

NUF6105FC

6-Channel EMI Filter with Integrated ESD Protection

The NUF6105FC is a six-channel (C–R–C) Pi-style EMI filter array with integrated ESD protection. Its typical component values of $R = 100\ \Omega$ and $C = 27\ \text{pF}$ deliver a cutoff frequency of 100 MHz and stop band attenuation greater than $-35\ \text{dB}$ from 800 MHz to 2.2 GHz.

This performance makes the part ideal for parallel interfaces with data rates up to 67 Mbps in applications where wireless interference must be minimized. The specified attenuation range is very effective in minimizing interference from 2G/3G, GPS, Bluetooth® and WLAN signals.

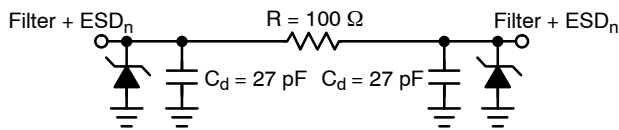
The NUF6105FC is available in the low-profile 12-lead 2.9 mm x 1.3 mm Flip-Chip surface mount package.

Features/Benefits

- $\pm 30\ \text{kV}$ ESD Protection on each channel (IEC61000-4-2 Level 4, Contact Discharge)
- $\pm 16\ \text{kV}$ ESD Protection on each channel (Human Body Model)
- R/C Values of $100\ \Omega$ and $27\ \text{pF}$ deliver Exceptional S21 Performance Characteristics of 100 MHz $f_{3\text{dB}}$ and $-35\ \text{dB}$ Stop Band Attenuation from 800 MHz to 2.2 GHz
- Integrated EMI/ESD System Solution in Flip-Chip Package Offers Exceptional Cost, System Reliability and Space Savings
- Pb-Free Package is Available*

Applications

- EMI Filtering for LCD and Camera Data Lines
- EMI Filtering and Protection for I/O Ports and Keypads



See Table 1 for pin description

Figure 1. Electrical Schematic

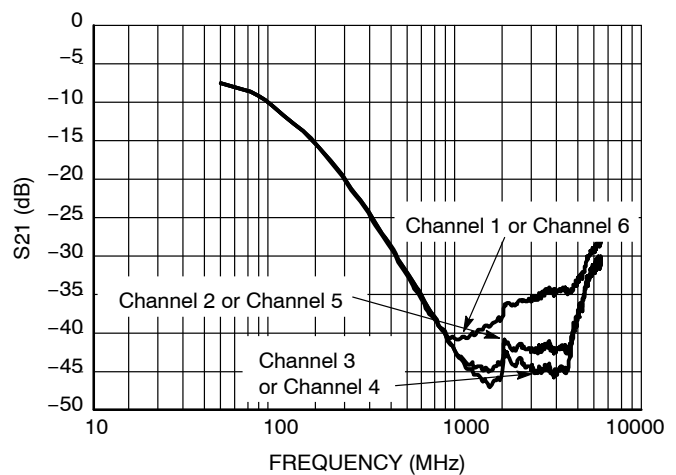


Figure 1. Insertion Loss Curve (S21 Measurement)



ON Semiconductor®

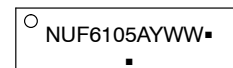
<http://onsemi.com>



A1

FLIP-CHIP
CASE 499D
PLASTIC

MARKING DIAGRAM



- A = Assembly Location
 - Y = Year
 - WW = Work Week
 - = Pb-Free Package
- (Note: Microdot may be in either location)

ORDERING INFORMATION

Device	Package	Shipping†
NUF6105FCT1	Flip-Chip	3000/Tape & Reel
NUF6105FCT1G	Flip-Chip (Pb-Free)	3000/Tape & Reel

†For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

*For additional information on our Pb-Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

NUF6105FC

Table 1. FUNCTIONAL PIN DESCRIPTION

Filter	Device Pins	Description
Filter 1	1 & 12	Filter + ESD Channel 1
Filter 2	2 & 11	Filter + ESD Channel 2
Filter 3	3 & 10	Filter + ESD Channel 3
Filter 4	4 & 9	Filter + ESD Channel 4
Filter 5	5 & 8	Filter + ESD Channel 5
Filter 6	6 & 7	Filter + ESD Channel 6
Ground Pad	GND	Ground

MAXIMUM RATINGS

Parameter	Symbol	Value	Unit
ESD Discharge IEC61000-4-2 Contact Discharge Machine Model Human Body Model	V_{PP}	30 30 16	kV
DC Power per Resistor	P_R	100	mW
DC Power per Package	P_T	600	mW
Junction Temperature	T_J	150	°C
Operating Temperature Range	T_{OP}	-40 to 85	°C
Storage Temperature Range	T_{STG}	-55 to 150	°C
Maximum Lead Temperature for Soldering Purposes (1.8 in from case for 10 seconds)	T_L	260	°C

Stresses exceeding Maximum Ratings may damage the device. Maximum Ratings are stress ratings only. Functional operation above the Recommended Operating Conditions is not implied. Extended exposure to stresses above the Recommended Operating Conditions may affect device reliability.

ELECTRICAL CHARACTERISTICS ($T_J = 25^\circ\text{C}$ unless otherwise noted)

Parameter	Symbol	Test Conditions	Min	Typ	Max	Unit
Maximum Reverse Working Voltage	V_{RWM}				5.0	V
Breakdown Voltage	V_{BR}	$I_R = 10\text{ mA}$	6.0	7.0	8.0	V
Leakage Current	I_R	$V_{RWM} = 3.3\text{ V}$			100	nA
Resistance	R_A	$I_R = 20\text{ mA}$	80	100	120	Ω
Diode Capacitance	C_d	$V_R = 2.5\text{ V}, f = 1.0\text{ MHz}$		27		pF
Line Capacitance	C_L	$V_R = 2.5\text{ V}, f = 1.0\text{ MHz}$		54		pF
3 dB Cut-Off Frequency (Note 1)	f_{3dB}	Above this frequency, appreciable attenuation occurs		100		MHz
6 dB Cut-Off Frequency (Note 1)	f_{6dB}	Above this frequency, appreciable attenuation occurs		180		MHz

1. 50 Ω source and 50 Ω load termination.

NUF6105FC

TYPICAL PERFORMANCE CURVES ($T_A = 25^\circ\text{C}$ unless otherwise specified)

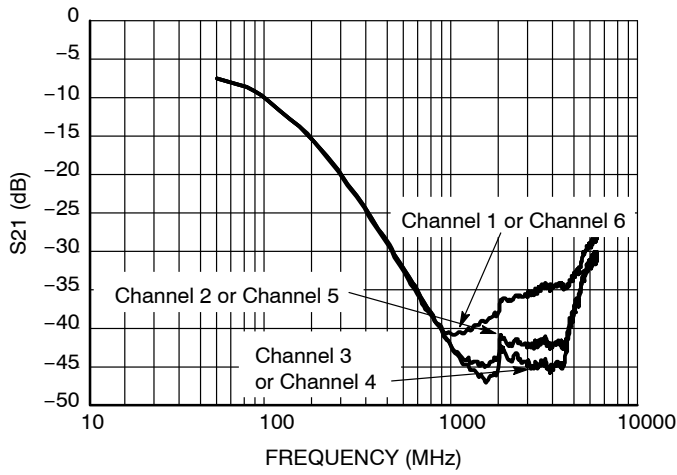


Figure 1. Insertion Loss Curve (S21 Measurement)

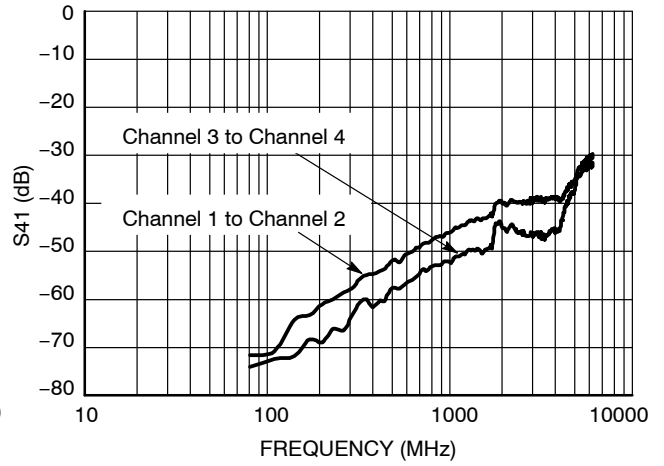


Figure 2. Analog Crosstalk Curve (S41 Measurement)

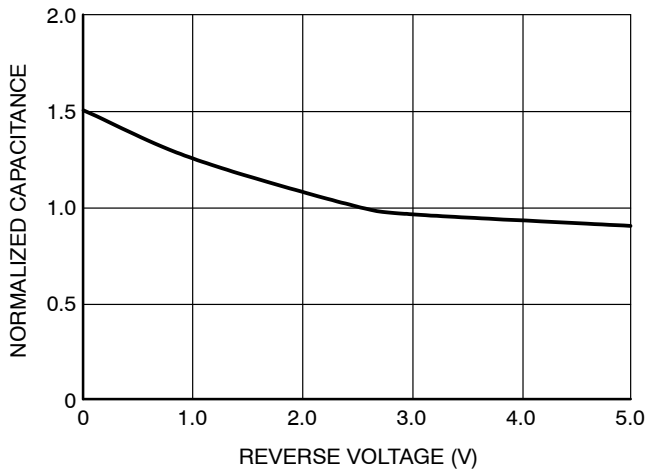


Figure 3. Typical Capacitance vs. Reverse Biased Voltage (Normalized Capacitance C_d at 2.5 V)

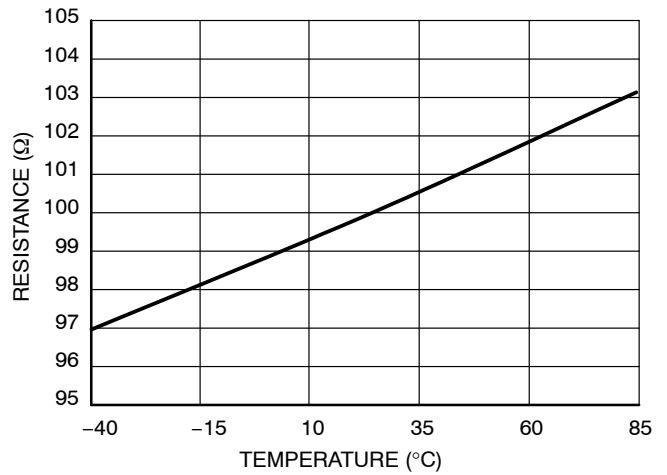


Figure 4. Resistance Over Temperature

Theory of Operation

The NUF6105FC combines ESD protection and EMI filtering conveniently into a small package for today’s size constrained applications. The capacitance inherent to a typical protection diode is utilized to provide the capacitance value necessary to create the desired frequency response based upon the series resistance in the filter. By combining this functionality into one device, a large number of discrete components are integrated into one small package saving valuable board space and reducing BOM count and cost in the application.

Application Example

The accepted practice for specifying bandwidth in a filter is to use the 3 dB cutoff frequency. Utilizing points such as the 6 dB or 9 dB cutoff frequencies results in signal degradation in an application. This can be illustrated in an application example. A typical application would include EMI filtering of data lines in a camera or display interface. In such an example it is important to first understand the signal and its spectral content. By understanding these things, an appropriate filter can be selected for the desired application. A typical data signal is pattern of 1’s and 0’s transmitted over a line in a form similar to a square wave. The maximum frequency of such a signal would be the pattern 1-0-1-0 such that for a signal with a data rate of 100 Mbps, the maximum frequency component would be 50 MHz. The next item to consider is the spectral content of the signal, which can be understood with the Fourier series

approximation of a square wave, shown below in Equations 1 and 2 in the Fourier series approximation.

From this it can be seen that a square wave consists of odd order harmonics and to fully construct a square wave n must go to infinity. However, to retain an acceptable portion of the waveform, the first two terms are generally sufficient. These two terms contain about 85% of the signal amplitude and allow a reasonable square wave to be reconstructed. Therefore, to reasonably pass a square wave of frequency x the minimum filter bandwidth necessary is 3x. All ON Semiconductor EMI filters are rated according to this principle. Attempting to violate this principle will result in significant rounding of the waveform and cause problems in transmitting the correct data. For example, take the filter with the response shown in Figure 5 and apply three different data waveforms. To calculate these three different frequencies, the 3 dB, 6 dB, and 9 dB bandwidths will be used.

Equation 1:

$$x(t) = \frac{1}{2} + \frac{2}{\pi} \sum_{n=1}^{\infty} \left[\frac{1}{2n-1} \sin((2n-1)\omega_0 t) \right] \quad (\text{eq. 1})$$

Equation 2 (simplified form of Equation 1):

$$x(t) = \frac{1}{2} + \frac{2}{\pi} \left[\frac{\sin(\omega_0 t)}{1} + \frac{\sin(3\omega_0 t)}{3} + \frac{\sin(5\omega_0 t)}{5} + \dots \right] \quad (\text{eq. 2})$$



Figure 5. Filter Bandwidth

From the above paragraphs it is shown that the maximum supported frequency of a waveform that can be passed through the filter can be found by dividing the bandwidth by a factor of three (to obtain the corresponding data rate

multiply the result by two). The following table gives the bandwidth values and the corresponding maximum supported frequencies and the third harmonic frequencies.

Table 2. Frequency Chart

Bandwidth	Maximum Supported Frequency	Third Harmonic Frequency
3 dB – 100 MHz	33.33 MHz (f_1)	100 MHz
6 dB – 200 MHz	66.67 MHz (f_2)	200 MHz
9 dB – 300 MHz	100 MHz (f_3)	300 MHz

Considering that 85% of the amplitude of the square is in the first two terms of the Fourier series approximation most of the signal content is at the fundamental (maximum supported) frequency and the third harmonic frequency. If a signal with a frequency of 33.33 MHz is input to this filter, the first two terms are sufficiently passed such that the signal is only mildly affected, as is shown in Figure 6a. If a signal

with a frequency of 66.67 MHz is input to this same filter, the third harmonic term is significantly attenuated. This serves to round the signal edges and skew the waveform, as is shown in Figure 6b. In the case that a 100 MHz signal is input to this filter, the third harmonic term is attenuated even further and results in even more rounding of the signal edges as is shown in Figure 6c. The result is the degradation of the data being transmitted making the digital data (1's and 0's) more difficult to discern. This does not include effects of other components such as interconnect and other path losses which could further serve to degrade the signal integrity. While some filter products may specify the 6 dB or 9 dB bandwidths, actually using these to calculate supported frequencies (and corresponding data rates) results in significant signal degradation. *To ensure the best signal integrity possible, it is best to use the 3 dB bandwidth to calculate the achievable data rate.*

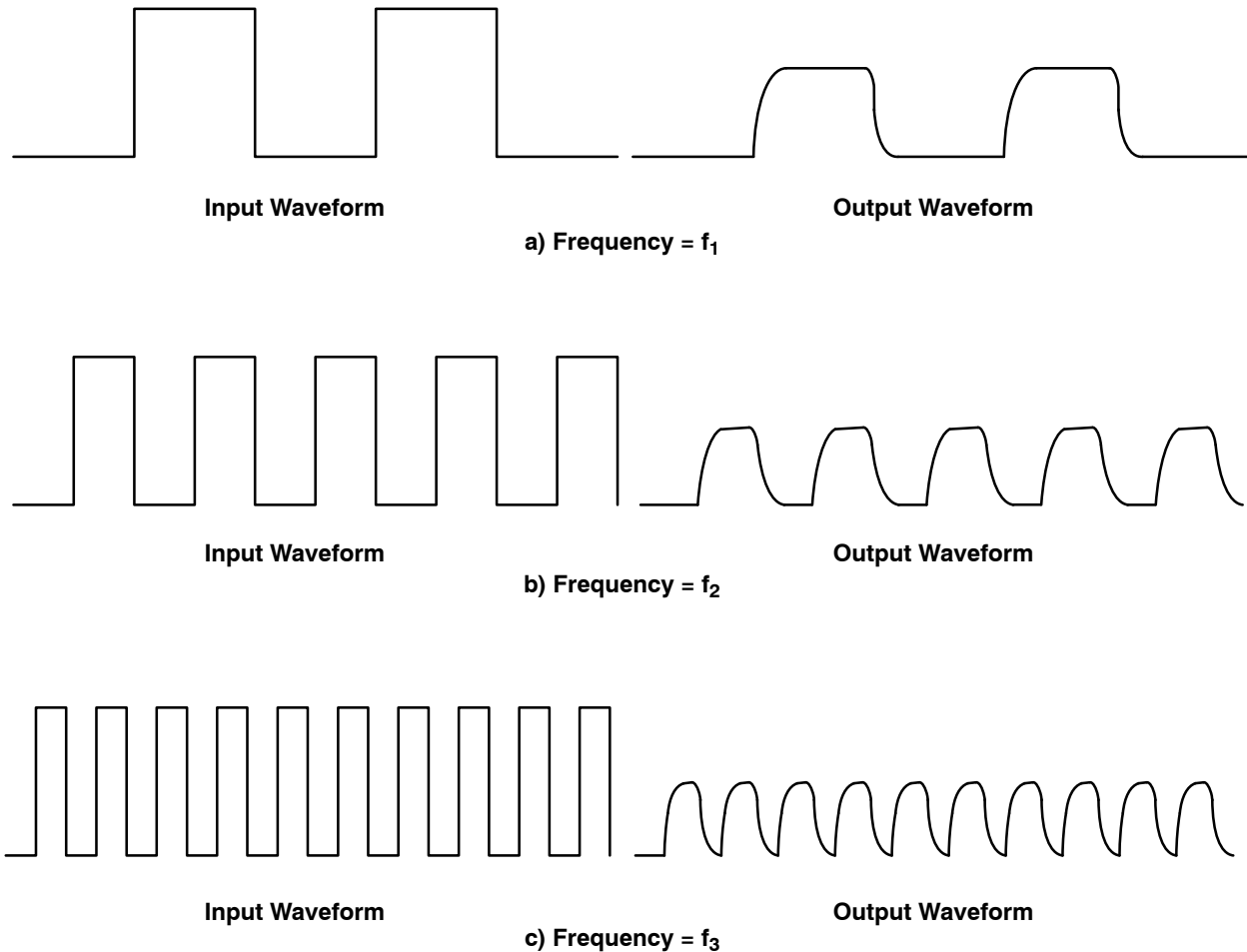


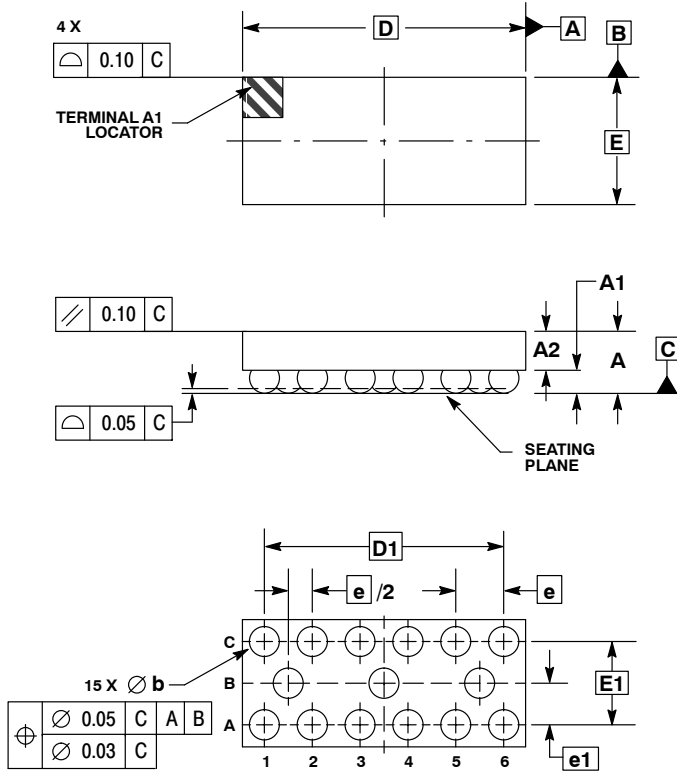
Figure 6. Input and Output Waveforms of Filter



15 PIN FLIPCHIP CSP
CASE 499D-01
ISSUE O

DATE 18 OCT 2002

SCALE 4:1



NOTES:

1. DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 1994.
2. CONTROLLING DIMENSION: MILLIMETER.
3. COPLANARITY APPLIES TO SPHERICAL CROWNS OF SOLDER BALLS.

MILLIMETERS		
DIM	MIN	MAX
A	---	0.700
A1	0.210	0.270
A2	0.380	0.430
D	2.960 BSC	
E	1.330 BSC	
b	0.290	0.340
e	0.500 BSC	
e1	0.435 BSC	
D1	2.500 BSC	
E1	0.870 BSC	

GENERIC
DEVICE MARKING



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