

Green Mode Power Switch

FSL206MR

Description

The FSL206MR integrated Pulse-Width Modulator (PWM) and SENSEFET® is specifically designed for high-performance offline Switched-Mode Power Supplies (SMPS) while minimizing external components. This device integrates high-voltage power regulators that combine an avalanche-rugged SENSEFET with a Current-Mode PWM control block.

The integrated PWM controller includes: a 7.8 V regulator, eliminating the need for auxiliary bias winding; Under-Voltage Lockout (UVLO) protection; Leading-Edge Blanking (LEB); an optimized gate turn-on/turn-off driver; EMI attenuator; Thermal Shutdown (TSD) protection; temperature-compensated precision current sources for loop compensation; soft-start during startup; and fault-protection circuitry such as Overload Protection (OLP), Over-Voltage Protection (OVP), Abnormal Over-Current Protection (AOCP), and Line Under-Voltage Protection (LUV).

The internal high-voltage startup switch and the Burst-Mode operation with very low operating current reduce the power loss in Standby Mode. As a result, it is possible to reach a power loss of 150 mW with no bias winding and 25 mW (for FSL206MR) or 30 mW (for FSL206MRBN) with a bias winding under no-load conditions when the input voltage is 265 Vac.

Features

- Internal Avalanche-Rugged SENSEFET 650 V
- Precision Fixed Operating Frequency: 67 kHz
- No-Load < 150 mW at 265 Vac without Bias Winding; < 25 mW with Bias Winding for FSL206MR, < 30 mW with Bias Winding for FSL206MRBN
- No Need for Auxiliary Bias Winding
- Frequency Modulation for Attenuating EMI
- Line Under-Voltage Protection (LUV)
- Pulse-by-Pulse Current Limiting
- Low Under-Voltage Lockout (UVLO)
- Ultra-Low Operating Current: 300 µA
- Built-In Soft-Start and Startup Circuit
- Various Protections: Overload Protection (OLP), Over-Voltage Protection (OVP), Thermal Shutdown (TSD), Abnormal Over-Current Protection (AOCP) Auto-Restart Mode for All Protections

Applications

- SMPS for STB, DVD & DVCD Players
- SMPS for Auxiliary Power

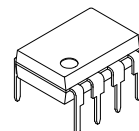
Related Resources

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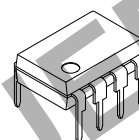


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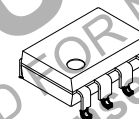
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**PDIP8 9.42x6.38, 2.54P
CASE 646CM**



**PDIP8 9.59x6.6, 2.54P
CASE 646CN**



**PDIP8 GW
CASE 709AJ**

MARKING DIAGRAM

\$Y&E&Z&2&K
FSL206MR

\$Y	= ON Semiconductor Logo
&E	= Designated Space
&Z	= Assembly Plant Code
&2	= 2-Digit Date code format
&K	= 2-Digits Lot Run Traceability Code
FSL206MR	= Specific Device Code Data

\$Y&Z&2&K
L206MRB

\$Y	= ON Semiconductor Logo
&Z	= Assembly Plant Code
&2	= 2-Digit Date code format
&K	= 2-Digits Lot Run Traceability Code
L206MRB	= Specific Device Code Data

ORDERING INFORMATION

See detailed ordering and shipping information on page 2 of this data sheet.

ORDERING INFORMATION

Part Number	Operating Temperature	Top Mark	PKG	Packing Method	Output Power Table (Note 1)			
					Current Limit	$R_{DS(ON),MAX}$	230 Vac $\pm 15\%$ (Note 2)	85 ~265 Vac
							Open Frame (Note 3)	Open Frame (Note 3)
FSL206MRN	-40 ~ 115°C	FSL206MR	8-DIP	Tube	0.6 A	19 Ω	12 W	7 W
FSL206MRBN		L206MRB						
FSL206MRL		FSL206MR	8-LSOP	Tube				
FSL206MRLX				Tape & Reel				

1. The junction temperature can limit the maximum output power.
2. 230 Vac or 100/115 Vac with doubler. The maximum power with CCM operation.
3. Maximum practical continuous power in an open-frame design at 50°C ambient.

APPLICATION DIAGRAM

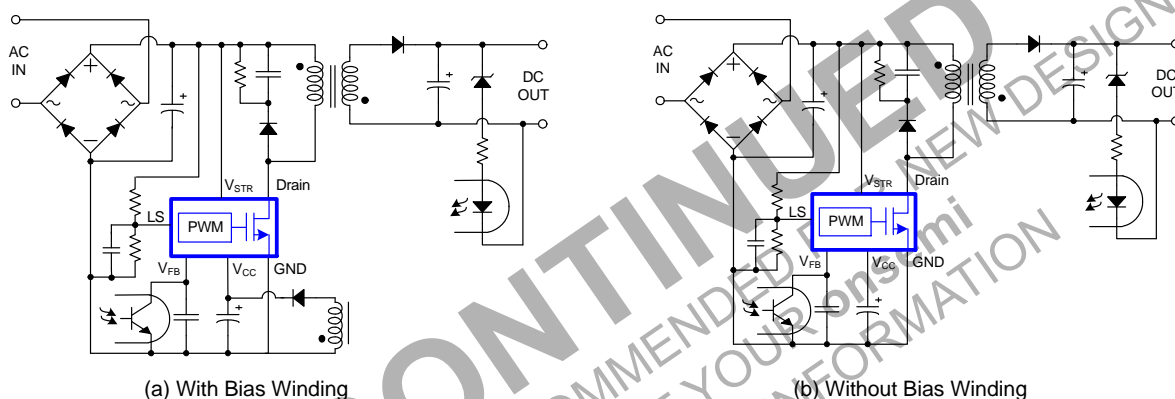


Figure 1. Typical Application

INTERNAL BLOCK DIAGRAM

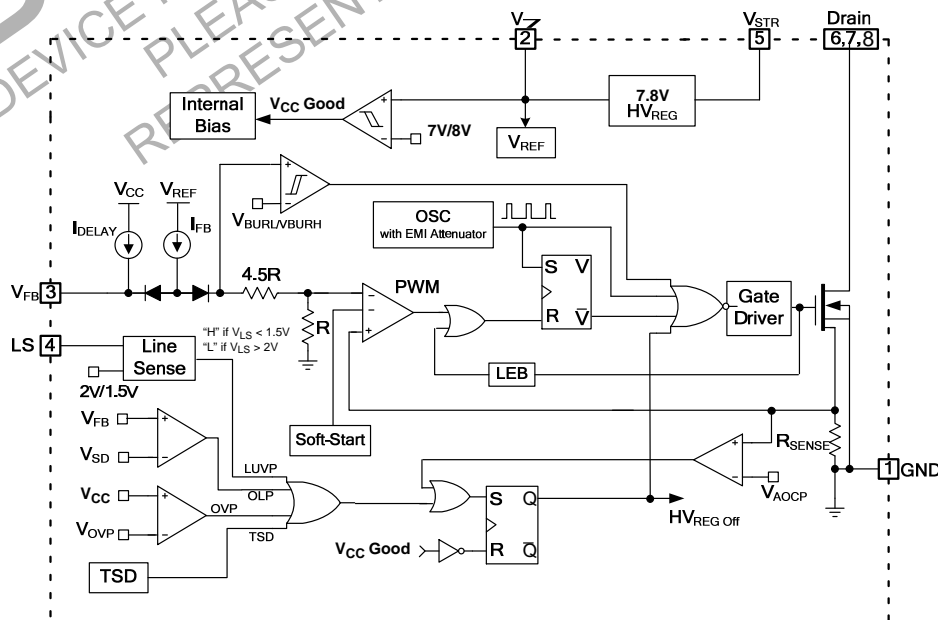


Figure 2. Internal Block Diagram

FSL206MR

PIN CONFIGURATION

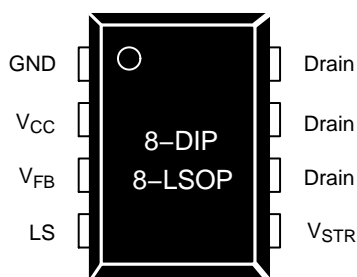


Figure 3. Pin Configuration

PIN DEFINITIONS

Pin No.	Name	Description
1	GND	Ground. SENSEFET source terminal on the primary side and internal control ground.
2	VCC	Positive Supply Voltage Input. Although connected to an auxiliary transformer winding, current is supplied from pin 5 (VSTR) via an internal switch during startup (see Figure 2). Once VCC reaches the UVLO upper threshold (12 V), the internal startup switch opens and device power is supplied via the auxiliary transformer winding.
3	VFB	Feedback Voltage. Non-inverting input to the PWM comparator, with a 0.11 mA current source connected internally and a capacitor and opto-coupler typically connected externally. There is a delay while charging external capacitor CFB from 2.4 V to 5 V using an internal 2.7 μ A current source. This delay prevents false triggering under transient conditions, but allows the protection mechanism to operate under true overload conditions.
4	LS	Line Sense Pin This pin is used to protect the device when the input voltage is lower than the rated input voltage range. If this pin is not used, connect to ground.
5	VSTR	Startup. Connected to the rectified AC line voltage source. At startup, the internal switch supplies internal bias and charges an external storage capacitor placed between the VCC pin and ground. Once VCC reaches 8 V, all internal blocks are activated. After that, the internal high-voltage regulator (HV REG) turns on and off irregularly to maintain VCC at 7.8 V.
6, 7, 8	Drain	Drain. Designed to connect directly to the primary lead of the transformer and capable of switching a maximum of 650 V. Minimizing the length of the trace connecting these pins to the transformer decreases leakage inductance.

FSL206MR

ABSOLUTE MAXIMUM RATINGS (T_A = 25°C unless otherwise specified)

Symbol	Parameter	Min	Max	Unit
V _{STR}	V _{STR} Pin Voltage	-0.3	650	V
V _{DS}	Drain Pin Voltage	-0.3	650	V
V _{CC}	Supply Voltage	-	26	V
V _{LS}	LS Pin Voltage	-	Internally Clamped Voltage (Note 4)	V
V _{FB}	Feedback Voltage Range	-0.3	Internally Clamped Voltage (Note 4)	V
I _{DM}	Drain Current Pulsed (Note 5)	-	1.5	A
E _{AS}	Single-Pulsed Avalanche Energy (Note 6)	-	11	mJ
P _D	Total Power Dissipation	-	1.3	W
T _J	Operating Junction Temperature	-40	+150	°C
T _A	Operating Ambient Temperature	-40	+125	°C
T _{STG}	Storage Temperature	-55	+150	°C
ESD	Human Body Model, JESD22-A114	-	4	kV
	Charged Device Model, JESD22-C101	-	2	

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

4. V_{FB} is clamped by internal clamping diode (13 V I_{CLAMP_MAX} < 100 μA). After Shutdown, before V_{CC} reaching V_{STOP}, V_{SD} < V_{FB} < V_{CC}.
5. Repetitive rating; pulse-width limited by maximum junction temperature.
6. L = 21 mH, starting T_J = 25°C

THERMAL IMPEDANCE (T_A = 25°C unless otherwise specified)

Symbol	Parameter	Value	Unit
θ _{JA}	Junction-to-Ambient Thermal Impedance (Note 7)	93	°C/W

7. JEDEC recommended environment, JESD51-2 and test board, JESD51-10 with minimum land pattern for 8DIP and JESD51-3 with minimum land pattern for 8LSOP.

ELECTRICAL CHARACTERISTICS (T_A = 25°C unless otherwise specified)

Symbol	Parameter	Test Condition	Min	Typ	Max	Unit
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SENSEFET SECTION

BV _{DSS}	Drain-Source Breakdown Voltage	V _{CC} = 0 V, I _D = 250 μA	650	-	-	V
I _{DSS}	Zero Gate Voltage Drain Current	V _{DS} = 650 V, V _{GS} = 0 V	-	-	50	μA
		V _{DS} = 520 V, V _{GS} = 0 V, T _A = 125°C (Note 8)	-	-	250	μA
R _{DS(ON)}	Drain-Source On-State Resistance (Note 9)	V _{GS} = 10 V, I _D = 0.3 A	-	14	19	Ω
C _{ISS}	Input Capacitance	V _{GS} = 0 V, V _{DS} = 25 V, f = 1 MHz	-	162	-	pF
C _{OSS}	Output Capacitance	V _{GS} = 0 V, V _{DS} = 25 V, f = 1 MHz	-	14.9	-	pF
C _{RSS}	Reverse Transfer Capacitance	V _{GS} = 0 V, V _{DS} = 25 V, f = 1 MHz	-	2.7	-	pF
t _r	Rise Time	V _{DS} = 325 V, I _D = 0.5 A, R _G = 25 Ω	-	6.1	-	ns
t _f	Fall Time	V _{DS} = 325 V, I _D = 0.5 A, R _G = 25 Ω	-	43.6	-	ns

CONTROL SECTION

f _{OSC}	Switching Frequency	V _{FB} = 4 V, V _{CC} = 10 V	61	67	73	KHz
Δf _{OSC}	Switching Frequency Variation	-25°C < T _J < 85°C	-	±5	±10	%
f _M	Frequency Modulation (Note 8)		-	±3	-	kHz
D _{MAX}	Maximum Duty Cycle	V _{FB} = 4 V, V _{CC} = 10 V	66	72	78	%
D _{MIN}	Minimum Duty Ratio	V _{FB} = 0 V, V _{CC} = 10 V	0	0	0	%

FSL206MR

ELECTRICAL CHARACTERISTICS (T_A = 25°C unless otherwise specified) (continued)

Symbol	Parameter	Test Condition	Min	Typ	Max	Unit
V _{START}	UVLO Threshold Voltage	V _{FB} = 0 V, V _{CC} Sweep	7	8	9	V
V _{STOP}		After Turn-on	6	7	8	V
I _{FB}	Feedback Source Current	V _{FB} = 0, V _{CC} = 10 V	90	110	130	μA
t _{S/S}	Internal Soft-Start Time	V _{FB} = 4 V, V _{CC} = 10 V	10	15	20	ms

BURST-MODE SECTION

V _{BURH}	Burst-Mode HIGH Threshold Voltage	V _{CC} = 10 V V _{FB} Increase	FSL206MR	0.66	0.83	1.00	V
			FSL206MRB	0.40	0.50	0.60	V
V _{BURL}	Burst-Mode LOW Threshold Voltage	V _{CC} = 10 V V _{FB} Decrease	FSL206MR	0.59	0.74	0.89	V
			FSL206MRB	0.28	0.35	0.42	V
HYS _{BUR}	Burst-Mode Hysteresis		FSL206MR	–	90	–	mV
			FSL206MRB	–	150	–	mV

PROTECTION SECTION

I _{LIM}	Peak Current Limit	V _{FB} = 4 V, di/dt = 300 mA/μs, V _{CC} = 10 V	0.54	0.60	0.66	A
t _{CLD}	Current Limit Delay Time (Note 8)		–	100	–	ns
V _{SD}	Shutdown Feedback Voltage	V _{CC} = 10 V	4.5	5.0	5.5	V
I _{DELAY}	Shutdown Delay Current	V _{FB} = 4 V	2.1	2.7	3.3	μA
t _{LEB}	Leading Edge Blanking Time (Note 8)		250	–	–	ns
V _{AOCP}	Abnormal Over-Current Protection (Note 8)		–	0.7	–	V
V _{OVP}	Over-Voltage Protection	V _{FB} = 4 V, V _{CC} Increase	23.0	24.5	26.0	V
V _{LS_OFF}	Line-Sense Protection On to Off	V _{FB} = 3 V, V _{CC} = 10 V, V _{LS} Decrease	1.9	2.0	2.1	V
V _{LS_ON}	Line-Sense Protection Off to On	V _{FB} = 3 V, V _{CC} = 10 V, V _{LS} Increase	1.4	1.5	1.6	V
TSD	Thermal Shutdown Temperature (Note 8)		125	135	150	°C
HYSTSD	TSD Hysteresis Temperature (Note 8)		–	60	–	°C

HIGH VOLTAGE REGULATOR SECTION

H _{HVR}	HV Regulator Voltage	V _{FB} = 0 V, V _{STR} = 40 V	–	7.8	–	V
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TOTAL DEVICE SECTION

I _{OP1}	Operating Supply Current (Control Part Only, without Switching)	V _{CC} = 15 V, 0 V < V _{FB} < V _{BURL}	–	0.3	0.5	mA
I _{OP2}	Operating Supply Current (Control Part Only, without Switching)	V _{CC} = 8 V, 0 V < V _{FB} < V _{BURL}	–	0.25	0.45	mA
I _{OP3}	Operating Supply Current (Note 8) (While Switching)	V _{CC} = 15 V, V _{BURL} < V _{FB} < V _{SD}	–	–	1.3	mA
I _{CH}	Startup Charging Current	V _{CC} = 0 V, V _{STR} > 40 V	1.6	1.9	2.4	mA
I _{START}	Startup Current	V _{CC} = Before V _{START} , V _{FB} = 0 V	–	100	150	μA
V _{STR}	Minimum V _{STR} Supply Voltage	V _{CC} = V _{FB} = 0 V, V _{STR} Increase	–	26	–	V

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

8. Though guaranteed by design, it is not 100% tested in production.

9. Pulse test: pulse width = 300 ms, duty cycle = 2%.

TYPICAL PERFORMANCE CHARACTERISTICS

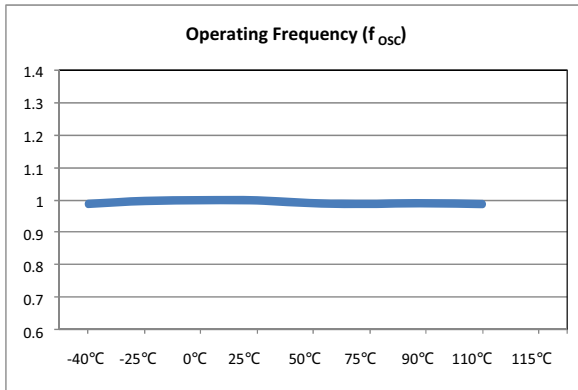


Figure 4. Operating Frequency vs. Temperature

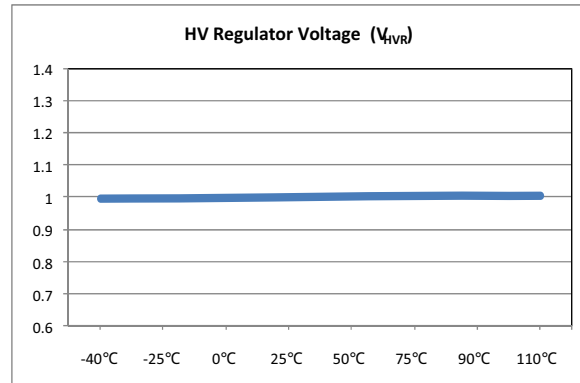


Figure 5. HV Regulator Voltage vs. Temperature

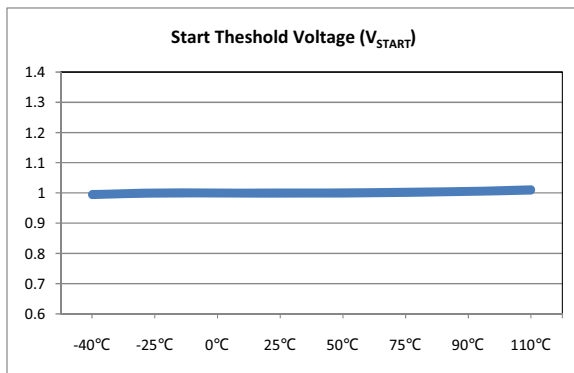


Figure 6. Start Threshold Voltage vs. Temperature

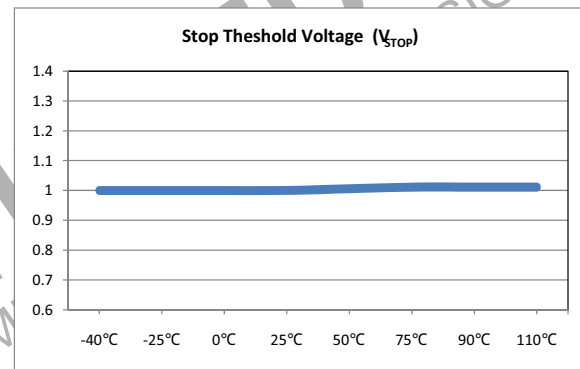


Figure 7. Stop Threshold Voltage vs. Temperature

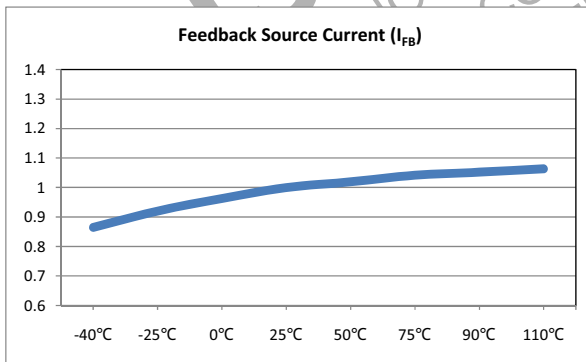


Figure 8. Feedback Source Current vs. Temperature

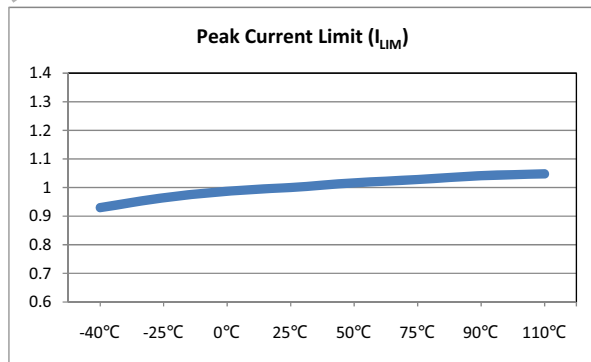


Figure 9. Peak Current Limit vs. Temperature

TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

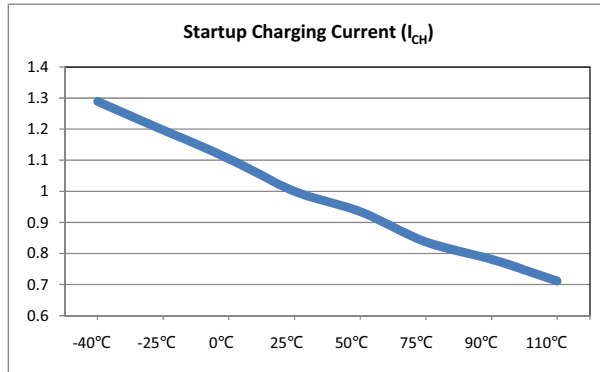
(These Characteristic graphs are normalized at $T_A = 25^\circ\text{C}$.)

Figure 10. Startup Charging Current vs. Temperature

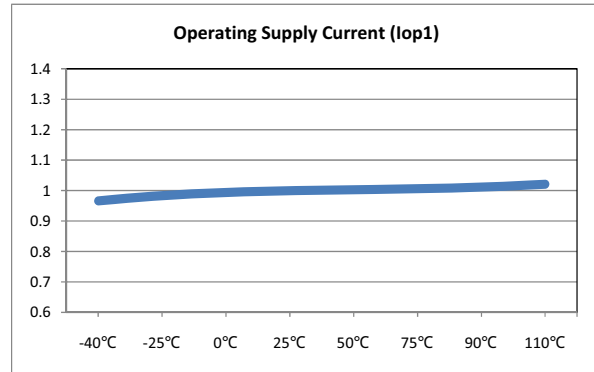


Figure 11. Operating Supply Current 1 vs. Temperature

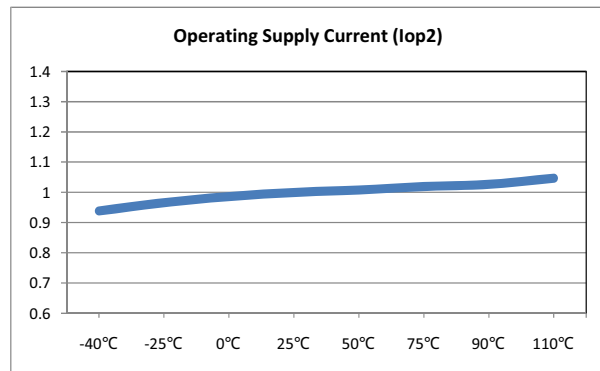


Figure 12. Operating Supply Current 2 vs. Temperature

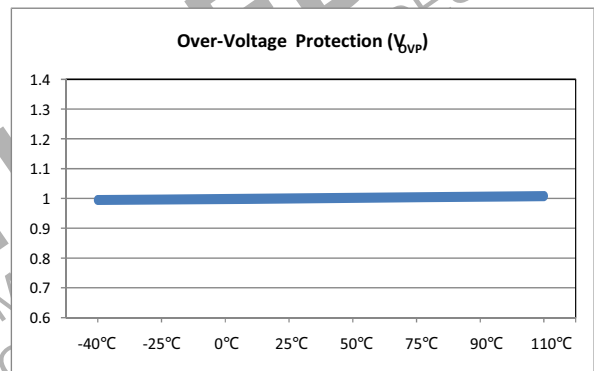


Figure 13. Over-Voltage Protection Voltage vs. Temperature

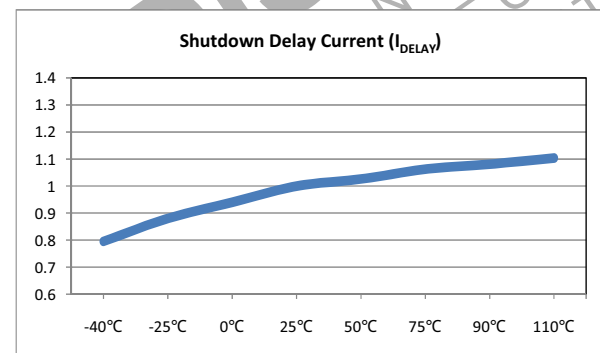


Figure 14. Shutdown Delay Current vs. Temperature

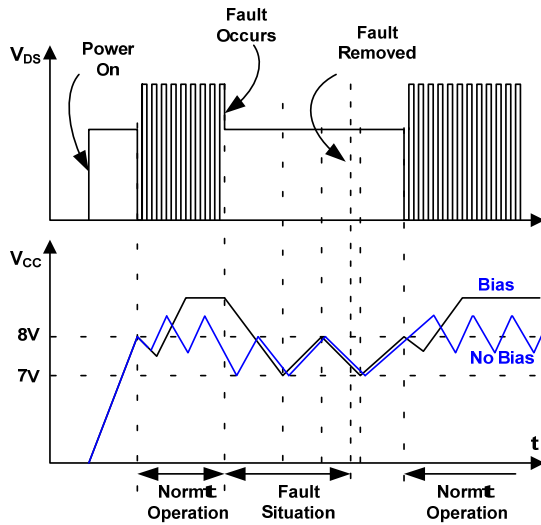


Figure 18. Auto-Restart Protection Waveforms

Overload Protection (OLP)

Overload is defined as the load current exceeding a preset level due to an unexpected event. In this situation, the protection circuit should be activated to protect the SMPS. However, even when the SMPS is operating normally, the overload protection (OLP) circuit can be activated during the load transition or startup. To avoid this undesired operation, the OLP circuit is activated after a specified time to determine whether it is a transient situation or a true overload situation. The Current-Mode feedback path limits the current in the SENSEFET when the maximum PWM duty cycle is attained. If the output consumes more than this maximum power, the output voltage (V_O) decreases below its rating voltage. This reduces the current through the opto-coupler LED, which also reduces the opto-coupler transistor current, increasing the feedback voltage (V_{FB}). If V_{FB} exceeds 2.4 V, the feedback input diode is blocked and the 2.7 μ A current source (I_{DELAY}) starts to charge C_{FB} slowly up. In this condition, V_{FB} increases until it reaches 5 V, when the switching operation is terminated, as shown in Figure 19. The shutdown delay is the time required to charge C_{FB} from 2.4 V to 5 V with 2.7 μ A current source.

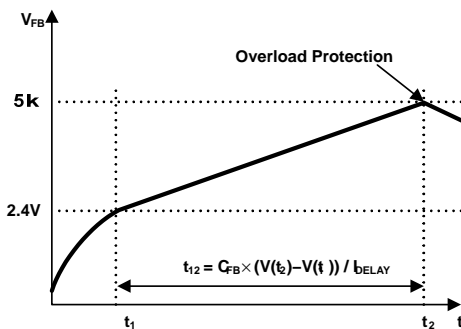


Figure 19. Overload Protection (OLP)

Abnormal Over-Current Protection (AOCP)

When the secondary rectifier diodes or the transformer pin are shorted, a steep current with extremely high di/dt can flow through the SENSEFET during the LEB time. Even though the power switch has OLP (Overload Protection), it is not enough to protect the FPS in that abnormal case, since severe current stress is imposed on the SENSEFET until OLP triggers. The power switch includes the internal AOCP (Abnormal Over-Current Protection) circuit shown in Figure 20. When the gate turn-on signal is applied to the power SENSEFET, the AOCP block is enabled and monitors the current through the sensing resistor. The voltage across the resistor is compared with a preset AOCP level. If the sensing resistor voltage is greater than the AOCP level, the set signal is applied to the latch, resulting in the shutdown of the SMPS.

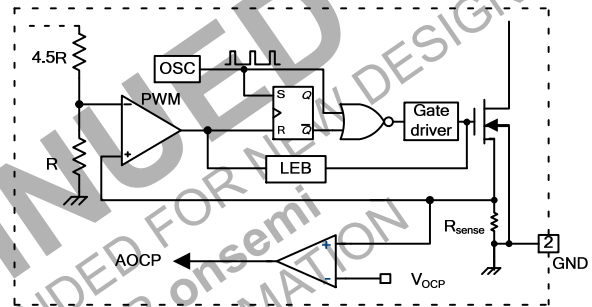


Figure 20. Abnormal Over-Current Protection

Thermal Shutdown (TSD)

The SENSEFET and control IC being integrated makes it easier to detect the temperature of the SENSEFET. When the junction temperature exceeds $\sim 135^\circ\text{C}$, thermal shutdown is activated and the power switch is restarted after temperature decreases to 60°C .

Over-Voltage Protection (OVP)

In the event of a malfunction in the secondary-side feedback circuit or an open feedback loop caused by a soldering defect, the current through the opto-coupler transistor becomes almost zero (refer to Figure 17). Then V_{FB} climbs up in a similar manner to the overload situation, forcing the preset maximum current to be supplied to the SMPS until the overload protection is activated. Because excess energy is provided to the output, the output voltage may exceed the rated voltage before the overload protection is activated, resulting in the breakdown of the devices in the secondary side. To prevent this situation, an over-voltage protection (OVP) circuit is employed. In general, V_{CC} is proportional to the output voltage and the FPS uses V_{CC} instead of directly monitoring the output voltage. If V_{CC} exceeds 24.5 V, OVP circuit is activated, resulting in termination of the switching operation. To avoid undesired activation of OVP during normal operation, V_{CC} should be designed to be below 24.5 V.

Line Under-Voltage Protection (LUVP)

If the input voltage of the converter is lower than the minimum operating voltage, the converter input current increases too much, causing components failure. If the input voltage is low, the converter should be protected. In the FSL206MR, the LUVP circuit senses the input voltage using the LS pin and, if this voltage is lower than 1.5 V, the LUVP signal is generated. The comparator has 0.5 V hysteresis. If the LUVP signal is generated, the output drive block is shut down and the output voltage feedback loop is saturated.

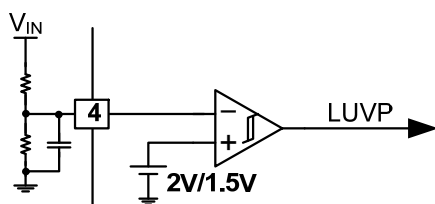


Figure 21. Line UVP Circuit

Soft-Start

The FSL206MR has an internal soft-start circuit that slowly increases the feedback voltage, together with the SENSEFET current, after it starts. The typical soft-start time is 15 ms, as shown in Figure 22, where progressive increments of the SENSEFET current are allowed during the startup phase. The pulse width to the power switching device is progressively increased to establish the correct working conditions for transformers, inductors, and capacitors. The voltage on the output capacitors is progressively increased with the intention of smoothly establishing the required output voltage. It also helps prevent transformer saturation and reduce the stress on the secondary diode.

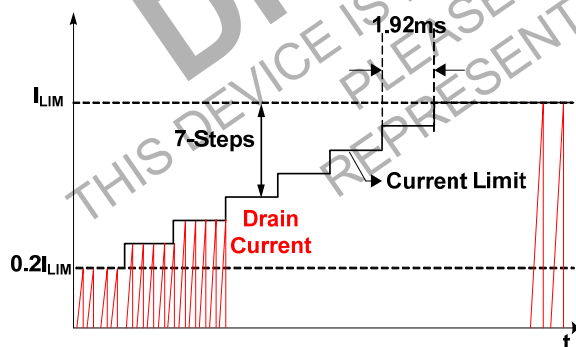


Figure 22. Internal Soft-Start

Burst Operation

To minimize power dissipation in Standby Mode, the power switch enters Burst Mode. As the load decreases, the feedback voltage decreases. As shown in Figure 23, the device automatically enters Burst Mode when the feedback voltage drops below V_{BURH} . Switching continues until the feedback voltage drops below V_{BURL} . At this point, switching stops and the output voltages start to drop at a rate dependent on the standby current load. This causes the feedback voltage to rise. Once it passes V_{BURH} , switching resumes. The feedback voltage then falls and the process repeats. Burst Mode alternately enables and disables switching of the SENSEFET and reduces switching loss in Standby Mode.

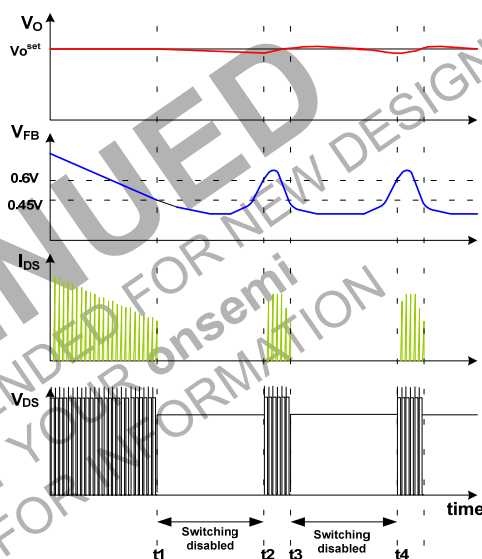


Figure 23. Burst-Mode Operation

MECHANICAL CASE OUTLINE

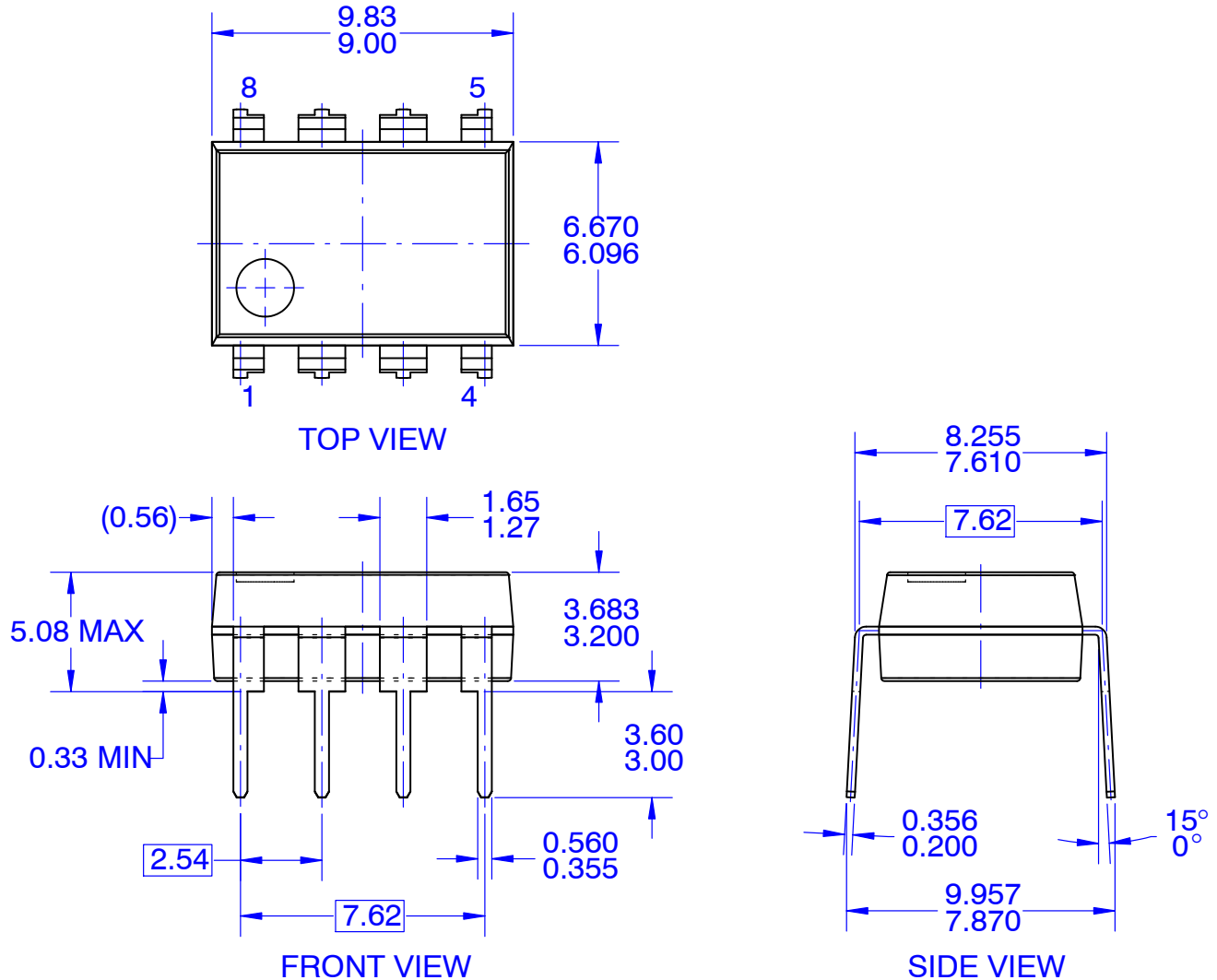
PACKAGE DIMENSIONS

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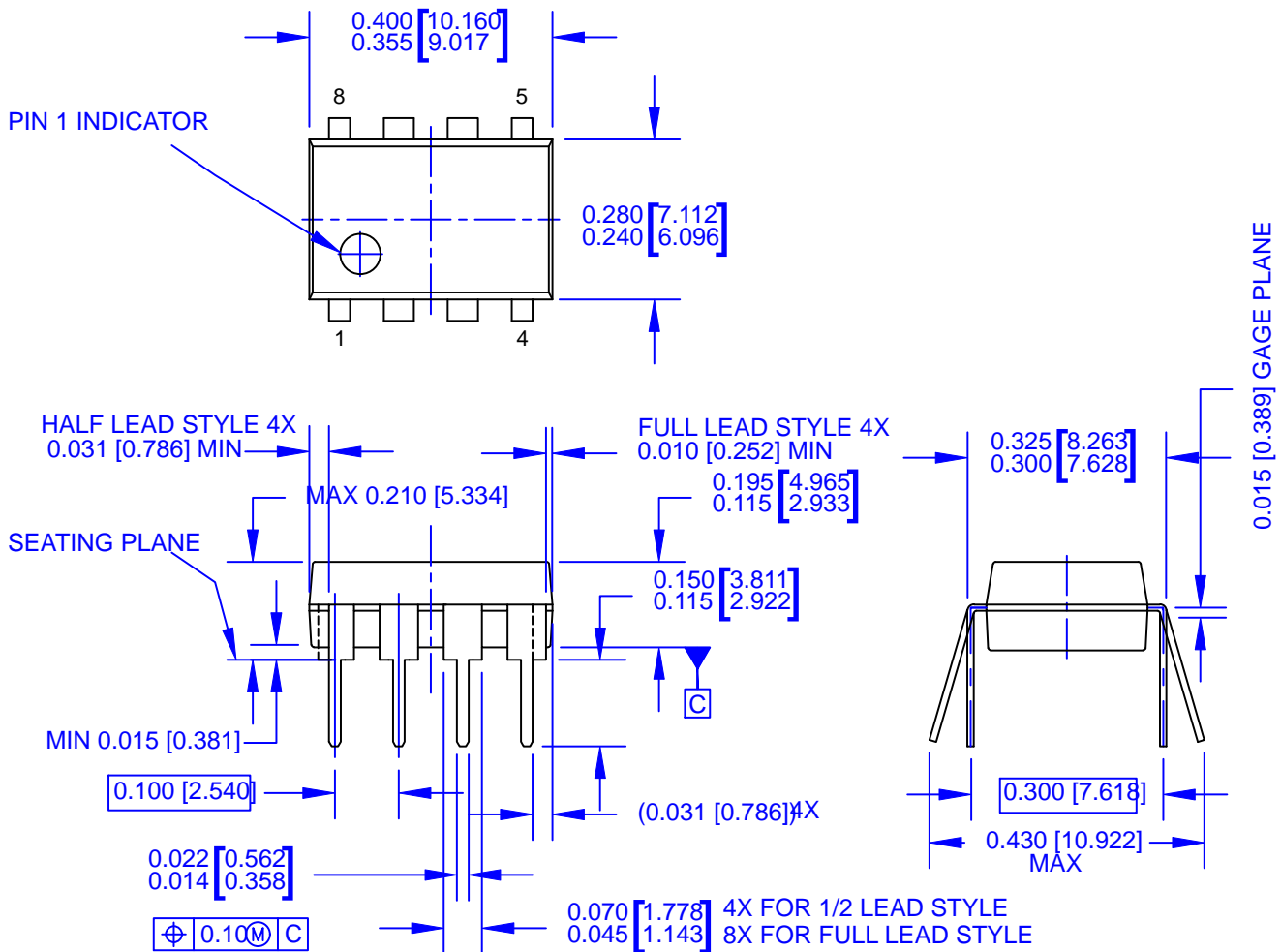
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- B. ALL DIMENSIONS ARE IN MILLIMETERS
- C. DIMENSIONS ARE EXCLUSIVE OF BURRS, MOLD FLASH, AND TIE BAR EXTRUSIONS
- D. DIMENSIONS AND TOLERANCES PER ASME Y14.5M-2009

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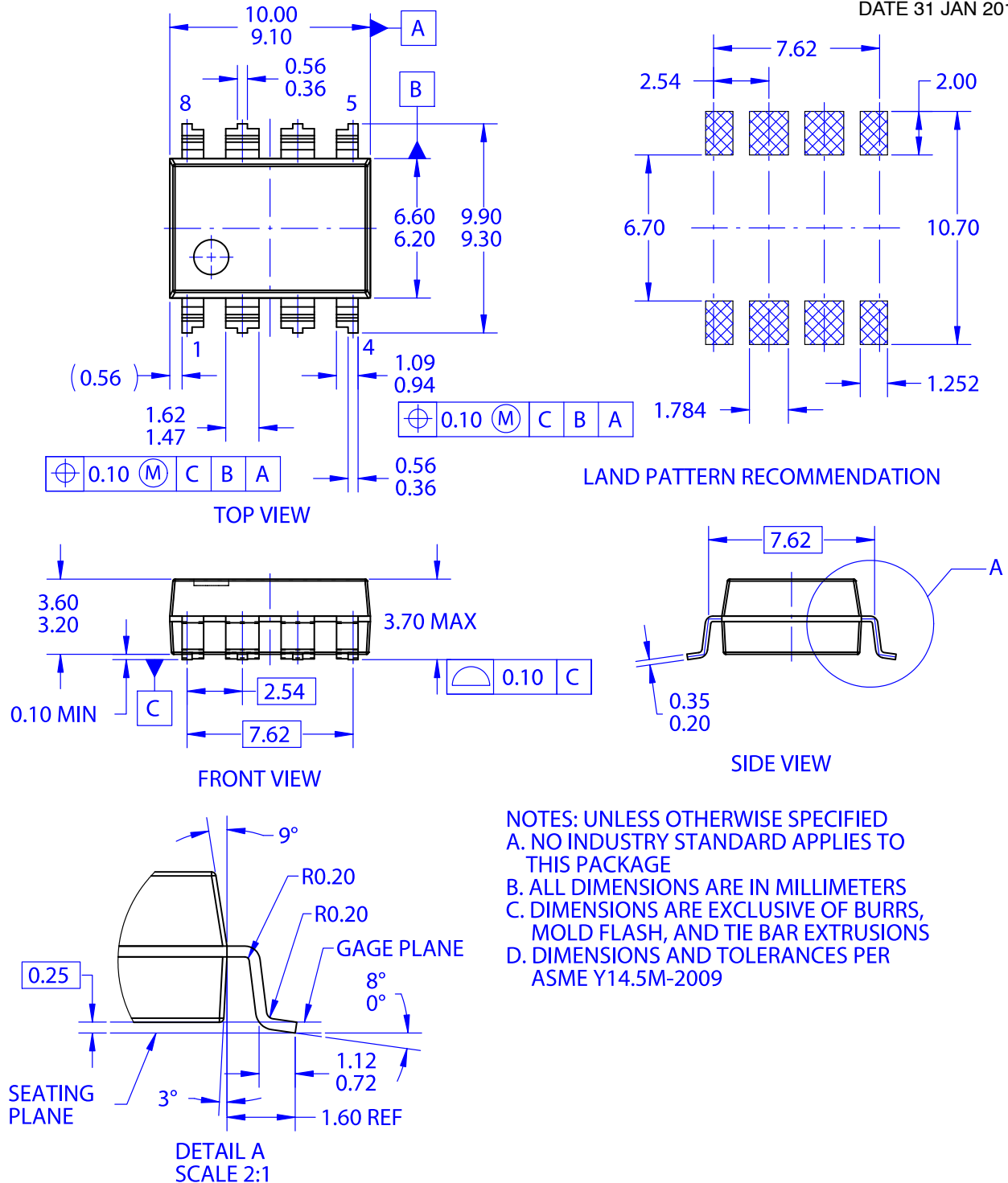
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- B) CONTROLLING DIMS ARE IN INCHES
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