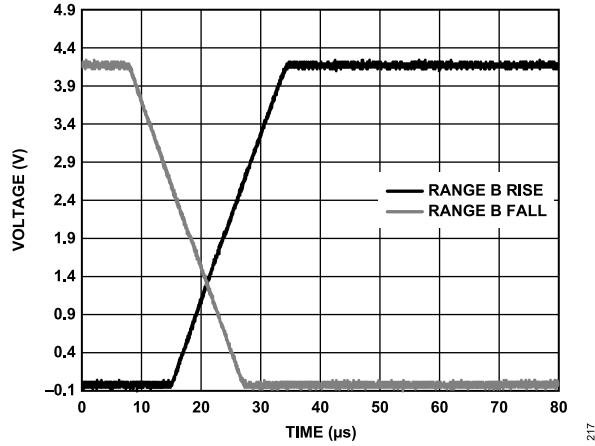


Figure 117. PPMU Force Voltage Transient Response, Range B, 0.0 V to 4.0 V, Uncalibrated,  $C_{LOAD} = 200 \text{ pF}$

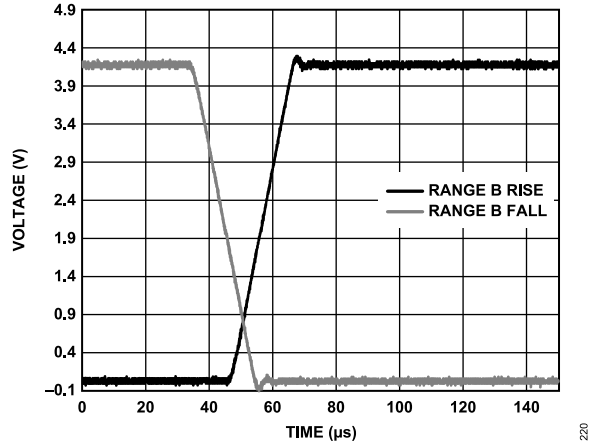


Figure 120. PPMU Force Voltage Transient Response, Range B, 0.0 V to 4.0 V, Uncalibrated,  $C_{LOAD} = 2000 \text{ pF}$

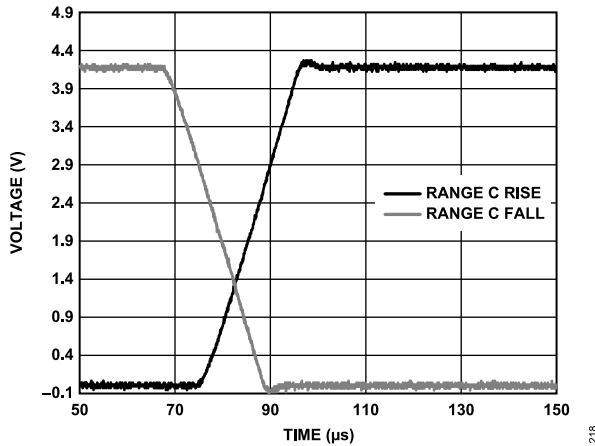


Figure 118. PPMU Force Voltage Transient Response, Range C, 0.0 V to 4.0 V, Uncalibrated,  $C_{LOAD} = 200 \text{ pF}$

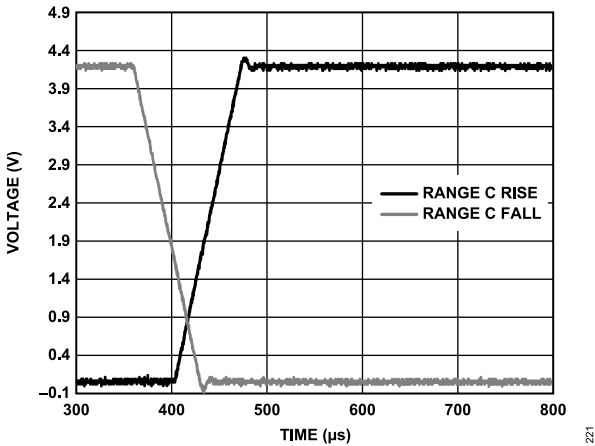


Figure 121. PPMU Force Voltage Transient Response, Range C, 0.0 V to 4.0 V, Uncalibrated,  $C_{LOAD} = 2000 \text{ pF}$

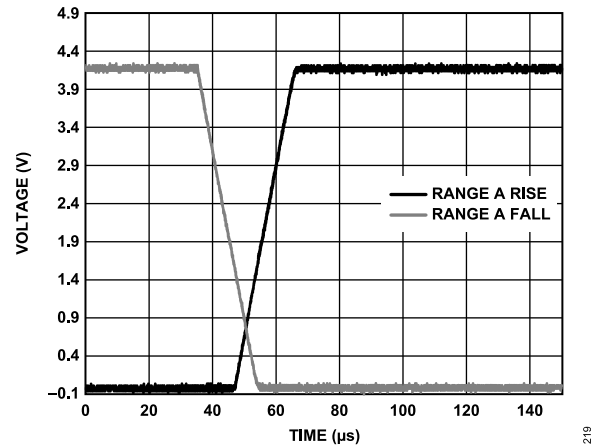


Figure 119. PPMU Force Voltage Transient Response, Range A, 0.0 V to 4.0 V, Uncalibrated,  $C_{LOAD} = 2000 \text{ pF}$

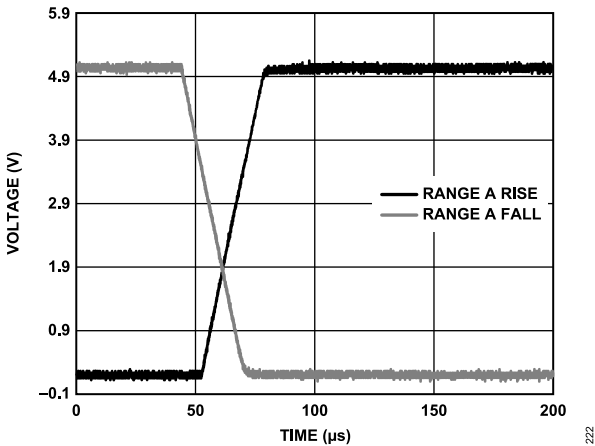


Figure 122. PPMU Force Current Transient Response, Range A, Full-Scale Transition, Uncalibrated,  $C_{LOAD} = 200 \text{ pF}$ ,  $R_{LOAD} = 127 \text{ } \Omega$



TYPICAL PERFORMANCE CHARACTERISTICS

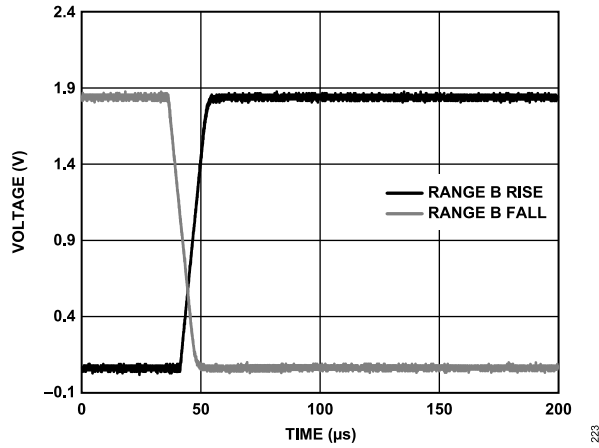


Figure 123. PPMU Force Current Transient Response, Range B, Full-Scale Transition, Uncalibrated,  $C_{LOAD} = 200 \text{ pF}$ ,  $R_{LOAD} = 1.8 \text{ k}\Omega$

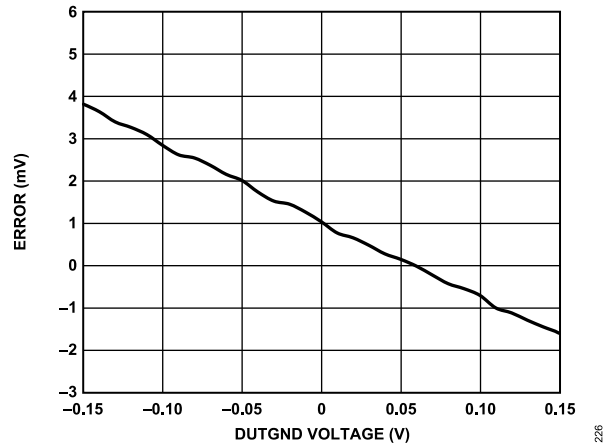


Figure 126. Typical DUTGND Transfer Function Voltage Error, Drive Low,  $V_{IL} = 0.0 \text{ V}$

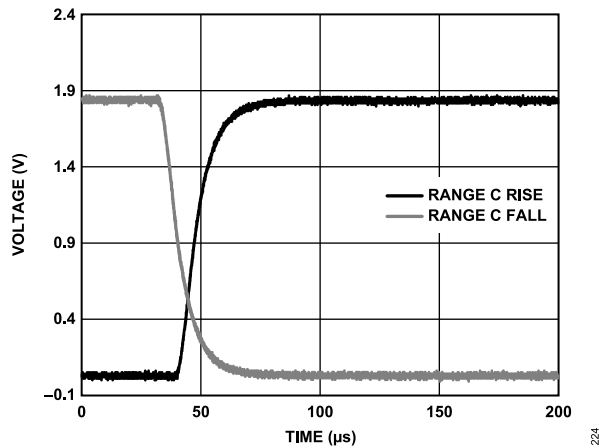


Figure 124. PPMU Force Current Transient Response, Range C, Full-Scale Transition, Uncalibrated,  $C_{LOAD} = 200 \text{ pF}$ ,  $R_{LOAD} = 18.5 \text{ k}\Omega$

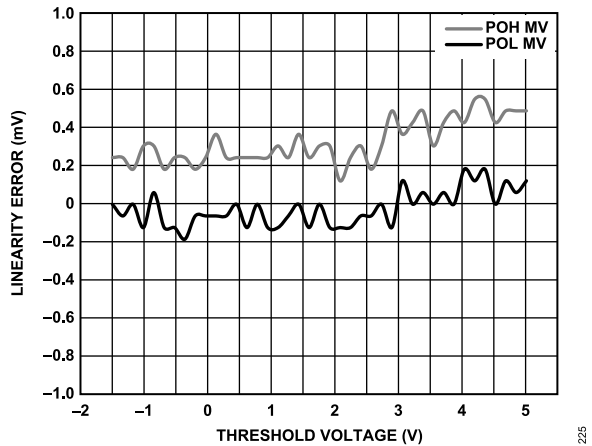


Figure 125. PPMU Go/No-Go Comparator Threshold INL

## THEORY OF OPERATION

## SERIAL PROGRAMMABLE INTERFACE (SPI)

## SPI Hardware Interconnect Details

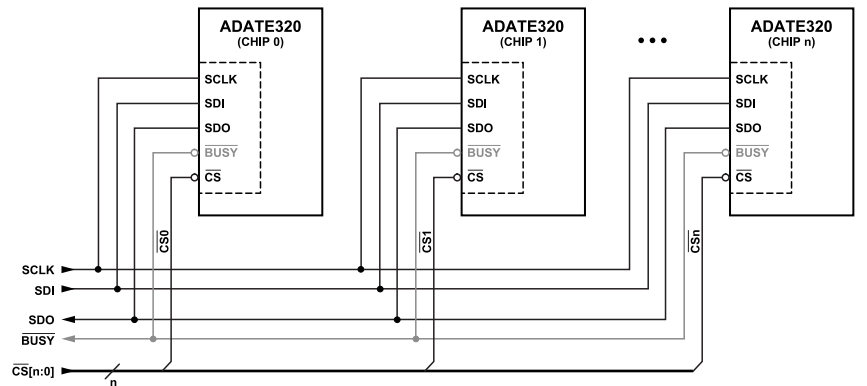


Figure 127. Multiple SPI with a Shared SDO Line

SPI Reset Sequence and the  $\overline{\text{RST}}$  Pin

The internal state of the ADATE320 is indeterminate following power-up. For this reason, it is necessary to perform a valid hardware reset sequence as soon as the power supplies are stabilized. The ADATE320 provides an active low reset pin ( $\overline{\text{RST}}$ ) for this purpose. Asserting  $\overline{\text{RST}}$  asynchronously initiates a reset sequence. Furthermore, the  $\overline{\text{RST}}$  pin must be asserted before and during the power-up cycling sequence, and released only after all power supplies are guaranteed to be stable.

A soft reset sequence can also be initiated under SPI software control by writing to the SPI\_RESET bit (see Figure 146). In the case of a soft reset, the sequence begins on the first rising edge of SCLK following the release of  $\overline{\text{CS}}$ , subject to the normal setup and hold times. Certain actions occur immediately upon the initiation of the reset request, whereas other actions require several cycles of SCLK.

The following asynchronous actions occur immediately following the detection of the reset request, whether it was hardware ( $\overline{\text{RST}}$ ) or software (SPI) initiated:

- ▶ Assert open-drain  $\overline{\text{BUSY}}$  pin
- ▶ Force all control registers to their default reset states as defined in Table 29
- ▶ Clear all calibration registers to their default reset states as defined in Table 29
- ▶ Override all DAC analog outputs and force dc levels to  $V_{\text{DUTGND}}$ , disable the driver and PPMU functions
- ▶ Enable active loads with  $\text{IOH}_x = \text{IOL}_x = 100 \mu\text{A}$  (uncalibrated and expected to vary with offset from device to device); soft connect DUT $_x$  pins to  $\text{VCOM} = V_{\text{DUTGND}}$

The device remains in this static reset state indefinitely until the clocked portion of the sequence begins with either the first rising edge of SCLK following the release of  $\overline{\text{RST}}$  in the case of an asynchronous hardware reset, or the second rising edge of SCLK

following the release of  $\overline{\text{CS}}$  in the case of a software SPI reset. Regardless of how the reset sequence was initiated, the clocked portion of the sequence requires 744 SCLK cycles to run through to completion, and the open-drain  $\overline{\text{BUSY}}$  pin (if available) remains asserted until all clock cycles are received. The following actions occur during the clocked portion of the reset sequence:

- ▶ Complete initialization of internal SPI controller
- ▶ Write default values to appropriate DAC  $X_2$  registers
- ▶ Enable the thermal alarm with a 100°C threshold
- ▶ Disable the PPMU clamp and overvoltage detect (OVD) alarms

The 744 rising edges of SCLK release  $\overline{\text{BUSY}}$  and start a self timed DAC deglitch period of approximately 3  $\mu\text{s}$ . DAC voltages begin to change as soon as the deglitch circuits time out. An additional 10  $\mu\text{s}$  is required to settle to the final values. A full reset sequence thus requires approximately 30  $\mu\text{s}$ , comprising 16  $\mu\text{s}$  (744 cycles  $\times$  20 ns) for post reset initialization, 3  $\mu\text{s}$  for DAC deglitch, and another 10  $\mu\text{s}$  for DAC analog level settling.

SPI Clock Cycles and the  $\overline{\text{BUSY}}$  Pin

The ADATE320 offers a digital  $\overline{\text{BUSY}}$  output pin to indicate that the SPI controller requires more SCLK cycles to be input on the SCLK pin. The device may be operated without this pin, but care must be exercised to ensure that the required number of SCLK cycles are provided in each case to complete each SPI instruction.

After any valid SPI instruction is written to the ADATE320, the  $\overline{\text{BUSY}}$  pin is asserted to indicate a busy status of the DAC update and calibration routines. The  $\overline{\text{BUSY}}$  pin is an open-drain output capable of sinking a minimum of 2 mA from the VDD supply. It is recommended to tie the  $\overline{\text{BUSY}}$  pin to VDD with an external 1 k $\Omega$  pull-up resistor.

It is not a requirement to wait for release of  $\overline{\text{BUSY}}$  prior to a subsequent assertion of the  $\overline{\text{CS}}$  pin. As long as the minimum number of

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SCLK cycles following the previous release of  $\overline{CS}$  is met according to the  $t_{CSAM}$  parameter, the  $\overline{CS}$  pin can again be asserted for another SPI operation. With the one exception of recovery from a reset request (either by hardware assertion of  $\overline{RST}$  pin or software setting of the internal SPI\_RESET control bit), there is no scenario in normal operation of the ADATE320 in which the user must wait for release of  $\overline{BUSY}$  before asserting the  $\overline{CS}$  pin for a subsequent SPI operation. The only requirement on the assertion of  $\overline{CS}$  is that the  $t_{CSAM}$  parameter has been met as defined in Figure 2 and Table 14.

It is very important, however, that the SCLK pin continue to operate for as long as the  $\overline{BUSY}$  pin state remains active. This period of time is defined by the parameter  $t_{BUSW}$  and is defined in Figure 2, Table 14, and Table 23. If the SCLK pin does not remain active for at least the number of cycles specified, operations pending to the internal processor may not fully complete. In such a case, a temporary malfunction of the ADATE320 may occur, or unexpected results may be obtained. After the device releases the  $\overline{BUSY}$  pin (or the required minimum number of clock cycles is satisfied), SCLK may again be stopped to prevent any unwanted digital noise from coupling into the analog functions. In every case (with no exception for reset recovery), it is the purpose of the  $\overline{BUSY}$  pin to notify the supervisory ASIC or FPGA that it is again safe to stop the SCLK

signal. Running SCLK for extra periods when  $\overline{BUSY}$  is not active is never a problem except for the possibility of adding unwanted digital switching noise into analog functions.

The required length of the  $\overline{BUSY}$  period ( $t_{BUSW}$ ) is variable depending on the particular preceding SPI instruction, but it is always deterministic. It depends only on factors such as whether the previous instruction involved a write to one or more DAC addresses, and, if so, how many channels were involved and whether calibration was enabled. Table 23 details the length of the  $t_{BUSW}$  requirement in units of rising edge SCLK cycles for each possible SPI instruction scenario, including recovery from a hardware  $\overline{RST}$  reset.

Because  $t_{BUSW}$  is deterministic, it is therefore possible to predict in advance the minimum number of rising edge SCLK cycles that are required to complete any given SPI instruction, which makes it possible to operate the device without a need to monitor the  $\overline{BUSY}$  pin. For applications in which it is neither possible nor desirable to monitor the pin, it is acceptable to use the deterministic information provided in Table 23 to guarantee the minimum number of cycles is provided. Either way, it is necessary to honor the minimum number of required rising edge SCLK cycles, as defined by  $t_{BUSW}$ , following the release of  $\overline{CS}$  for each of the SPI instruction scenarios listed.

**Table 23.  $\overline{BUSY}$  Minimum SCLK Cycle Requirements**

SPI Instruction Type (Single- or Dual-Channel Operation)	Minimum $t_{BUSW}$ (SCLK Cycles)
Following Release of Asynchronous $\overline{RST}$ Reset Pin (Hardware Reset)	744
Following Assertion of the SPI_RESET Control Bit (Software Reset)	744
Write to No Operation (NOP) (Address 0x00, Address 0x20, Address 0x50, Address 0x60)	3
Write to a Valid Address That Is Not a DAC (Address > 0x10)	3
Write to Any DAC Except VILx or VIHx (Address 0x01 to Address 0x0F, Except Address 0x01 and Address 0x03)	18
Write to VILx or VIHx DAC (Address 0x01 or Address 0x03)	21

**SPI Read/Write Register Definition**

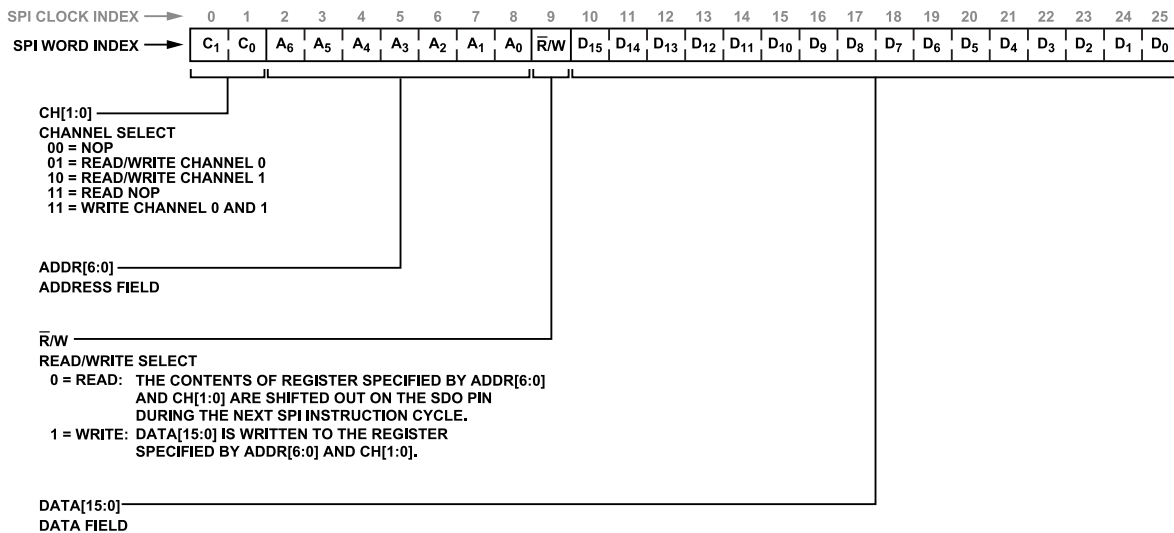


Figure 128. SPI Word Definition

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The ADATE320 is configured through a collection of 16-bit registers as defined in [Table 29](#). Mode configuration, DAC level settings, calibration constants, and alarm flags status can all be controlled and monitored by accessing the respective registers.

Specific access to any 16-bit register is made through a serial programmable interface (SPI). A single SPI control register is exposed to the user by this hardware SPI interface. The format of the SPI Control Register is illustrated in [Figure 128](#). The SPI control register includes address and channel information, read/write direction, and a 16-bit data field. Any valid SPI write instruction cycle populates all these fields, and the ADATE320 subsequently operates on the addressed channel and register using the data provided. Any valid SPI read instruction cycle populates only the address and channel fields, and the ADATE320 makes the addressed register contents in the 16-bit data field available for subsequent readout at the SDO pin.

Detailed SPI timing diagrams for each read/write operation type are provided in [Figure 2](#) through [Figure 7](#). Respective dc and ac timing parameters are provided in [Table 13](#) and [Table 14](#), respectively.

A typical hardware wiring diagram for the SPI is illustrated in [Figure 127](#).

## LEVEL SETTING DACS

### DAC Update Modes

The ADATE320 provides 32 16-bit integrated level setting DACs organized as two channel banks of 16 DACs each. The detailed mapping of each DAC register to each pin electronics function is shown in [Table 29](#). Each DAC can be individually programmed by writing data to the respective SPI register address and channel.

The ADATE320 provides two methods for updating analog DAC levels: DAC immediate update mode and DAC deferred update mode. At the release of the  $\overline{CS}$  pin associated with any valid SPI write to a DAC address, the update of the analog levels can start immediately or can be deferred, depending on the state of the DAC\_LOAD\_MODE control bit in the DAC control register (see [Figure 145](#)). Initiation of the analog level update sequence (and triggering of the on-chip deglitch circuit) begins four SCLK cycles following the associated release of  $\overline{CS}$  pin. For the purpose of this data sheet, the analog level update sequence is assumed to start coincident with the release of  $\overline{CS}$ . The DAC update mode can be selected independently for each channel bank.

If the DAC\_LOAD\_MODE control bit for a given channel bank is cleared, the DACs assigned to that channel bank are placed in the DAC immediate update mode. Writing to any DAC within that channel causes the corresponding analog levels to be updated immediately following the associated release of  $\overline{CS}$ . Because all analog levels are updated on a per channel basis, any previously pending DAC writes queued to that channel (while in an earlier deferred update mode) are also updated at this time. This situation can arise if DAC writes are queued to the channel while in deferred

update mode, and then the DAC\_LOAD\_MODE bit is subsequently changed to immediate update mode before writing to the respective DAC\_LOAD control bit (see [Figure 145](#)). The queued data is not lost. Note that writing to the DAC\_LOAD control bit has no effect while in immediate update mode.

If the DAC\_LOAD\_MODE control bit for a given channel is set, then the DACs assigned to that channel bank is in the deferred update mode. Writing to any DAC of that channel only queues the DAC data into that channel. The analog update of queued DAC levels is deferred until the respective DAC\_LOAD control bit is set (see [Figure 145](#)). The DAC deferred update mode, in conjunction with the respective DAC\_LOAD control bit, provides the means to queue all DAC level writes to a given channel bank before synchronously updating the analog levels with a single SPI command.

The OVDH and OVDL DAC levels do not fit neatly within a particular channel bank. However, they must be updated as a part of the channel bank to which they are assigned, as shown in [Table 29](#).

The ADATE320 provides a feature in which a single SPI write operation can address two channels at one time. With this feature, a single SPI write operation can address corresponding DACs on both channels at the same time, even though the channels may be configured with different DAC update modes. In such a case, the device behaves as expected. For example, if both channels are in immediate update mode, the update of analog levels of both channel banks begins following the associated release of the  $\overline{CS}$  pin. If both channels are in deferred update mode, the update of analog levels is deferred for both channels until the corresponding DAC\_LOAD control bit is set. If one channel is in deferred update mode and the other is in immediate update mode, the deferred channel defers analog updates until the corresponding DAC\_LOAD bit is written, and the immediate channel begins analog updates immediately following release of the  $\overline{CS}$  pin.

An on-chip deglitch circuit with a period of approximately 3  $\mu$ s is provided to prevent DAC-to-DAC crosstalk within a channel whenever an analog update is processed. Each DAC channel bank has its own dedicated deglitch circuitry, and each channel may therefore operate independently.

A deglitch circuit can be retriggered if an analog level update is initiated before a previous update operation on that channel completes. Analog transitions at the DAC outputs do not begin until after the deglitch circuit times out. Final settling to full precision requires an additional 7  $\mu$ s beyond the end of the 3  $\mu$ s deglitch interval. The total DAC settling time following the release of the associated  $\overline{CS}$  pin is approximately 10  $\mu$ s maximum. Note that an extended retriggering sequence of the deglitch circuit on one channel may cause the apparent settling time of analog levels on that channel to appear delayed longer than the specified 10  $\mu$ s.

A typical DAC update sequence is illustrated in [Figure 129](#). In this example, consecutive immediate mode DAC updates are written in direct succession. This example was chosen to illustrate what hap-

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pens when a DAC update command is written before the previous update command finishes its deglitch and settling sequence.

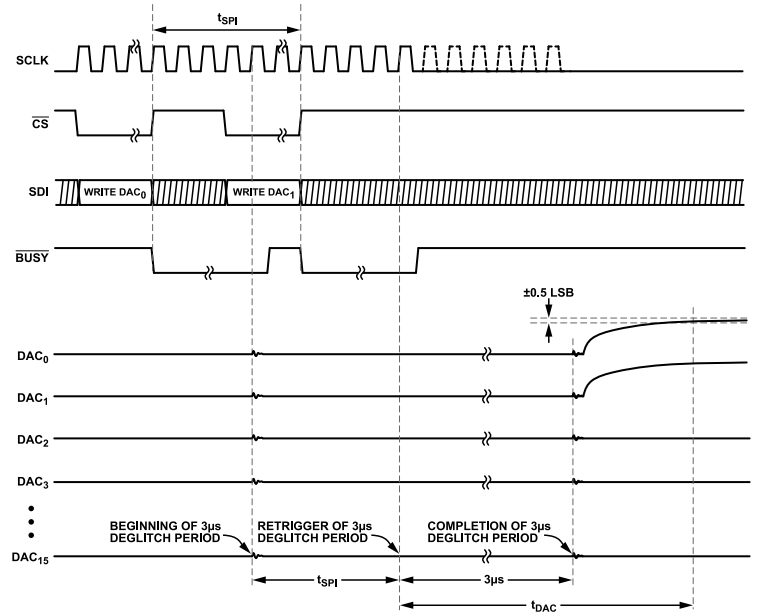


Figure 129. SPI DAC Write Timing Diagram and Settling of DC Levels

**DAC Levels and V THERM Pin Transfer Function**

Table 24. Detailed DAC Code to/from Voltage Level Transfer Functions

Level	Programmable Range (0x0000 to 0xFFFF)	DAC-to-Level and Level-to-DAC Transfer Functions
VILx, VIHx, VITx/VCOMx, VOLx, VOHx, POLx, POHx, VCHx, VCLx, PCHx, PCLx, OVDHx, OVDLx, PPMUx (FV), PCHx (FI), PCLx (FI)	-2.5 V to +7.5 V	$V_{DUTx} = (4 \times (DAC/2^{16}) - 1) \times (V_{REF} - V_{REFGND}) + V_{DUTGND}$ $DAC = ((V_{DUTx} - V_{DUTGND}) / (V_{REF} - V_{REFGND})) / (4 \times (V_{REF} - V_{REFGND})) \times 2^{16}$
IOHx, IOLx	-12.5 mA to +37.5 mA	$I_{DUTx} = (4 \times (DAC/2^{16}) - 1) \times (V_{REF} - V_{REFGND}) \times (25 \text{ mA}/5)$ $DAC = ((I_{DUTx} \times (5/25 \text{ mA})) + (V_{REF} - V_{REFGND})) / (4 \times (V_{REF} - V_{REFGND})) \times 2^{16}$
PPMUx (FI, Range A), PCHx and PCLx (FV, Range A)	-80 mA to +80 mA	$I_{DUTx} = (4 \times (DAC/2^{16}) - 2) \times (V_{REF} - V_{REFGND}) \times (80 \text{ mA}/5)$ $DAC = ((I_{DUTx}/80 \text{ mA} \times 5) + 2 \times (V_{REF} - V_{REFGND})) / (4 \times (V_{REF} - V_{REFGND})) \times 2^{16}$
PPMUx (FI, Range B), PCHx and PCLx (FV, Range B)	-2 mA to +2 mA	$I_{DUTx} = (4 \times (DAC/2^{16}) - 2) \times (V_{REF} - V_{REFGND}) \times (2 \text{ mA}/5)$ $DAC = ((I_{DUTx}/2 \text{ mA} \times 5) + 2 \times (V_{REF} - V_{REFGND})) / (4 \times (V_{REF} - V_{REFGND})) \times 2^{16}$
PPMUx (FI, Range C), PCHx and PCLx (FV, Range C)	-200 µA to +200 µA	$I_{DUTx} = (4 \times (DAC/2^{16}) - 2) \times (V_{REF} - V_{REFGND}) \times (200 \text{ µA}/5)$ $DAC = ((I_{DUTx}/200 \text{ µA} \times 5) + 2 \times (V_{REF} - V_{REFGND})) / (4 \times (V_{REF} - V_{REFGND})) \times 2^{16}$
PPMUx (FI, Range D) PCHx and PCLx (FV, Range D)	-20 µA to +20 µA	$I_{DUTx} = (4 \times (DAC/2^{16}) - 2) \times (V_{REF} - V_{REFGND}) \times (20 \text{ µA}/5)$ $DAC = ((I_{DUTx}/20 \text{ µA} \times 5) + 2 \times (V_{REF} - V_{REFGND})) / (4 \times (V_{REF} - V_{REFGND})) \times 2^{16}$
PPMUx (FI, Range E) PCHx and PCLx (FV, Range E)	-4 µA to +4 µA	$I_{DUTx} = (4 \times (DAC/2^{16}) - 2) \times (V_{REF} - V_{REFGND}) \times (4 \text{ µA}/5)$ $DAC = ((I_{DUTx}/4 \text{ µA} \times 5) + 2 \times (V_{REF} - V_{REFGND})) / (4 \times (V_{REF} - V_{REFGND})) \times 2^{16}$

Table 25. Load Transfer Functions

Load Level	Transfer Functions	Notes
IOLx	$VIOLx / (2 \times (V_{REF} - V_{REFGND})) \times 25 \text{ mA}$	VIOLx DAC levels are not referenced to $V_{DUTGND}$
IOHx	$VIOHx / (2 \times (V_{REF} - V_{REFGND})) \times 25 \text{ mA}$	VIOHx DAC levels are not referenced to $V_{DUTGND}$

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Table 26. PPMU Transfer Functions

PPMU Mode	Transfer Functions <sup>1</sup>	Uncalibrated PPMU DAC Settings to Achieve Specified PPMU Range
FV	$V_{DUTX} = \text{PPMUX}$	$-1.5 \text{ V} < \text{PPMUX} < +4.5 \text{ V}$
FI	$I_{DUTX} = (\text{PPMUX} - (V_{REF} - V_{REFGND})) / (5 \times R_{PPMU})$	$0.0 \text{ V} < \text{PPMUX} < 5.0 \text{ V}$
MV	$V_{PPMU\_Mx} = V_{DUTX}$ (internal sense path)	Not applicable
MV	$V_{PPMU\_Mx} = V_{PPMU\_Sx}$ (external sense path)	Not applicable
MI	$V_{PPMU\_Mx} = (V_{REF} - V_{REFGND}) + (5 \times I_{DUTX} \times R_{PPMU}) + V_{DUTGND}$	Not applicable

<sup>1</sup>  $R_{PPMU} = 12.5 \Omega$  for Range A,  $500 \Omega$  for Range B,  $5.0 \text{ k}\Omega$  for Range C,  $50 \text{ k}\Omega$  for Range D, and  $250 \text{ k}\Omega$  for Range E.

Table 27. Temperature Sensor Transfer Function

Temperature	Output
0 K	0.00 V
300 K	3.00 V
$T_{\text{KELVIN}}$	$0.00 \text{ V} + (T_{\text{KELVIN}}) \times 10 \text{ mV/K}$

## DAC Gain and Offset Correction

Each analog function within the ADATE320 has independent gain (m) and offset (c) calibration registers that allow digital trim of first-order errors in the analog signal chain. These registers correct errors in the pin electronics transfer functions as well as errors intrinsic to the DAC itself.

The m and c registers are volatile and must be reloaded after each power-on cycle as part of a calibration routine if values other than the defaults are required. The registers are not cleared by any reset operation (although the DAC\_CAL\_ENABLE bit is cleared following reset).

The gain and offset calibration function can be bypassed by clearing the DAC\_CAL\_ENABLE bit in the DAC control register (see Figure 145). This bypass mode is available only on a per chip basis. In other words, it is not possible to bypass the calibration function for a specific subset of the DACs.

The calibration function, when enabled, adjusts the numerical data sent to each DAC according to the following equation:

$$X_2 = \left( \left( \frac{m+1}{2^{16}} \right) \times X_1 \right) + (c - 2^{15}) \quad (1)$$

where:

$X_2$  is the 16-bit data-word gated into the physical DAC, and returned by subsequent SPI read from that same DAC.

$m$  is the code in the respective DAC gain calibration register (the default code is  $0x\text{FFFF} = 2^{16} - 1$ ).

$X_1$  is the 16-bit data-word written by the user to the DAC via the SPI.

$c$  is the code in the respective DAC offset calibration register (the default code is  $0x8000 = 2^{15}$ ).

From Equation 1, it can be seen that the gain applied to any written  $X_1$  data is always  $\leq 1.0$ , with the effect that the effective output of a DAC can only be made smaller in magnitude by the calibration mechanism. To compensate for this imposed limitation, each of the analog signal paths in the pin electronics functions are guaranteed

by designed to have a gain  $\geq 1.0$  when the default m register values are applied. A signal path gain  $\geq 1.0$  guarantees that proper gain calibration can always be achieved by multiplying down.

DAC  $X_2$  Registers and SPI Readback

When data is written via the SPI to a particular DAC, that data is operated on in accordance with Equation 1. The results are stored in an  $X_2$  register associated with that DAC (see Figure 130).

There is only a single physical  $X_2$  register per DAC, and it is the value of this  $X_2$  register that is eventually gated into the physical DAC at the time of analog update, which can be either in immediate or deferred mode. It is also this register value that is returned to the user during an SPI read operation addressed to that DAC channel. In the special case of a dual channel write to a DAC, both of the associated  $X_2$  registers are sequentially updated using the appropriate m register and c register for each channel.

When enabled, the calibration function applies this operation to the  $X_2$  registers only after a SPI write to the respective  $X_1$  registers. The  $X_2$  registers are not updated after write operations to either m register or c register or following any changes to functional modes or range settings of the device. For this reason, to ensure that calibration data is recalculated for any particular DAC, it is necessary to write fresh data to that DAC after changes are first made to the associated m register and c register, and any associated functional modes and ranges for that DAC function.

For each DAC, there is only a single  $X_2$  register, and generally there is one dedicated and unique set of m calibration register and c calibration register assigned. In several special cases (for example, the PPMU DAC) there is still only one  $X_2$  register per DAC, but there are several different choices for m register and c register depending on the particular configuration of mode and range control settings for the function. For those DACs, a choice of calibration register is made automatically based on the respective mode and range control settings in place for that function when the DAC is written.



## THEORY OF OPERATION

Table 28 describes detailed m register and c register selection as a function of mode and range control settings. For all DAC functions, it is necessary to ensure that the respective m register and c register values are put in place first, and that the desired mode and

range settings are updated prior to sending data to the DAC. It is only during the DAC write sequence that the calibration constants are selected and applied.

Table 28. m and c Calibration Register Selection

SPI Address [Channel]	DAC Name	Functional (DAC Usage) Description	m Register	c Register	DMC_ENAB LE (Address 0x1A[0]) <sup>1</sup>	LOAD_ENA BLE_x (Address 0x1B[0])	PPMU_MEAS _VI_x (Address 0x1C[6])	PPMU_FORCE_VI_x (Address 0x1C[5])	PPMU_RANG E_x (Address 0x1C[4:2])
0x01[0]	VIH0	Driver high level, Channel 0	0x21[0]	0x31[0]	X	X	X	X	XXX
0x01[1]	VIH1	Driver high level, Channel 1	0x21[1]	0x31[1]	X	X	X	X	XXX
0x02[0]	VIT0/ VCOM0	Driver term level, Channel 0	0x22[0]	0x32[0]	X	0	X	X	XXX
		Load commutation voltage, Channel 0	0x42[0]	0x52[0]	X	1	X	X	XXX
0x02[1]	VIT1/ VCOM0	Driver termination level, Channel 1	0x22[1]	0x32[1]	X	0	X	X	XXX
		Load commutation voltage, Channel 1	0x42[1]	0x52[1]	X	1	X	X	XXX
0x03[0]	VIL0	Driver low level, Channel 0	0x23[0]	0x33[0]	X	X	X	X	XXX
0x03[1]	VIL1	Driver low level, Channel 1	0x23[1]	0x33[1]	X	X	X	X	XXX
0x04[0]	VCH0	Reflection clamp high level, Channel 0	0x24[0]	0x34[0]	X	X	X	X	XXX
0x04[1]	VCH1	Reflection clamp high level, Channel 1	0x24[1]	0x34[1]	X	X	X	X	XXX
0x05[0]	VCL0	Reflection clamp low level, Channel 0	0x25[0]	0x35[0]	X	X	X	X	XXX
0x05[1]	VCL1	Reflection clamp low level, Channel 1	0x25[1]	0x35[1]	X	X	X	X	XXX
0x06[0]	VOH0	Normal window comparator high level, Channel 0	0x26[0]	0x36[0]	0	X	X	X	XXX
		Differential mode comparator high level, Channel 0	0x46[0]	0x56[0]	1	X	X	X	XXX
0x06[1]	VOH1	Normal window comparator high level, Channel 1	0x26[1]	0x36[1]	X	X	X	X	XXX
0x07[0]	VOL0	Normal window comparator low level, Channel 0	0x27[0]	0x37[0]	0	X	X	X	XXX
		Differential mode comparator low level, Channel 0	0x47[0]	0x57[0]	1	X	X	X	XXX
0x07[1]	VOL1	Normal window comparator low level, Channel 1	0x27[1]	0x37[1]	X	X	X	X	XXX
0x08[0]	VIOH0	Load IOHx level, Channel 0	0x28[0]	0x38[0]	X	X	X	X	XXX
0x08[1]	VIOH1	Load IOHx level, Channel 1	0x28[1]	0x38[1]	X	X	X	X	XXX
0x09[0]	VIOL0	Load IOL level, Channel 0	0x29[0]	0x39[0]	X	X	X	X	XXX
0x09[1]	VIOL1	Load IOL level, Channel 1	0x29[1]	0x39[1]	X	X	X	X	XXX
0x0A[0]	PPMU0	PPMU VIN FV level, Channel 0	0x2A[0]	0x3A[0]	X	X	X	0	XXX
		PPMU VIN FI level Range A, Channel 0	0x4A[0]	0x5A[0]	X	X	X	1	111
		PPMU VIN FI level Range B, Channel 0	0x4B[0]	0x5A[0]	X	X	X	1	110

## THEORY OF OPERATION

Table 28. m and c Calibration Register Selection (Continued)

SPI Address [Channel]	DAC Name	Functional (DAC Usage) Description	m Register	c Register	DMC_ENAB	LOAD_ENA	PPMU_MEAS	PPMU_FORCE_ VI_x (Address 0x1C[5])	PPMU_RANG E_x (Address 0x1C[4:2])
					LE (Address 0x1A[0]) <sup>1</sup>	BLE_x (Address 0x1B[0])	_VI_x (Address 0x1C[6])		
		PPMU VIN FI level Range C, Channel 0	0x4C[0]	0x5A[0]	X	X	X	1	101
		PPMU VIN FI level Range D, Channel 0	0x4D[0]	0x5A[0]	X	X	X	1	100
		PPMU VIN FI level Range E, Channel 0	0x4E[0]	0x5A[0]	X	X	X	1	0XX
0x0A[1]	PPMU1	PPMU VIN FV level, Channel 1	0x2A[1]	0x3A[1]	X	X	X	0	XXX
		PPMU VIN FI level Range A, Channel 1	0x4A[1]	0x5A[1]	X	X	X	1	111
		PPMU VIN FI level Range B, Channel 1	0x4B[1]	0x5A[1]	X	X	X	1	110
		PPMU VIN FI level Range C, Channel 1	0x4C[1]	0x5A[1]	X	X	X	1	101
		PPMU VIN FI level Range D, Channel 1	0x4D[1]	0x5A[1]	X	X	X	1	100
		PPMU VIN FI level Range E, Channel 1	0x4E[1]	0x5A[1]	X	X	X	1	0XX
0x0B[0]	PCH0	PPMU current clamp (FV) high lev- el, Channel 0	0x44[0]	0x54[0]	X	X	X	0	XXX
		PPMU voltage clamp (FI) high lev- el, Channel 0	0x2B[0]	0x3B[0]				1	
0x0B[1]	PCH1	PPMU current clamp (FV) high lev- el, Channel 1	0x44[1]	0x54[1]	X	X	X	0	XXX
		PPMU voltage clamp (FI) high lev- el, Channel 1	0x2B[1]	0x3B[1]				1	
0x0C[0]	PCL0	PPMU current clamp (FV) low lev- el, Channel 0	0x45[0]	0x55[0]	X	X	X	0	XXX
		PPMU voltage clamp (FI) low level, Channel 0	0x2C[0]	0x3C[0]				1	
0x0C[1]	PCL1	PPMU current clamp (FV) low lev- el, Channel 1	0x45[1]	0x55[1]	X	X	X	0	XXX
		PPMU voltage clamp (FI) low level, Channel 1	0x2C[1]	0x3C[1]	X	X	X	1	
0x0D[0]	POH0	PPMU go/no-go MV high level, Channel 0	0x2D[0]	0x3D[0]	X	X	0	X	XXX
		PPMU go/no-go MI Range A high level, Channel 0	0x61[0]	0x5D[0]	X	X	1	X	111
		PPMU go/no-go MI Range B high level, Channel 0	0x62[0]	0x5D[0]	X	X	1	X	110
		PPMU go/no-go MI Range C high level, Channel 0	0x63[0]	0x5D[0]	X	X	1	X	101
		PPMU go/no-go MI Range D high level, Channel 0	0x64[0]	0x5D[0]	X	X	1	X	100
		PPMU go/no-go MI Range E high level, Channel 0	0x65[0]	0x5D[0]	X	X	1	X	0XX



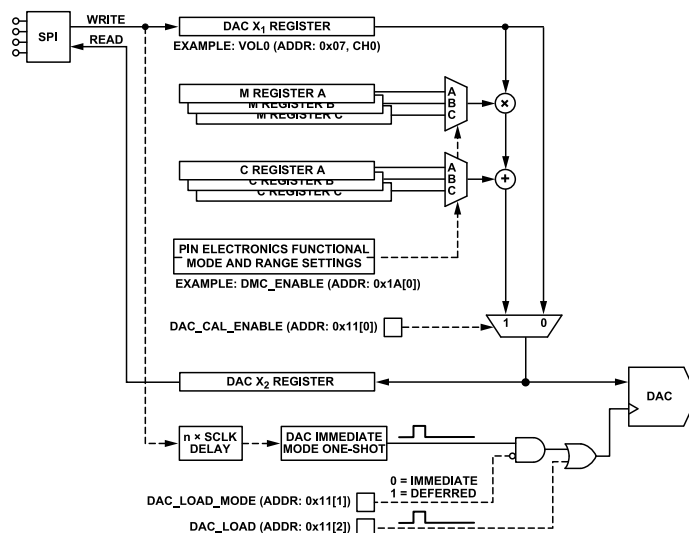
## THEORY OF OPERATION

Table 28. m and c Calibration Register Selection (Continued)

SPI Address [Channel]	DAC Name	Functional (DAC Usage) Description	m Register	c Register	DMC_ENAB	LOAD_ENA	PPMU_MEAS	PPMU_FORCE_ VI_x (Address 0x1C[5])	PPMU_RANG E_x (Address 0x1C[4:2])
					LE (Address 0x1A[0]) <sup>1</sup>	BLE_x (Address 0x1B[0])	_VI_x (Address 0x1C[6])		
0x0D[1]	POH1	PPMU go/no-go MV high level, Channel 1	0x2D[1]	0x3D[1]	X	X	0	X	XXX
		PPMU go/no-go MI Range A high level, Channel 1	0x61[1]	0x5D[1]	X	X	1	X	111
		PPMU go/no-go MI Range B high level, Channel 1	0x62[1]	0x5D[1]	X	X	1	X	110
		PPMU go/no-go MI Range C high level, Channel 1	0x63[1]	0x5D[1]	X	X	1	X	101
		PPMU go/no-go MI Range D high level, Channel 1	0x64[1]	0x5D[1]	X	X	1	X	100
		PPMU go/no-go MI Range E high level, Channel 1	0x65[1]	0x5D[1]	X	X	1	X	0XX
0x0E[0]	POLO	PPMU go/no-go MV low level, Channel 0	0x2E[0]	0x3E[0]	X	X	0	X	XXX
		PPMU go/no-go MI Range A low level, Channel 0	0x66[0]	0x5E[0]	X	X	1	X	111
		PPMU go/no-go MI Range B low level, Channel 0	0x67[0]	0x5E[0]	X	X	1	X	110
		PPMU go/no-go MI Range C low level, Channel 0	0x68[0]	0x5E[0]	X	X	1	X	101
		PPMU go/no-go MI Range D low level, Channel 0	0x69[0]	0x5E[0]	X	X	1	X	100
		PPMU go/no-go MI Range E low level, Channel 0	0x6A[0]	0x5E[0]	X	X	1	X	0XX
0x0E[1]	POL1	PPMU go/no-go MV low level, Channel 1	0x2E[1]	0x3E[1]	X	X	0	X	XXX
		PPMU go/no-go MI Range A low level, Channel 1	0x66[1]	0x5E[1]	X	X	1	X	111
		PPMU go/no-go MI Range B low level, Channel 1	0x67[1]	0x5E[1]	X	X	1	X	110
		PPMU go/no-go MI Range C low level, Channel 1	0x68[1]	0x5E[1]	X	X	1	X	101
		PPMU go/no-go MI Range D low level, Channel 1	0x69[1]	0x5E[1]	X	X	1	X	100
		PPMU go/no-go MI Range E low level, Channel 1	0x6A[1]	0x5E[1]	X	X	1	X	0XX
0x0F[0]	OVDL	Overvoltage detect low level	0x2F[0]	0x3F[0]	X	X	X	X	XXX
0x0F[1]	OVDH	Overvoltage detect high level	0x2F[1]	0x3F[1]	X	X	X	X	XXX

<sup>1</sup> X means don't care.

## THEORY OF OPERATION

Figure 130. DAC X<sub>2</sub> Registers and Calibration Diagram

## ALARM FUNCTIONS

The ADATE320 contains per channel overvoltage detectors (OVDL/OVDH), per channel PPMU voltage/current clamps (PCLx/PCHx), and a thermal alarm to detect and signal these respective fault conditions. Any of these functions may flag an alarm independently in the alarm state register (see Figure 152). The status of the alarms may be determined at any time by reading the SPI alarm state register. This register is read only, and its contents are cleared by the read operation. The alarm flag bits can then become set by any of the respective alarm functions. The individual fault condition flags are logically OR'ed together to drive the open-drain **ALARM** output pin to indicate that a fault condition has occurred (see Figure 134).

The various alarm flags can be either enabled or disabled (masked) using the alarm mask register (see Figure 151). The thermal alarm is enabled by default (mask bit clear), and the overvoltage and PPMU clamp alarms are all disabled by default (mask bits set).

The PPMU clamp alarm behavior depends on the mode of the PPMU. When in FI mode, the PPMU clamps behave as programmable voltage clamps. The high and low voltage clamp levels are set by the respective PCHx and PCLx level setting DACs. If the voltage on the DUTx pin reaches either the PCHx or PCLx setting, a PPMU clamp alarm is generated, but only if the clamps are enabled with the PPMU\_CLAMP\_ENABLE\_x control bit in the PPMU

control register (see Figure 150). Note that if the PPMU clamps are enabled and a PPMU clamp alarm is generated, the alarm can still be masked with the alarm mask register. However, if the voltage clamps are disabled, no PPMU clamp alarm is generated.

When the PPMU is in FV mode, the PPMU clamps behave as programmable current clamps. The source and sink current clamp levels are set with the respective PCHx and PCLx level setting DACs. The current clamps cannot be disabled by setting or clearing the PPMU\_CLAMP\_ENABLE\_x control bit—the clamps are always active when in PPMU FV mode. If the PCHx and PCLx levels are set outside their functional range, a  $\pm 140\%$  static current limit is left in effect. If the current on a DUTx pin reaches either the PCHx or PCLx clamp setting, or, alternatively, one of the static current limits, a PPMU clamp alarm results. The PPMU clamp alarm can be masked separately in the alarm mask register.

Refer to Figure 131 through Figure 134 for more information about PPMU clamp functions.

The only purpose of the various alarm circuits is to detect and indicate the presence of a fault condition of interest to the user. The only action the ADATE320 takes upon detection of a fault is to set the appropriate alarm state register flag bits in the alarm state register and then activate the open-drain **ALARM** pin. No other action is taken.

THEORY OF OPERATION

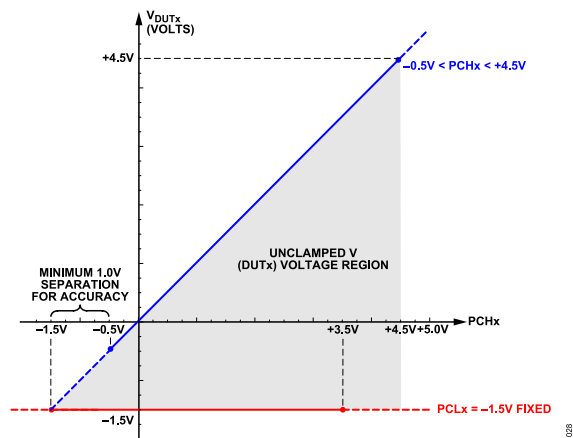


Figure 131. PPMU Voltage Clamp High, Functional Diagram (Voltage Clamp Low Fixed at -1.5 V)

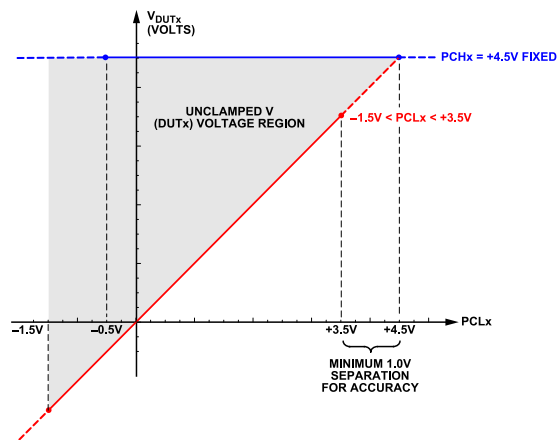


Figure 132. PPMU Voltage Clamp Low, Functional Diagram (Voltage Clamp High Fixed at 4.5 V)

THEORY OF OPERATION

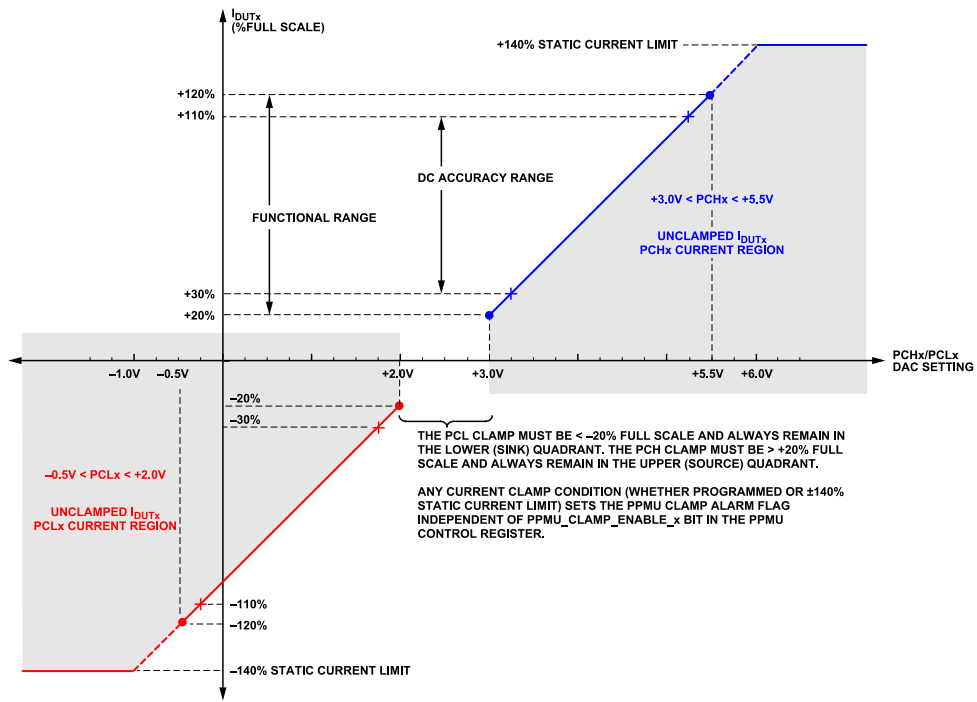


Figure 133. PPMU Current Clamp High and Low, Functional Diagram

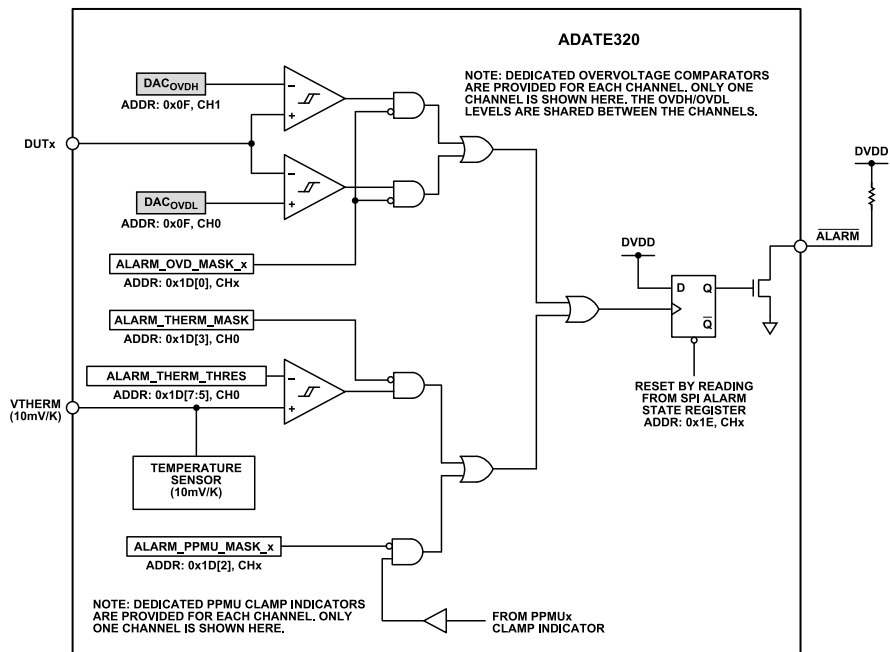


Figure 134. Fault Alarm Functional Block Diagram

## APPLICATIONS INFORMATION

### POWER SUPPLY, GROUNDING, AND TYPICAL DECOUPLING STRATEGY

The ADATE320 is internally divided into a digital core and an analog core.

The VDD and DGND pins provide power and ground for the digital core that includes the SPI, certain logic functions, and the digital calibration functions. DGND is the logic ground reference for the VDD supply. Therefore, bypass VDD adequately to DGND with good quality, low effective series resistance (ESR) bypass capacitors. To reduce transient digital switching noise coupling to the analog core, connect DGND to a dedicated external ground plane that is separated from the analog ground domains. If the application permits, the DGND pins can share a digital ground domain with the supervisory FPGA or ASIC that interfaces with the ADATE320 SPI. All CMOS inputs and outputs are referenced between VDD and DGND, and their valid levels must be guaranteed relative to these power supply pins.

The analog core of the device includes all analog ATE functional blocks such as the DACs, the driver, the comparator, the load, and the PPMU. The VCC and VEE supplies provide power to the analog core. AGND and PGND are analog ground and power ground references, respectively. PGND is generally noisier with analog switching transients, and it may also have large static dc currents. AGND is generally quieter and has relatively smaller static dc currents. These two grounds can be connected together outside the chip to a single shared analog ground plane. Regardless, keep PGND and AGND (whether separated or shared) separated from the DGND ground plane if system design constraints permit.

The transient frequencies generated by the analog core can be a full order of magnitude greater than those generated by the SPI and on-chip digital circuitry. Therefore, pay close attention to the decoupling of the VCC and VEE supplies. Each supply must be adequately bypassed to the PGND ground domain using the highest quality bypass capacitors available. Locate the decoupling capacitors as close to the device as practically possible. The decoupling capacitors must have very low ESR and effective series inductance (ESL). Commonly available ceramic capacitors may provide only a marginally low impedance path to ground at the frequencies

encountered in the ADATE320. Therefore, consider only the highest performance decoupling capacitors if possible. In accordance with generally accepted practices, a typical 10  $\mu$ F tantalum capacitor must also be shared across each power supply domain.

Pay particularly close attention to decoupling the VCC and VEE supplies in proximity to the transmission line at the DUTx pins of the device. To avoid undesired waveform aberrations and degradation of performance, it is important that all return currents to and from the transmission line have a direct and low impedance path back to the VCCDx and VEEDx pins adjacent to the respective DUTx pins. See [Figure 135](#) for a typical transmission line decoupling strategy.

The ADATE320 has a DUTGND reference input pin that senses the remote low frequency ground potential at the target device under test (DUT). With the exception of the VIOH and VIOL active load currents and VPMU when in PPMU FI mode, all DAC levels are adjusted on-chip relative to this DUTGND input. Furthermore, the PPMU measure output pins (PPMU\_Mx) are also referenced to DUTGND. The off-chip system analog-to-digital converter (ADC) that measures the PPMU\_Mx pins must therefore be referenced to DUTGND as well. Referencing the system ADC to AGND results in errors unless DUTGND is tied directly to AGND as close as possible to the ADATE320. For applications that do not distinguish between DUT ground reference and system analog ground reference, the DUTGND pin may be connected to the same ground plane as AGND.

Avoid routing digital lines under the device, because these lines can couple noise into the device. Generous use of an analog ground plane under the device shields noise coupling that can otherwise enter the device. The power supply distribution lines must provide very wide and low inductance paths to the respective supply planes. This is especially true for VCC and VEE. Attention to via inductance is extremely important in these supplies—it cannot be neglected. Fast switching signals routed in proximity to the ADATE320 must be adequately shielded, preferably with their proper ground returns to avoid radiating noise to other parts of the board. Route such lines as far away as possible from the analog inputs to the device, such as the AGND, DUTGND, VREF, and VREFGND reference inputs.

## APPLICATIONS INFORMATION

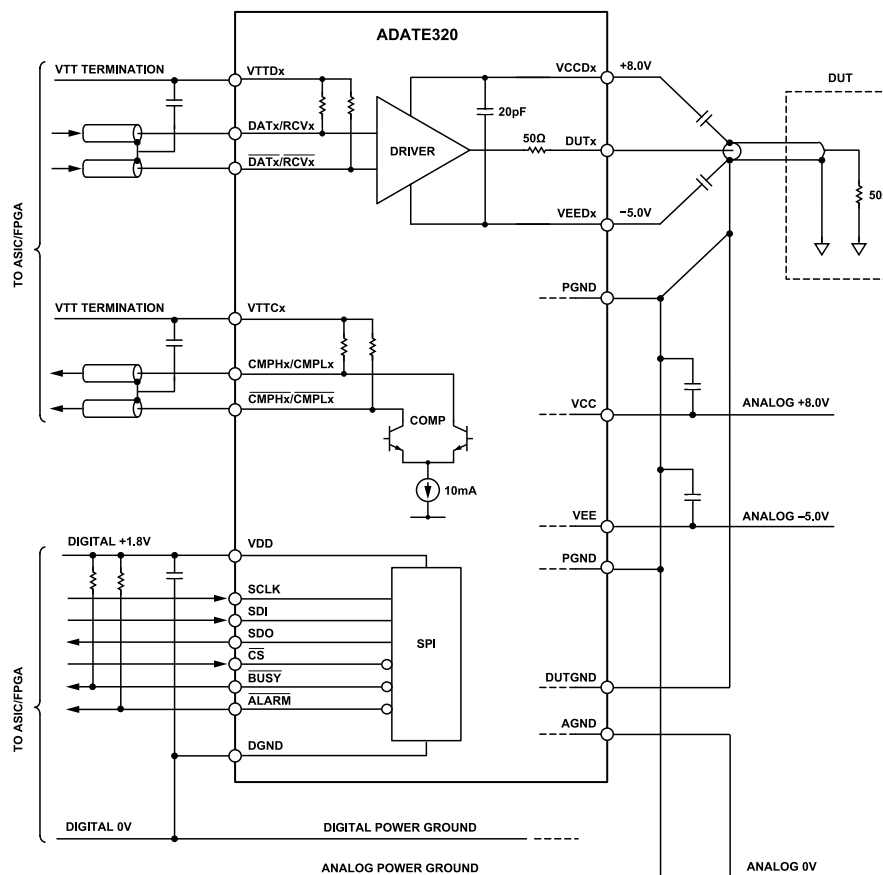


Figure 135. Power Supply and Transmission Path Decoupling Detail

## POWER SUPPLY SEQUENCING

The ADATE320 is designed to tolerate sequencing of power supplies in any order. It is therefore not critical that the power supplies be sequenced in any particular order; however, there are recommended best practices.

The ADATE320 has two analog power supplies ( $V_{CC}$ ,  $V_{EE}$ ) and one digital power supply ( $V_{DD}$ ). The analog supplies service all of the analog functions on the chip such as level setting DACs, driver, comparator, load, and PPMU. The digital supply services the SPI and all digital CMOS control circuitry.

There is careful separation between the analog and digital partitions of the chip, and significant effort has been made to decouple these two partitions both functionally and electrically. The analog partition remains in the default configuration in the absence of  $V_{DD}$ , and similarly, the digital partition remains in the default configuration in the absence of either (or both)  $V_{CC}$  and  $V_{EE}$ .

It is not possible to guarantee predictable behavior of the analog partition if either the  $V_{EE}$  or  $V_{CC}$  supply is poorly conditioned or absent. It is therefore recommended that any externally connected device be disconnected from the DUTx pin to prevent potential damage to that device while either of the  $V_{EE}$  or  $V_{CC}$  supplies is out of specification.

Assuming the  $V_{EE}$  and  $V_{CC}$  analog supplies are both applied and within specification, the analog partition ensures that all functions remain in the default configuration. This is true even when the  $V_{DD}$  supply is absent and digital CMOS control circuitry is not yet functioning. In such a case (or whenever the  $\overline{RST}$  pin is asserted), all of the level setting DACs takes the voltage present at the DUTGND input pin, and all SPI control bits assume their reset default values. The analog functions remain in this safe condition as long as  $V_{DD}$  remains absent or as long as the  $\overline{RST}$  pin remains asserted.

It is recommended that the  $\overline{RST}$  pin always be asserted during the time that the  $V_{DD}$  supply is being brought up. If this condition is met, the level setting DACs continue to hold the DUTGND potential after  $V_{DD}$  stabilizes and after the  $\overline{RST}$  pin is released. A fully clocked reset sequence then initializes the level setting DACs to the reset default conditions as specified in [Table 29](#).

The reset sequence is described in more detail in the [SPI Reset Sequence and the RST Pin](#) section.

In light of these considerations, it is recommended that the two analog supplies be applied first. It is preferable that the smaller valued supply ( $V_{EE}$ ) be applied before the larger valued supply ( $V_{CC}$ ). Bring up the digital  $V_{DD}$  supply next while the RST pin

## APPLICATIONS INFORMATION

is asserted. After  $V_{DD}$  is stable and the  $\overline{RST}$  pin is subsequently released, a fully clocked reset sequence must follow. This power supply sequence ensures that analog functions and all level setting DACs receive the proper configuration information during the digital partition reset sequence.

The power supplies must be removed in the reverse order.

Note that VREF and the high speed transmission line termination pins (VTTDx, VTTCx) are all part of the analog partition, but they are not treated as supplies. VREF can be managed independent of  $V_{CC}$  and  $V_{EE}$ , provided its potential never goes outside those of the  $V_{EE}$  and  $V_{CC}$  supply buses to prevent ESD protection diodes from becoming forward biased. The VTTDx and VTTCx pins do not have this restriction relative to  $V_{CC}$  and  $V_{EE}$ , but they must never go outside the absolute maximum ratings as measured with respect to PGND.

DETAILED FUNCTIONAL BLOCK DIAGRAMS

Figure 141 through Figure 144 illustrate the top-level functionality of the capabilities of the ADATE320 for the driver, comparator, active load, and PPMU.

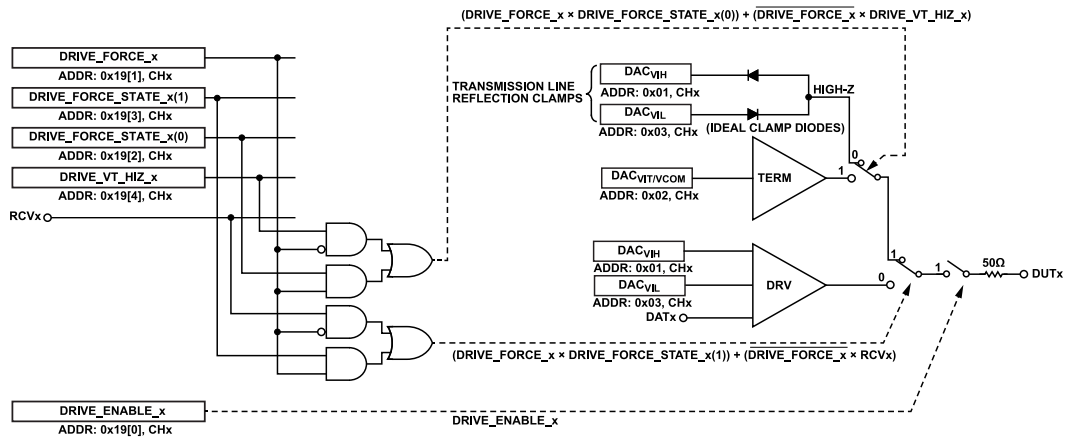


Figure 136. Driver Block Diagram

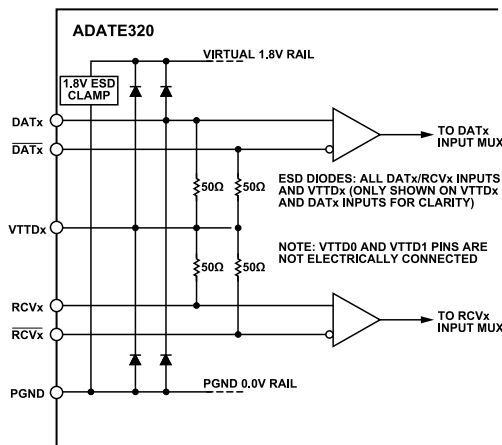


Figure 137. Driver Equivalent Input Stage Diagram



DETAILED FUNCTIONAL BLOCK DIAGRAMS

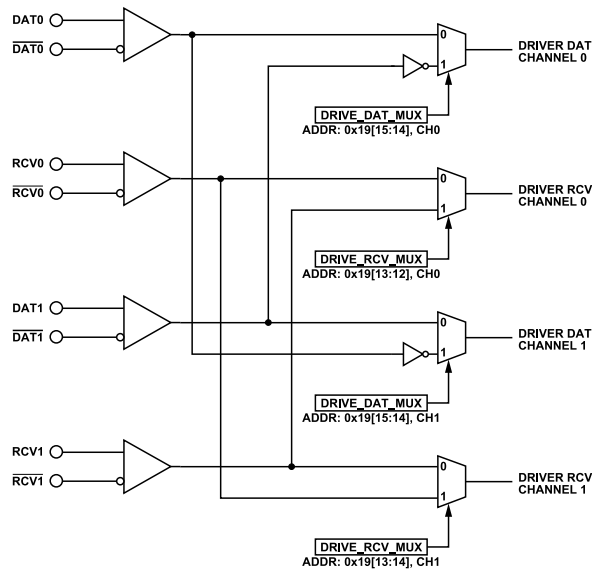


Figure 138. Driver Input Multiplex Diagram

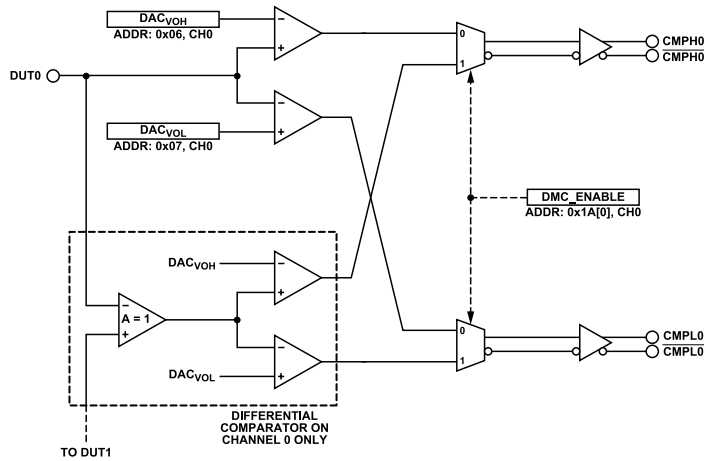


Figure 139. Comparator Functional Block Diagram

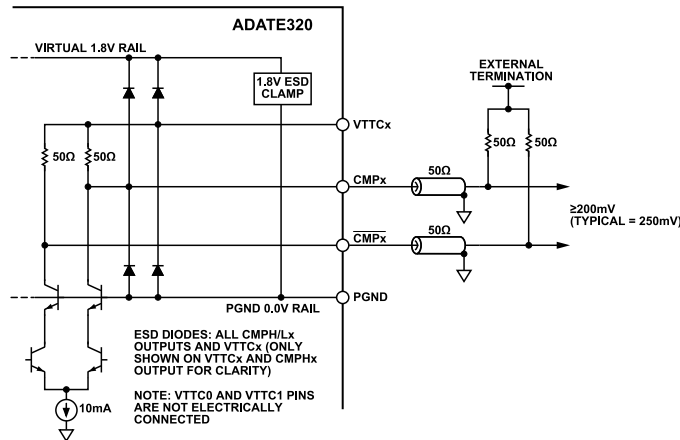


Figure 140. Comparator Equivalent Output Stage Diagram

DETAILED FUNCTIONAL BLOCK DIAGRAMS

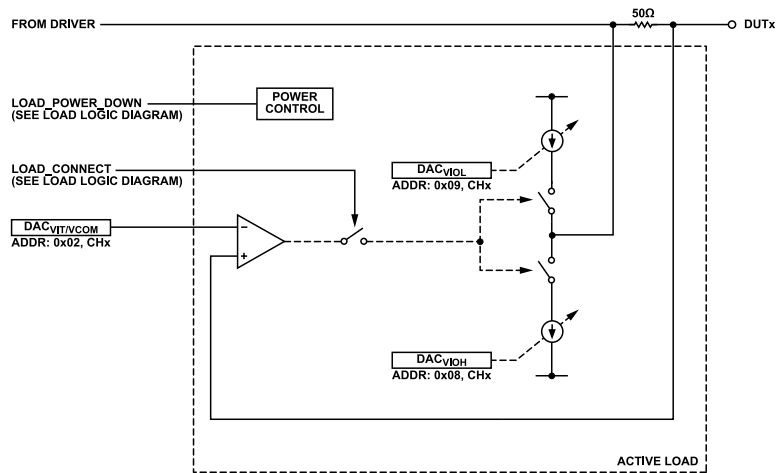


Figure 141. Active Load Functional Block Diagram

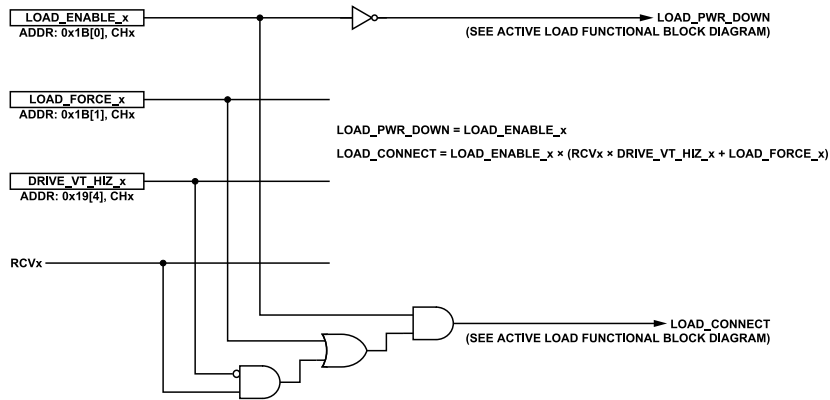


Figure 142. Active Load Functional Logic Diagram

DETAILED FUNCTIONAL BLOCK DIAGRAMS

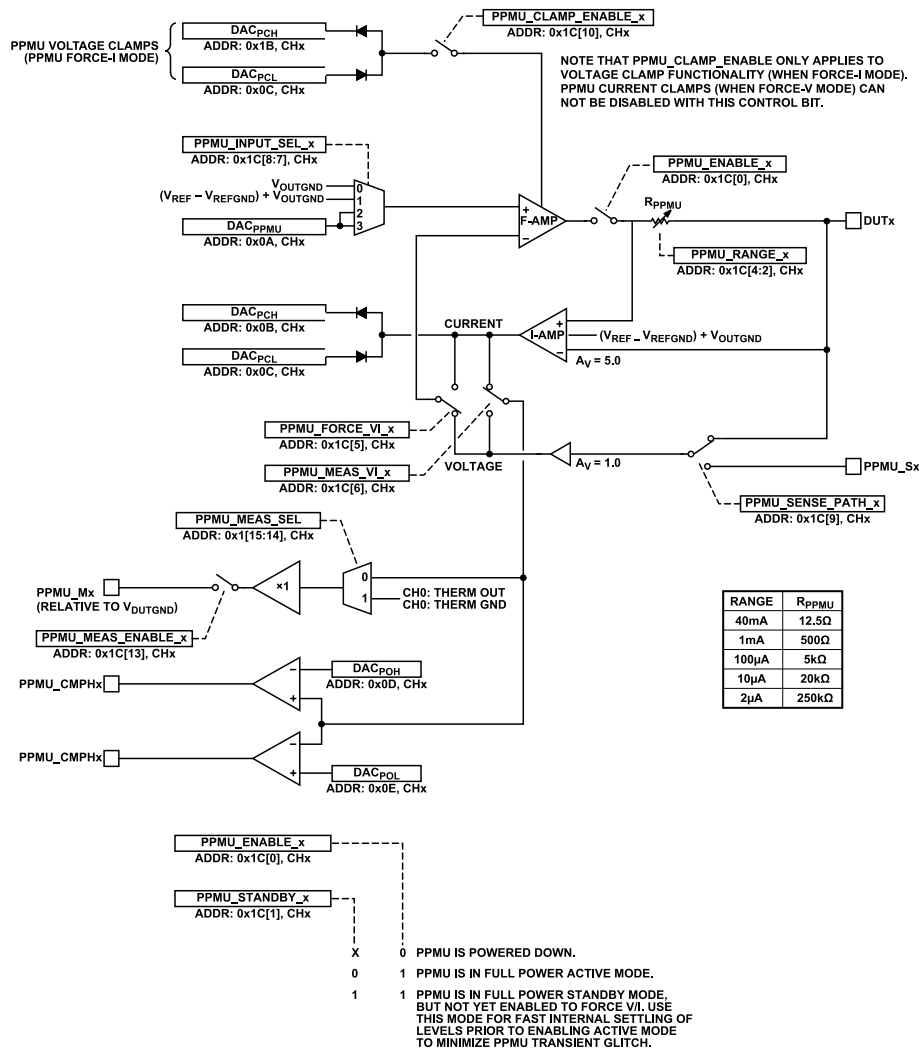


Figure 143. PPMU Functional Block Diagram

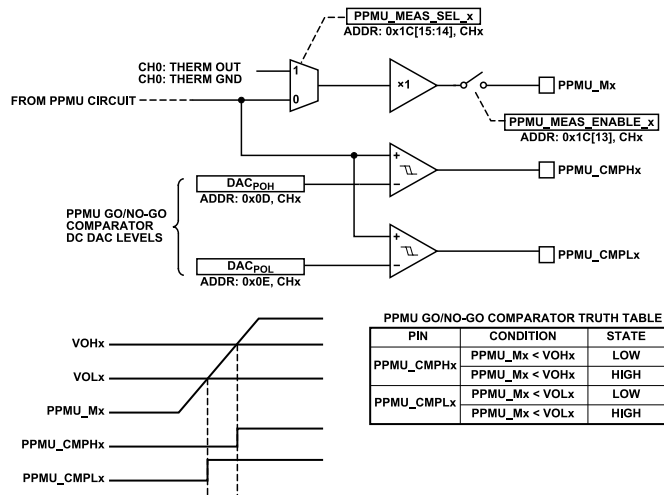


Figure 144. PPMU Go/No-Go Comparator Functional Block Diagram

## SPI REGISTER MEMORY MAP AND DETAILS

## MEMORY MAP

Table 29. SPI Register Memory Map

CH[1:0] <sup>1,2</sup>	Address (ADDR[6:0])	R/W	DATA[15:0] <sup>3</sup>	Register Description	Reset Value
XX	0x00	X	XXXX	NOP	
CC	0x01	R/W	DDDD	VIH DAC level (reset value = 0.0 V)	0x4000
CC	0x02	R/W	DDDD	VIT/VCOM DAC level (reset value = 0.0 V)	0x4000
CC	0x03	R/W	DDDD	VIL DAC level (reset value = 0.0 V)	0x4000
CC	0x04	R/W	DDDD	VCHx DAC level (reset value = $V_{MAX}$ )	0xFFFF
CC	0x05	R/W	DDDD	VLx DAC level (reset value = $V_{MIN}$ )	0x0000
CC	0x06	R/W	DDDD	VOHx DAC level (reset value = 4.0 V)	0xA666
CC	0x07	R/W	DDDD	VOLx DAC level (reset value = -1.0 V)	0x2666
CC	0x08	R/W	DDDD	VIOH DAC level (reset value $\geq 0 \mu A$ ) <sup>4</sup>	0x4000
CC	0x09	R/W	DDDD	VIOL DAC level (reset value $\geq 0 \mu A$ ) <sup>4</sup>	0x4000
CC	0x0A	R/W	DDDD	PPMU DAC level (reset value = 0.0 V)	0x4000
CC	0x0B	R/W	DDDD	PCHx DAC level (reset value = $V_{MAX}$ )	0xFFFF
CC	0x0C	R/W	DDDD	PCLx DAC level (reset value = $V_{MIN}$ )	0x0000
CC	0x0D	R/W	DDDD	POHx DAC level (reset value = 4.0 V)	0xA666
CC	0x0E	R/W	DDDD	POLx DAC level (reset value = -1.0 V)	0x2666
01	0x0F	R/W	DDDD	OVDL DAC level (reset value = $V_{MIN}$ )	0x0000
10	0x0F	R/W	DDDD	OVDH DAC level (reset value = $V_{MAX}$ )	0xFFFF
XX	0x10	X	XXXX	Reserved	
CC	0x11	R/W	DDDD	DAC control register	0x0000
01	0x12	R/W	DDDD	SPI control register	0x0000
XX	0x13 to 0x18	X	XXXX	Reserved	
CC	0x19	R/W	DDDD	DRV control register	0x0000
CC	0x1A	R/W	DDDD	CMP control register	0xFF00
CC	0x1B	R/W	DDDD	Load control register	0x0003
CC	0x1C	R/W	DDDD	PPMU control register	0x0000
01	0x1D	R/W	DDDD	Alarm mask register	0x0085
10	0x1D	R/W	DDDD	Alarm mask register	0x0005
CC	0x1E	R	DDDD	Alarm state register	0x0000
CC	0x1F	R/W	DDDD	Product serialization code register	Unique
XX	0x20	X	XXXX	NOP	
CC	0x21	R/W	DDDD	VIH (driver high level) m coefficient	0xFFFF
CC	0x22	R/W	DDDD	VIT (driver term level) m coefficient	0xFFFF
CC	0x23	R/W	DDDD	VIL (driver low level) m coefficient	0xFFFF
CC	0x24	R/W	DDDD	VCHx (driver reflection clamp) m coefficient	0xFFFF
CC	0x25	R/W	DDDD	VLx (driver reflection clamp) m coefficient	0xFFFF
CC	0x26	R/W	DDDD	VOHx (normal window comparator) m coefficient	0xFFFF
CC	0x27	R/W	DDDD	VOLx (normal window comparator) m coefficient	0xFFFF
CC	0x28	R/W	DDDD	VIOH (active load IOHx) m coefficient	0xFFFF
CC	0x29	R/W	DDDD	VIOL (active load IOL) m coefficient	0xFFFF
CC	0x2A	R/W	DDDD	PPMU (PPMU FV) m coefficient	0xFFFF
CC	0x2B	R/W	DDDD	PCHx (PPMU voltage clamp, FI) m coefficient	0xFFFF
CC	0x2C	R/W	DDDD	PCLx (PPMU voltage clamp, FI) m coefficient	0xFFFF
CC	0x2D	R/W	DDDD	POHx (PPMU comparator MV) m coefficient	0xFFFF

## SPI REGISTER MEMORY MAP AND DETAILS

Table 29. SPI Register Memory Map (Continued)

CH[1:0] <sup>1,2</sup>	Address (ADDR[6:0])	R/W	DATA[15:0] <sup>3</sup>	Register Description	Reset Value
CC	0x2E	R/W	DDDD	POLx (PPMU comparator MV) m coefficient	0xFFFF
01	0x2F	R/W	DDDD	OVDL m coefficient	0xFFFF
10	0x2F	R/W	DDDD	OVDH m coefficient	0xFFFF
XX	0x30	X	XXXX	Reserved	
CC	0x31	R/W	DDDD	VIH (driver high level) c coefficient	0x8000
CC	0x32	R/W	DDDD	VIT (driver term level) c coefficient	0x8000
CC	0x33	R/W	DDDD	VIL (driver low level) c coefficient	0x8000
CC	0x34	R/W	DDDD	VCHx (driver reflection clamp) c coefficient	0x8000
CC	0x35	R/W	DDDD	VLx (driver reflection clamp) c coefficient	0x8000
CC	0x36	R/W	DDDD	VOHx (normal window comparator) c coefficient	0x8000
CC	0x37	R/W	DDDD	VOLx (normal window comparator) c coefficient	0x8000
CC	0x38	R/W	DDDD	VIOH (active load IOHx) c coefficient	0x8000
CC	0x39	R/W	DDDD	VIOL (active load IOL) c coefficient	0x8000
CC	0x3A	R/W	DDDD	PPMU (PPMU FV) c coefficient	0x8000
CC	0x3B	R/W	DDDD	PCHx (PPMU voltage clamp, FI) c coefficient	0x8000
CC	0x3C	R/W	DDDD	PCLx (PPMU voltage clamp, FI) c coefficient	0x8000
CC	0x3D	R/W	DDDD	POHx (PPMU comparator MV) c coefficient	0x8000
CC	0x3E	R/W	DDDD	POLx (PPMU comparator MV) c coefficient	0x8000
01	0x3F	R/W	DDDD	OVDL c coefficient	0x8000
10	0x3F	R/W	DDDD	OVDH c coefficient	0x8000
XX	0x40 to 0x41	X	XXXX	Reserved	
CC	0x42	R/W	DDDD	VCOM (active load) m coefficient	0xFFFF
XX	0x43	X	XXXX	Reserved	
CC	0x44	R/W	DDDD	PCHx (PPMU current clamp, FV) m coefficient	0xFFFF
CC	0x45	R/W	DDDD	PCLx (PPMU current clamp, FV) m coefficient	0xFFFF
01	0x46	R/W	DDDD	VOHx (differential comparator) m coefficient	0xFFFF
01	0x47	R/W	DDDD	VOLx (differential comparator) m coefficient	0xFFFF
XX	0x48 to 0x49	X	XXXX	Reserved	
CC	0x4A	R/W	DDDD	PPMU FI Range A m coefficient	0xFFFF
CC	0x4B	R/W	DDDD	PPMU FI Range B m coefficient	0xFFFF
CC	0x4C	R/W	DDDD	PPMU FI Range C m coefficient	0xFFFF
CC	0x4D	R/W	DDDD	PPMU FI Range D m coefficient	0xFFFF
CC	0x4E	R/W	DDDD	PPMU FI Range E m coefficient	0xFFFF
XX	0x4F	X	XXXX	Reserved	
XX	0x50 to 0x51	X	XXXX	Reserved	
CC	0x52	R/W	DDDD	VCOM (active load) c coefficient	0x8000
XX	0x53	X	XXXX	Reserved	
CC	0x54	R/W	DDDD	PCHx (PPMU current clamp, FV) c coefficient	0x8000
CC	0x55	R/W	DDDD	PCLx (PPMU current clamp, FV) c coefficient	0x8000
01	0x56	R/W	DDDD	VOHx (differential comparator) c coefficient	0x8000
01	0x57	R/W	DDDD	VOLx (differential comparator) c coefficient	0x8000
XX	0x58 to 0x59	X	XXXX	Reserved	
CC	0x5A	R/W	DDDD	PPMU FI c coefficient	0x8000
XX	0x5B to 0x5C	X	XXXX	Reserved	

## SPI REGISTER MEMORY MAP AND DETAILS

Table 29. SPI Register Memory Map (Continued)

CH[1:0] <sup>1,2</sup>	Address (ADDR[6:0])	R/W	DATA[15:0] <sup>3</sup>	Register Description	Reset Value
CC	0x5D	R/W	DDDD	POHx (PPMU comparator MI) c coefficient	0x8000
CC	0x5E	R/W	DDDD	POLx (PPMU comparator MI) c coefficient	0x8000
XX	0x5F	X	XXXX	Reserved	
XX	0x60	X	XXXX	Reserved	
CC	0x61	R/W	DDDD	POHx (PPMU comparator MI Range A) m coefficient	0xFFFF
CC	0x62	R/W	DDDD	POHx (PPMU comparator MI Range B) m coefficient	0xFFFF
CC	0x63	R/W	DDDD	POHx (PPMU comparator MI Range C) m coefficient	0xFFFF
CC	0x64	R/W	DDDD	POHx (PPMU comparator MI Range D) m coefficient	0xFFFF
CC	0x65	R/W	DDDD	POHx (PPMU comparator MI Range E) m coefficient	0xFFFF
CC	0x66	R/W	DDDD	POLx (PPMU comparator MI Range A) m coefficient	0xFFFF
CC	0x67	R/W	DDDD	POLx (PPMU comparator MI Range B) m coefficient	0xFFFF
CC	0x68	R/W	DDDD	POLx (PPMU comparator MI Range C) m coefficient	0xFFFF
CC	0x69	R/W	DDDD	POLx (PPMU comparator MI Range D) m coefficient	0xFFFF
CC	0x6A	R/W	DDDD	POLx (PPMU comparator MI Range E) m coefficient	0xFFFF
XX	0x6B to 0x7F	X	XXXX	Reserved	

<sup>1</sup> X means don't care for the respective field.

<sup>2</sup> CC represents two contiguous binary channel bits.

<sup>3</sup> DDDD represents four-digit hexadecimal data.

<sup>4</sup> The active load VIOHx and VIOLx voltage offsets are guaranteed to be nonzero and positive. These offsets result in a nonzero current for each IOHx and IOLx level following valid reset sequence and prior to calibration. Furthermore, the active load is forced into the active on state following a reset, which facilitates a soft connect of the DUTx pin to VCOMx = 0.0 V following a valid reset sequence (with small but nonzero IOHx and IOLx currents).

## REGISTER DETAILS

Reserved bits in any register are undefined. In some cases, a physical but unused memory bit may be present.

Any SPI read operation from a reserved bit or register results in an unknown but deterministic readback value. Any SPI write operation to a reserved bit or register results in no action.

A write to a control bit or control register defined only on Channel 0 must be addressed to Channel 0. Any such write that is addressed

to only Channel 1 is ignored if no register or control bit is defined on Channel 1.

Furthermore, any such write that is addressed to both Channel 0 and Channel 1 (as a multichannel write) proceeds as if the write is addressed to both Channel 0 and Channel 1. If no register or control bit is defined at Channel 1, data addressed to the undefined Channel 1 is ignored. If a register or control bit is defined at Channel 1, it is filled as part of the multichannel write.

SPI REGISTER MEMORY MAP AND DETAILS

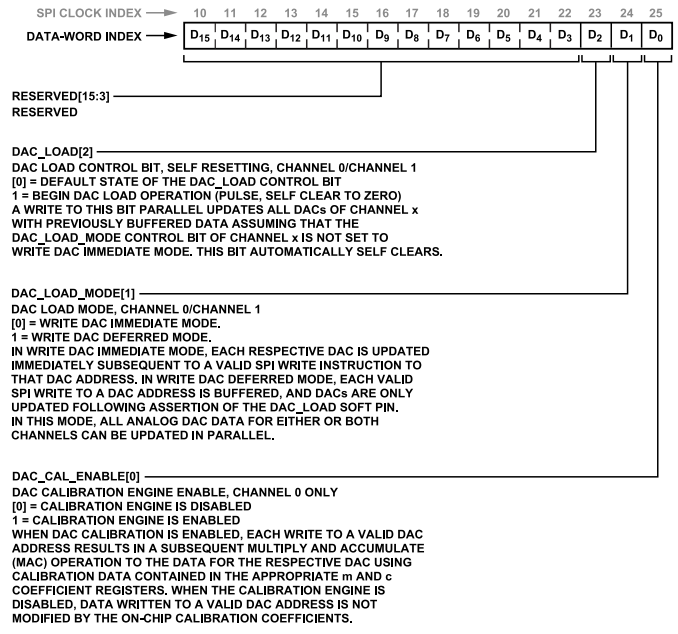


Figure 145. DAC Control Register (Address 0x11)

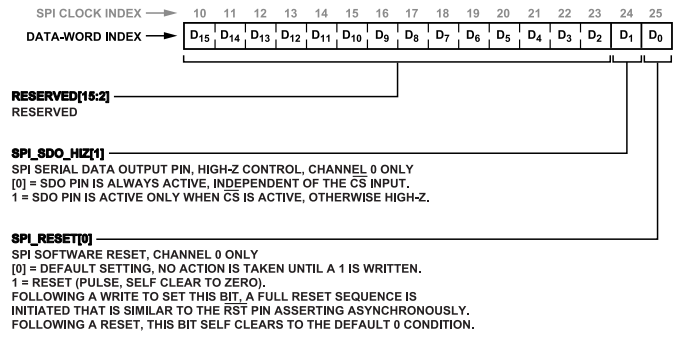


Figure 146. SPI Control Register (Address 0x12)

SPI REGISTER MEMORY MAP AND DETAILS

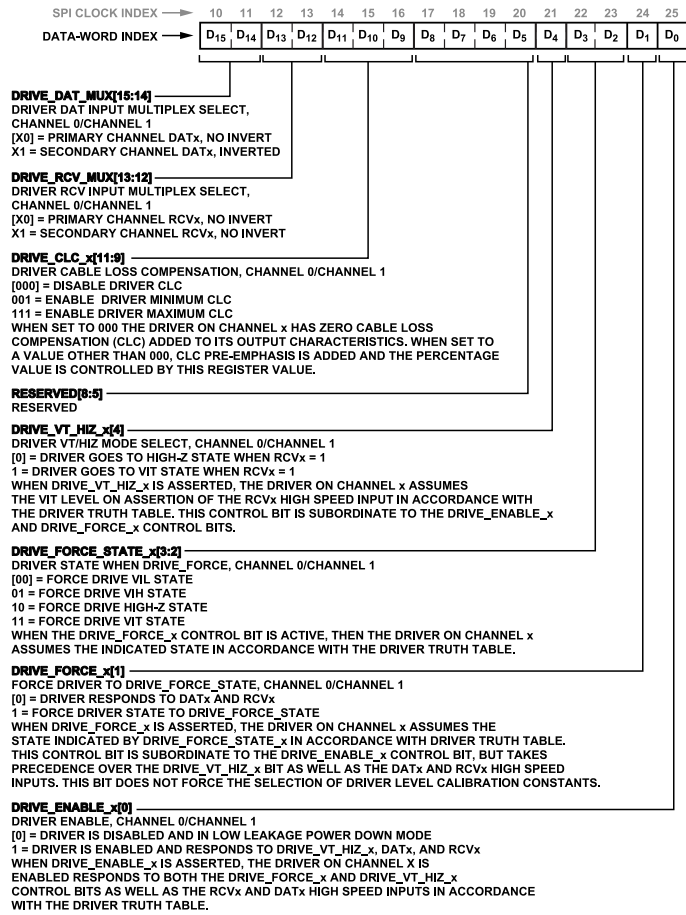


Figure 147. DRV Control Register (Address 0x19)



SPI REGISTER MEMORY MAP AND DETAILS

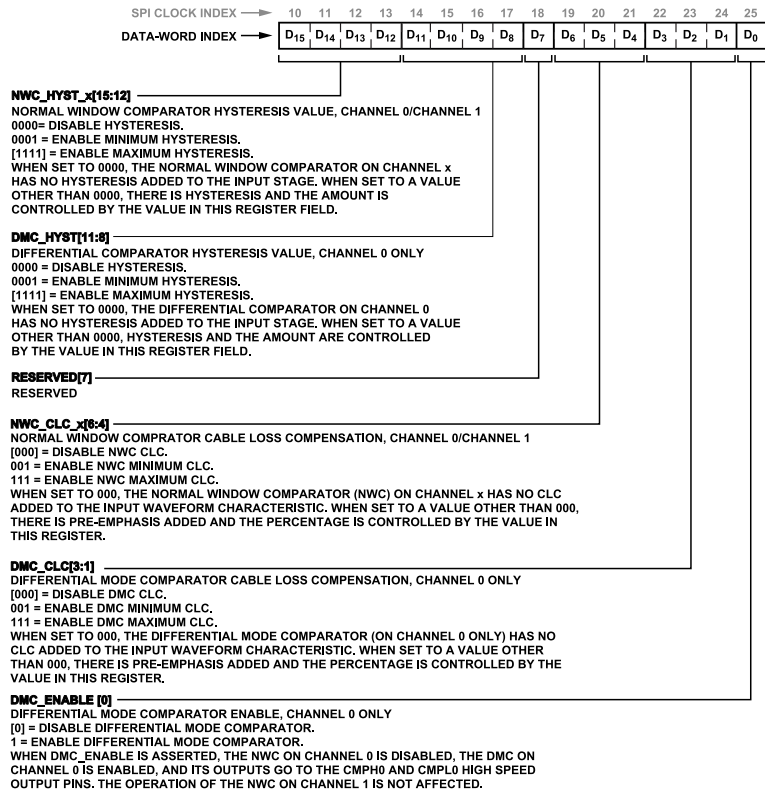


Figure 148. CMP Control Register (Address 0x1A)

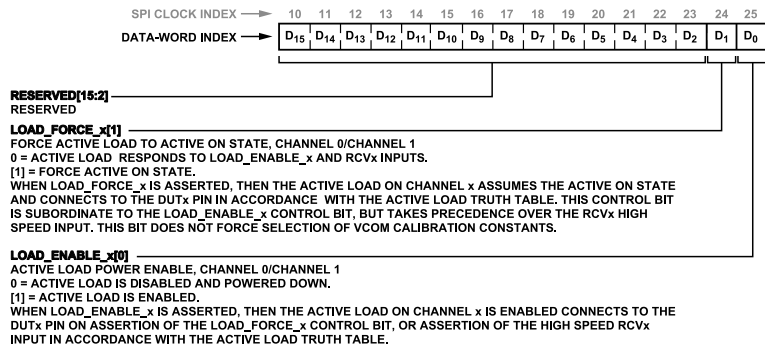


Figure 149. Load Control Register (Address 0x1B)

SPI REGISTER MEMORY MAP AND DETAILS

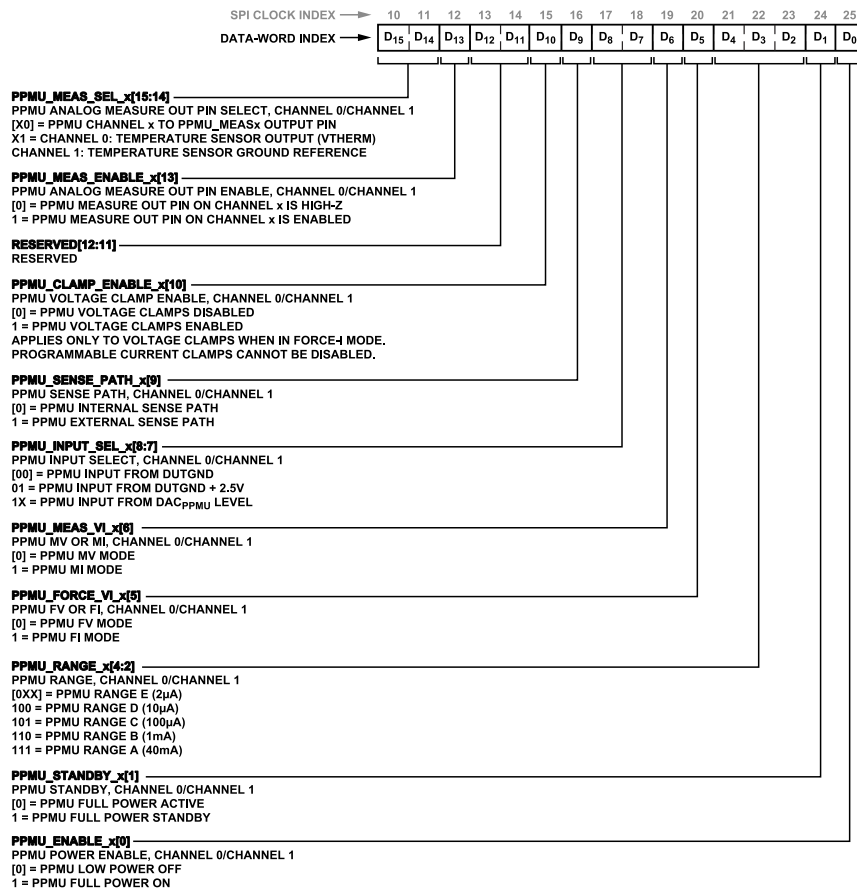


Figure 150. PPMU Control Register (Address 0x1C)

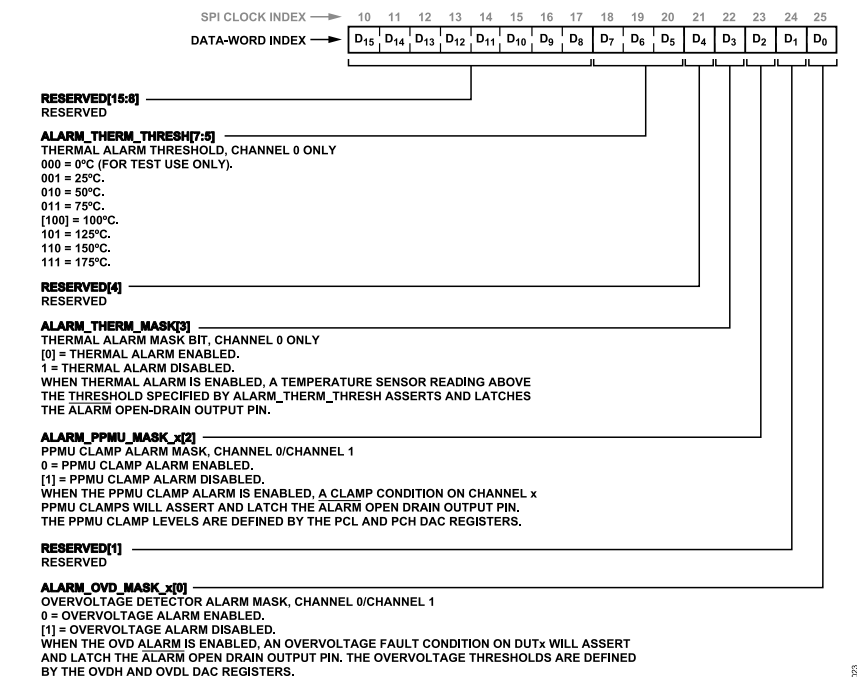


Figure 151. Alarm Mask Register (Address 0x1D)

SPI REGISTER MEMORY MAP AND DETAILS

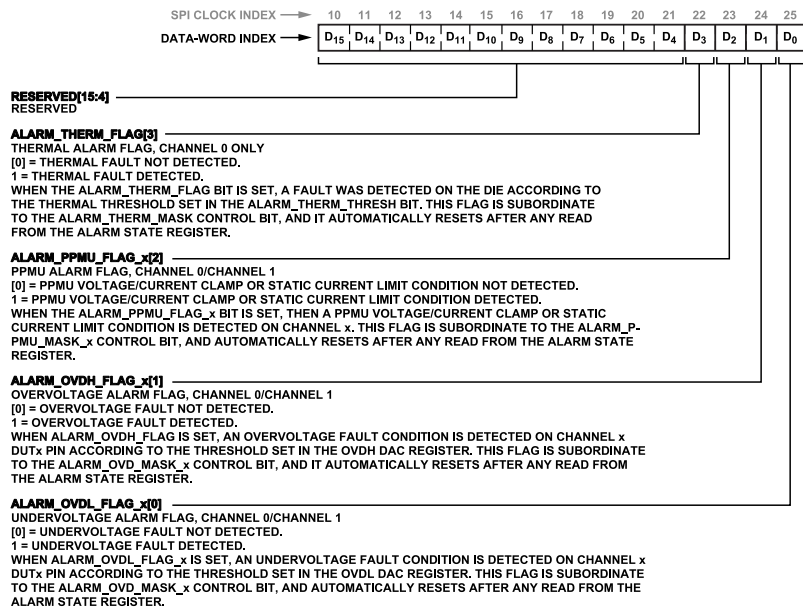


Figure 152. Alarm State Register (Address 0x1E) (Read Only)

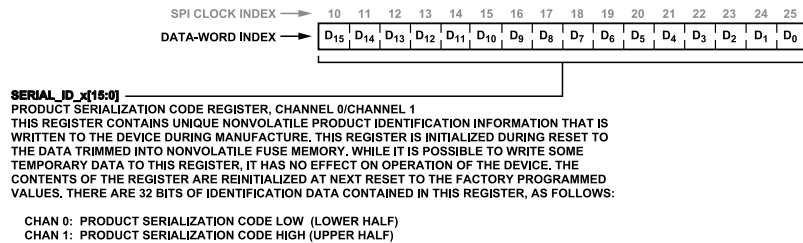


Figure 153. Product Serialization Code Register (Address 0x1F)

## DEFAULT TEST CONDITIONS

Table 30. Default Test Conditions

Name	SPI Address	Default Test Condition	Description
VIHx DAC Levels	Address 0x01[x]	2.0 V	
VITx/VCOMx DAC Levels	Address 0x02[x]	1.0 V	
VILx DAC Levels	Address 0x03[x]	0.0 V	
VOHx DAC Levels	Address 0x06[x]	5.0 V	
VOLx DAC Levels	Address 0x07[x]	-2.0 V	
POHx DAC Levels	Address 0x0D[x]	5.5 V	
POLx DAC Levels	Address 0x0E[x]	-2.0 V	
VCHx DAC Levels	Address 0x04[x]	5.0 V	
VCLx DAC Levels	Address 0x05[x]	-2.0 V	
PCHx DAC Levels	Address 0x0B[x]	7.0 V	
PCLx DAC Levels	Address 0x0C[x]	-2.0 V	
VIOHx DAC Levels	Address 0x08[x]	0.0 mA	
VIOLx DAC Levels	Address 0x09[x]	0.0 mA	
PPMux DAC Levels	Address 0x0A[x]	0.0 V	
OVDH DAC Level	Address 0x0F[1]	5.0 V	
OVDL DAC Level	Address 0x0F[0]	-2.0 V	
DAC Control Register	Address 0x11[0]	0x0000	DAC calibration disabled, DAC load mode is immediate
SPI Control Register	Address 0x12[1]	0x0000	SDO pin is always active, independent of $\overline{CS}$ state
DRV Control Registers	Address 0x19[x]	0x0000	Driver disabled in low leakage mode, DATx/RCVx inputs are multiplexed to primary channels, CLC is off, driver responds high-Z to RCVx inputs when enabled
CMP Control Registers	Address 0x1A[x]	0x0000	Normal window comparator mode, CLC is off, hysteresis is off
LOAD Control Registers	Address 0x1B[x]	0x0000	Active load is disabled and in power-down mode
PPMU Control Registers	Address 0x1C[x]	0x0000	PPMU disabled and in power-down mode, mode set FVMV Range E, input select $V_{DUTGND}$ internal sense path to $V_{DUTx}$ , PPMU_Mx pins high-Z, clamps disabled
ALARM Mask Registers	Address 0x1D[x]	0x0085	Disable PPMU and overvoltage detector alarm functions
Calibration m Coefficients	Not applicable	1.0 (0xFFFF)	
Calibration c Coefficients	Not applicable	0.0 (0x8000)	
DATx, RCVx Inputs	Not applicable	Static low	
SCLK Input	Not applicable	Static low	
DUTx Pins	Not applicable	Unterminated	
CMPHx, CMLPx Outputs	Not applicable	Unterminated	
$V_{DUTGND}$	Not applicable	0.0 V	

## EXTERNAL COMPONENTS

In addition to the external components identified in [Table 31](#) and [Table 32](#), see the [Power Supply, Grounding, and Typical Decoupling](#)

[Strategy](#) section for further information about recommended power supply decoupling capacitors.

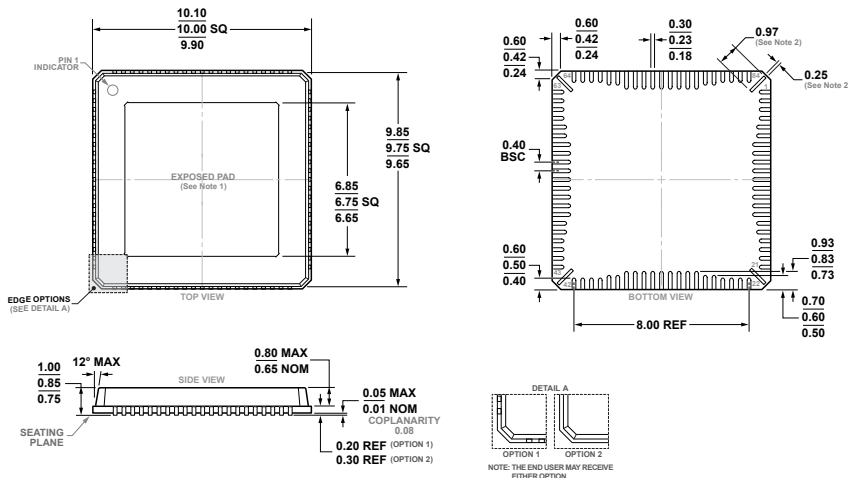
**Table 31. PPMU External Compensation Capacitors**

External Components Value (pF)	Location
1000 pF	Between the CFFB0 and CFFA0 pins
1000 pF	Between the CFFB1 and CFFA1 pins

**Table 32. Other External Components**

External Components Value (kΩ)	Location
10 kΩ	ALARM pull-up resistor to VDD
1 kΩ	BUSY pull-up resistor to VDD

OUTLINE DIMENSIONS



NOTES:  
 1. FOR PROPER CONNECTION OF THE EXPOSED PAD, REFER TO THE PIN CONFIGURATION AND FUNCTION DESCRIPTIONS SECTION OF THIS DATA SHEET.  
 2. TIEBARS MAY OR MAY NOT BE SOLDERED TO THE BOARD.

**Figure 154. 84-Lead Lead Frame Chip Scale Package [LFCSP]  
 10 mm x 10 mm Body and 0.85 mm Package Height  
 (CP-84-2)  
 Dimensions shown in millimeters**

Updated: August 09, 2023

ORDERING GUIDE

Model <sup>1</sup>	Temperature Range	Package Description	Package Option
ADATE320-1KCPZ	+25°C to +75°C	84-Lead LFCSP (10mm x 10mm w/ EP)	CP-84-2
ADATE320KCPZ	+25°C to +75°C	84-Lead LFCSP (10mm x 10mm w/ EP)	CP-84-2

<sup>1</sup> Z = RoHS Compliant Part.

EVALUATION BOARDS

Model <sup>1, 2</sup>	Description
EVAL-ADATE320EBZ	Evaluation Board

<sup>1</sup> Z = RoHS Compliant Part.

<sup>2</sup> The EVAL-ADATE320EBZ has an ADATE320-1KCPZ by default (part is soldered down).